

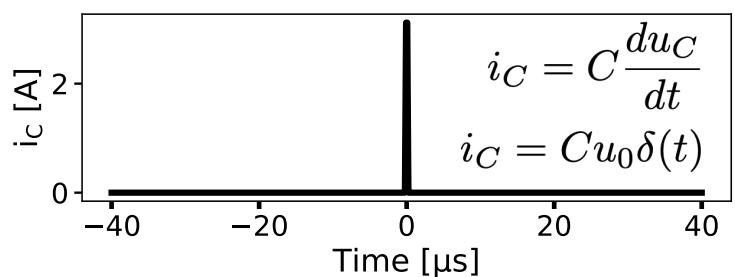
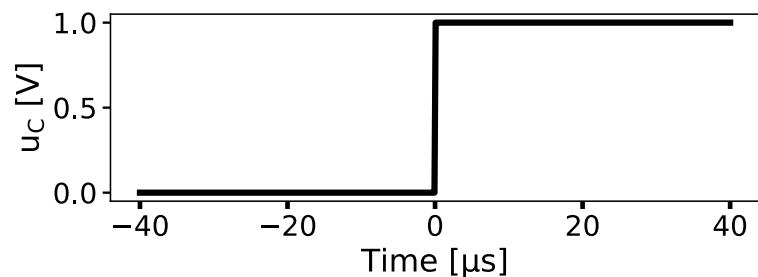
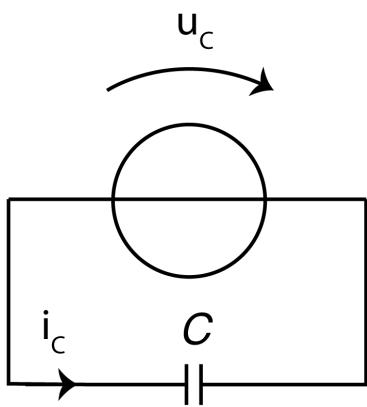
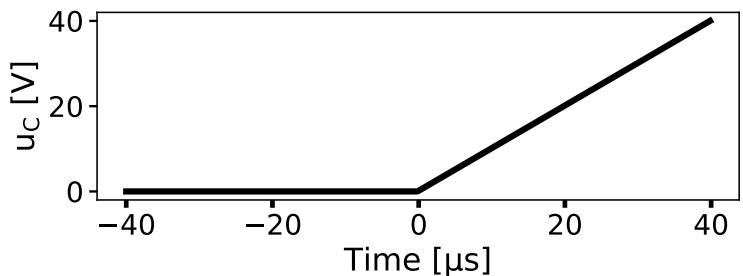
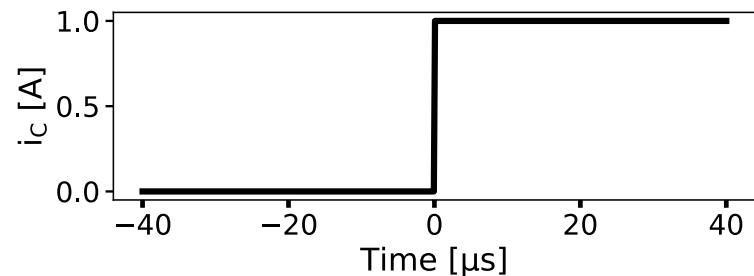
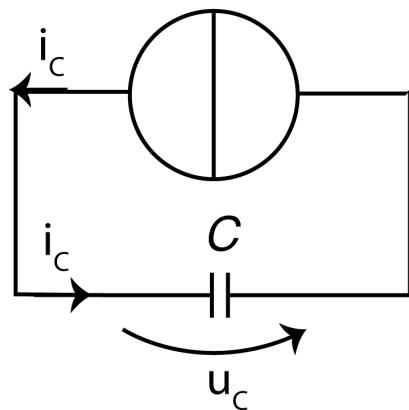
# Electrotechique-II

## MICRO-101: week 8

- What you have seen last time:
  - Time-dependent systems: Math recap (ch. 6.1), time-response of R and C (ch. 6.2.1 and 6.2.2) and RC circuits (ch. 6.3.1)
- Today:
  - Wrap up RC circuits (6.3.1)
  - Time-response of L (ch. 6.2.3.), RL circuits (ch. 6.3.2)
- Next time
  - Transient behavior with alternating voltages (ch. 6.4)
  - Solving of past exam (uploaded tomorrow to Moodle)

# 1-slide recap of last week

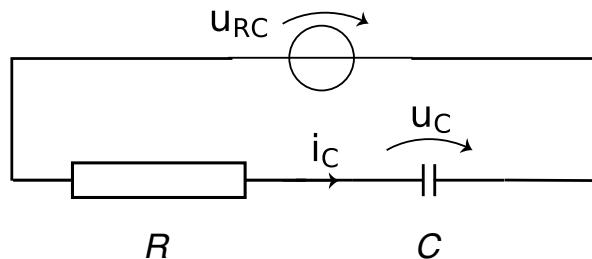
# C-circuit (Ch. 6.2.3)



# Today

# RC circuit (Ch. 6.3.1): the current

- Rule 1: In real systems (with non-zero resistance) – the voltage across a capacitor can not change suddenly (no voltage jump)



$$t < 0 :$$

$$i_C = 0$$

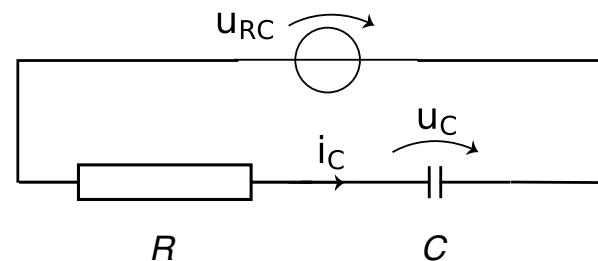
$$u_C = 0$$

$$t \geq 0 :$$

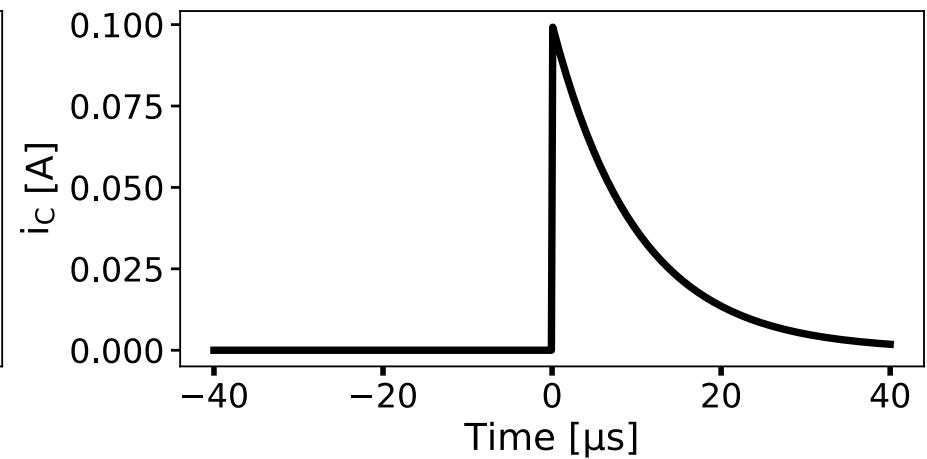
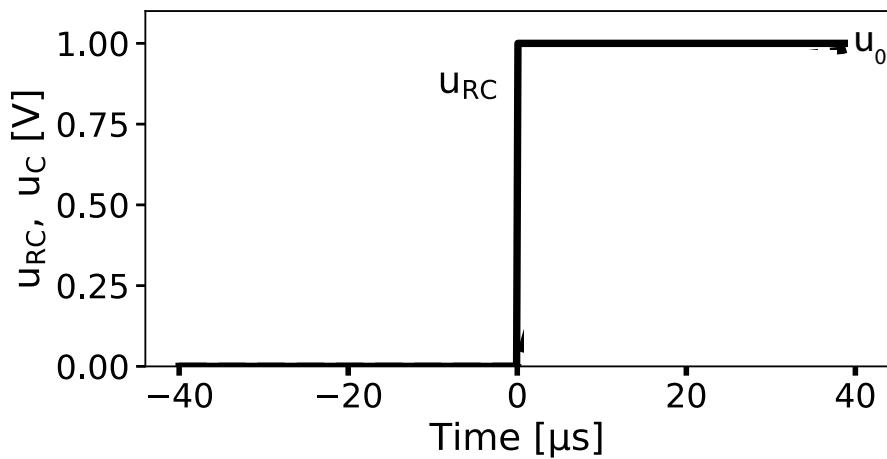
$$i_C = \frac{u_0}{R} e^{-\frac{t}{RC}}$$

# RC circuit (Ch. 6.3.1): the current

- Rule 1: In real systems (with non-zero resistance) – the voltage across a capacitor can not change suddenly (no voltage jump)

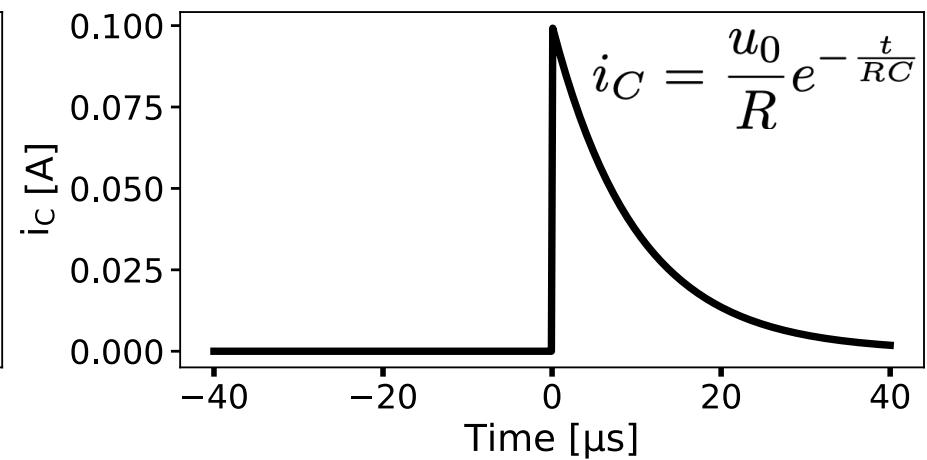
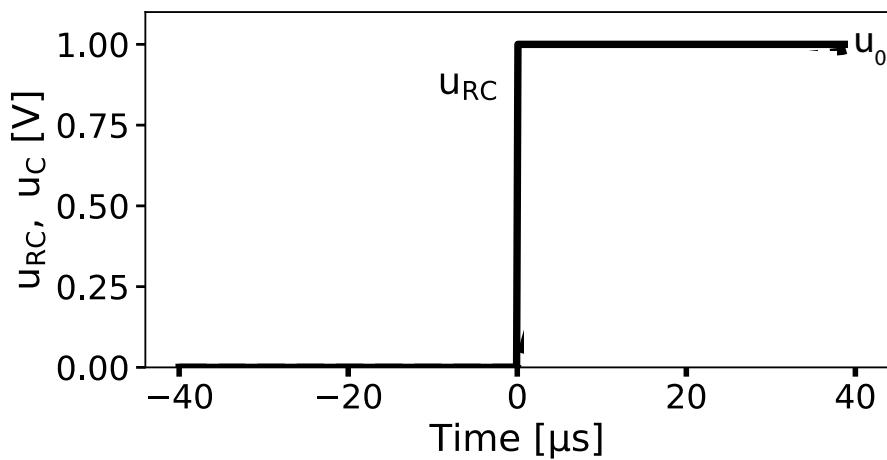


$$t \geq 0 : \quad i_C = \frac{u_0}{R} e^{-\frac{t}{RC}}$$



# RC circuit (Ch. 6.3.1): the current

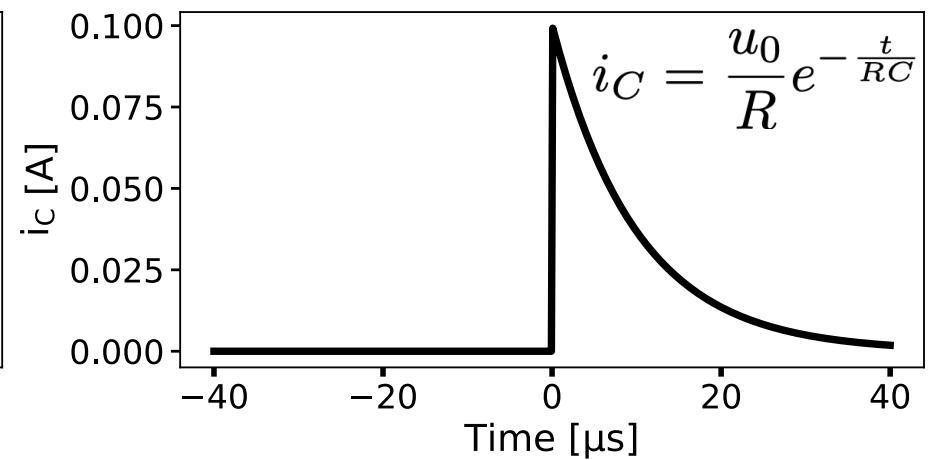
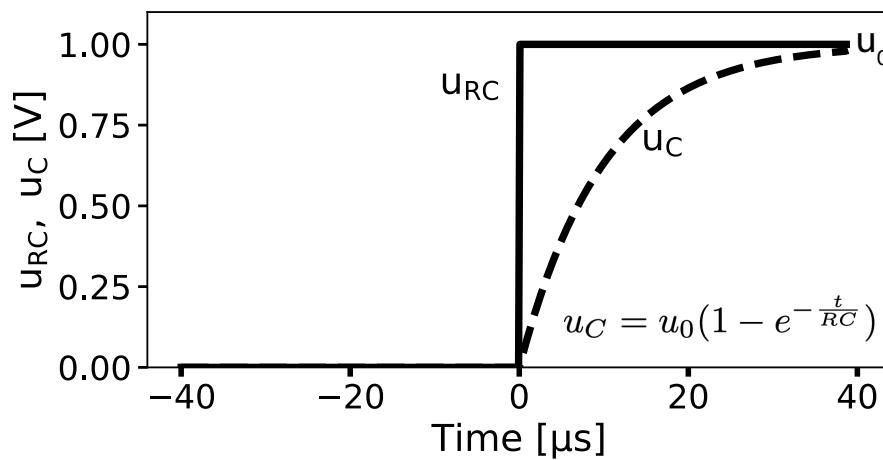
- Rule 1: In real systems (with non-zero resistance) – **the voltage across a capacitor can not change suddenly (no voltage jump)**
- The current spikes to finite value at  $t = 0$ :  $i_C(t = 0) = \frac{u_0}{R}$
- $i_C$  then decays with characteristic time  $\tau_{RC} = RC$ .
- The smaller  $R$ , the shorter  $\tau_{RC}$ , the higher  $i_C(t = 0) \rightarrow$  the more the system converges towards what we've seen for the C-circuit



# RC circuit (Ch. 6.3.1): the voltage

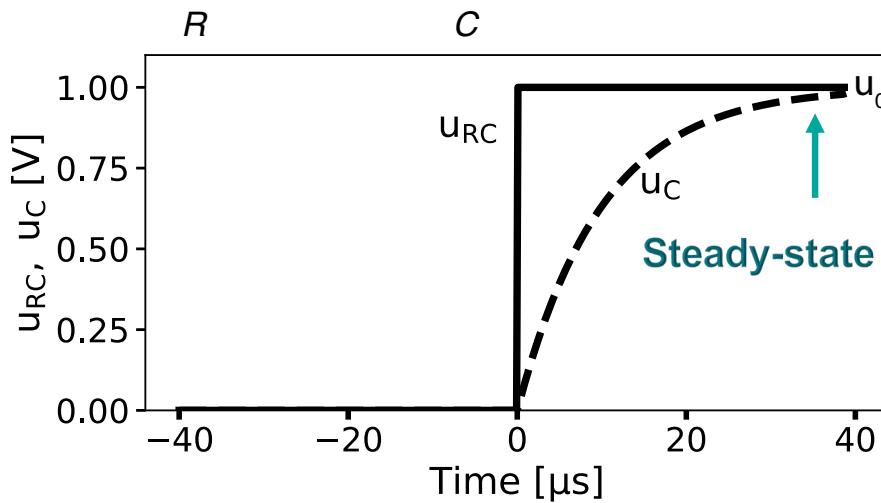
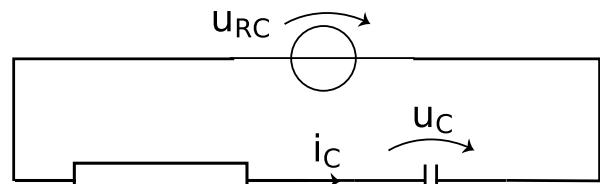
- Rule 1: In real systems (with non-zero resistance) – **the voltage across a capacitor can not change suddenly (no voltage jump)**
- No discontinuity in the voltage
- $u_C$  then increases with  $\tau_{RC}$  towards the value of the source.
- The smaller  $R$ , the shorter  $\tau_{RC}$ , the faster  $u_C$  converges towards  $u_{RC} \rightarrow$  what we've seen for the C-circuit

■ MICRO-101



# Overview RC circuit (Ch. 6.3.1)

- Guided example 1:



$$t < 0 :$$

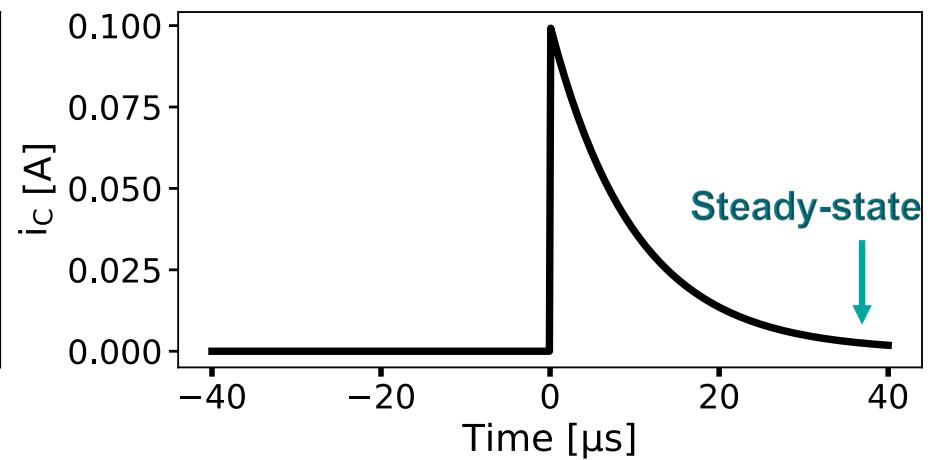
$$i_C = 0$$

$$u_C = 0$$

$$t \geq 0 :$$

$$i_C = \frac{u_0}{R} e^{-\frac{t}{RC}}$$

$$u_C = u_0 \left( 1 - e^{-\frac{t}{RC}} \right)$$

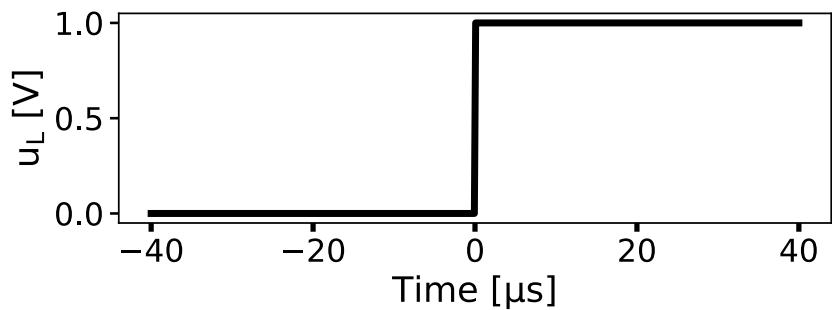
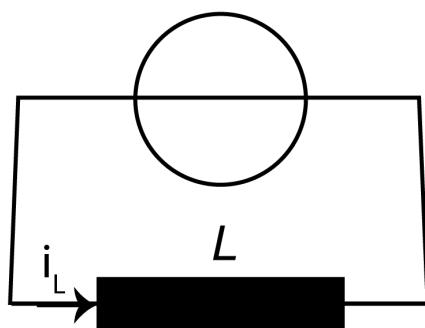


# Time response of L (Ch. 6.2.3)

Voltage source connected

$$u_L(t) = u_0 h(t)$$

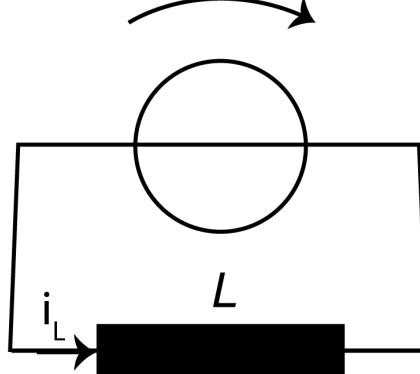
$$u_L(t) = L \frac{di_L}{dt}$$



# Time response of L (Ch. 6.2.3)

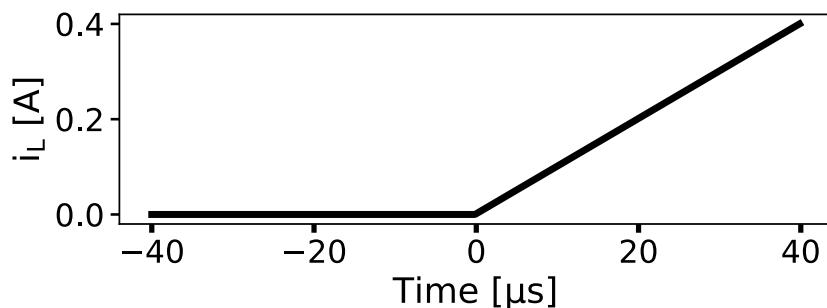
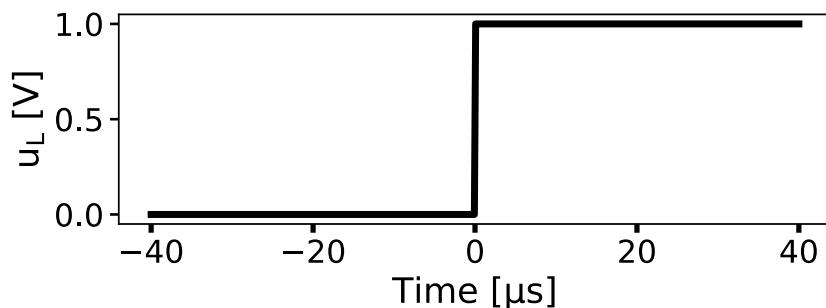
Voltage source connected

$$u_L(t) = u_0 h(t)$$



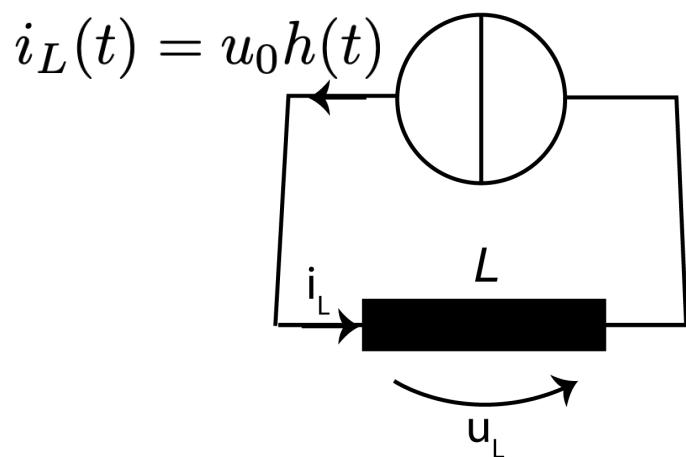
$$u_L(t) = L \frac{di_L}{dt}$$

Current does not react instantaneously to change in voltage:  
It slowly ramps up until it reaches infinity.  
For long times, the inductor behaves like a short.

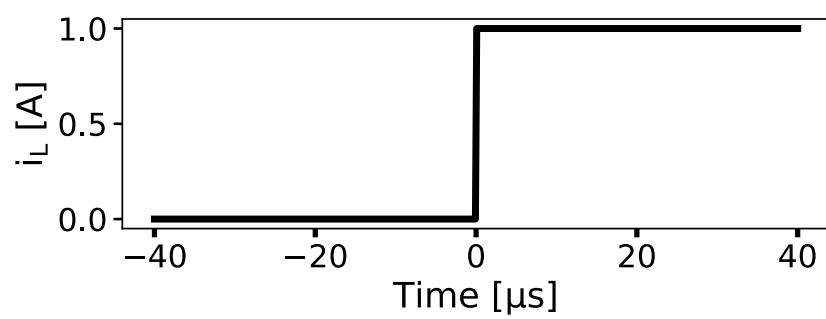


# Time response of L (Ch. 6.2.3)

Current source connected

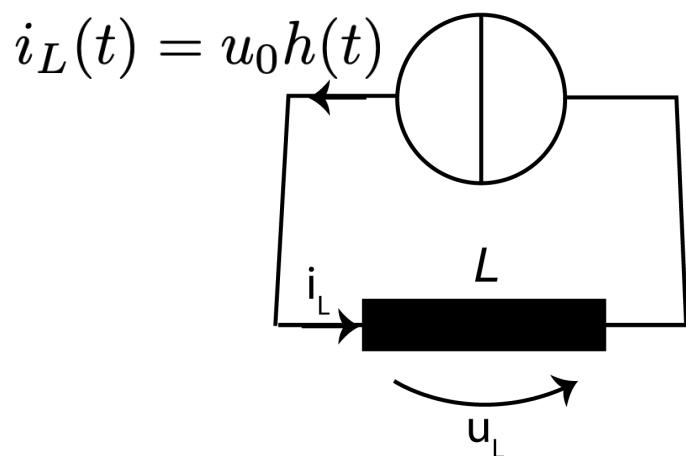


$$u_L(t) = L \frac{di_L}{dt}$$



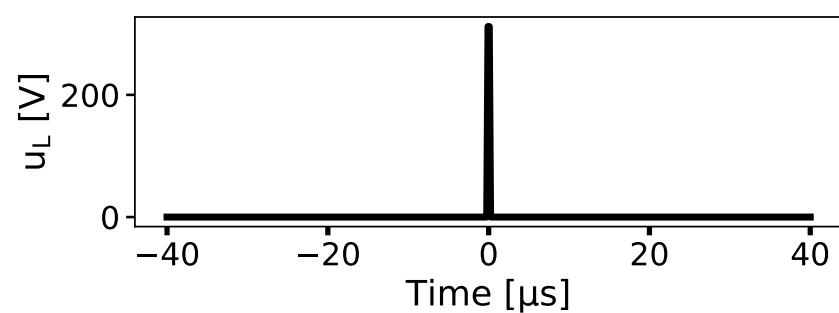
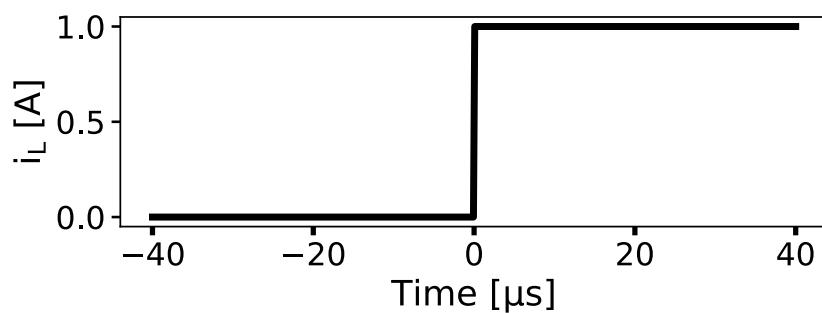
# Time response of L (Ch. 6.2.3)

Current source connected



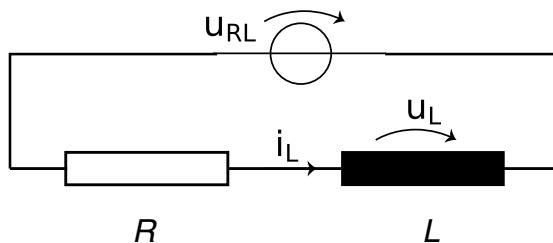
$$u_L(t) = L \frac{di_L}{dt}$$

Voltage spikes in reaction to change in current, so that the current flowing across the inductor equals the one of the source.



# RL circuit (Ch. 6.3.2)

- Guided example 2:



$$u_{RL} = u_0 h(t)$$

- Boundary conditions:

$$t < 0$$

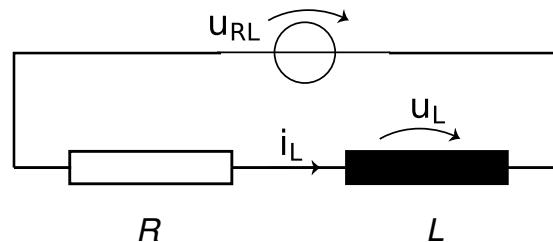
(calculation at the blackboard)

$$i_L = 0 \quad \text{No current flowing through the circuit!}$$

$$u_L = 0 \quad \text{No voltage drop across the inductor}$$

# RL circuit (Ch. 6.3.2)

- Rule 2: In real systems (with non-zero resistance) – the current flowing through an inductor can not change suddenly (no current jump)



- Boundary conditions:

$$t < 0$$

$$i_L = 0$$

$$u_L = 0$$

$$t \geq 0$$

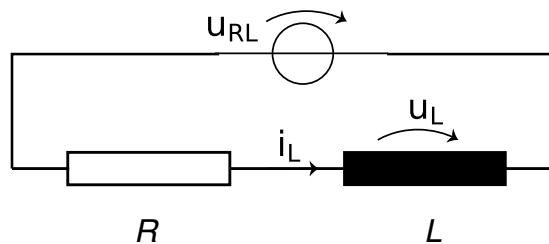
$$i_L = \frac{u_0}{R} \left(1 - e^{-\frac{tR}{L}}\right)$$

$$u_L = L \frac{di_L}{dt} = u_0 e^{-\frac{tR}{L}}$$

# RL circuit (Ch. 6.3.2)

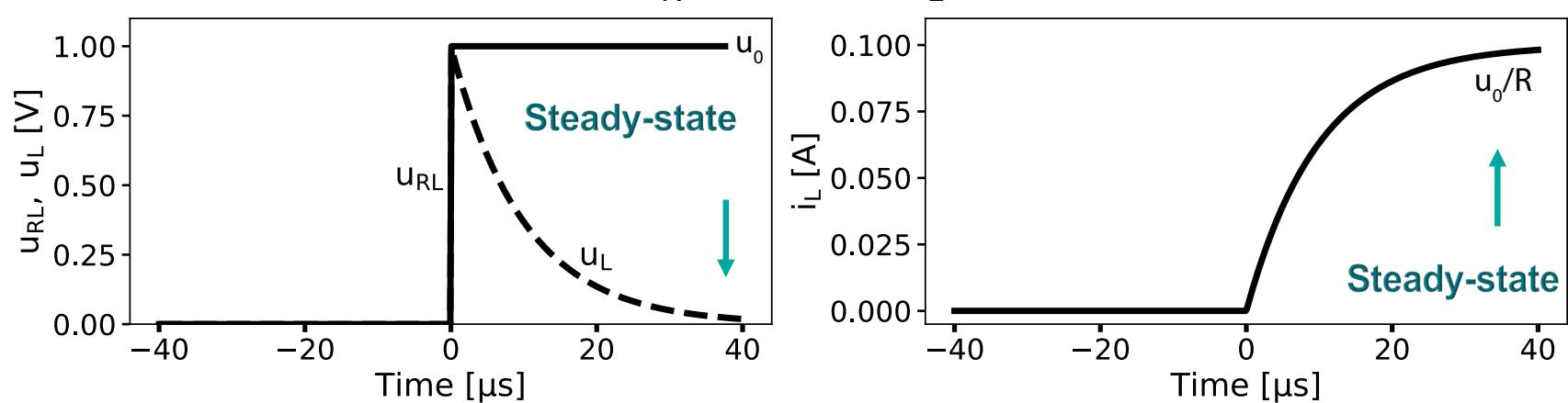
$$t \geq 0$$

- Guided example 2:



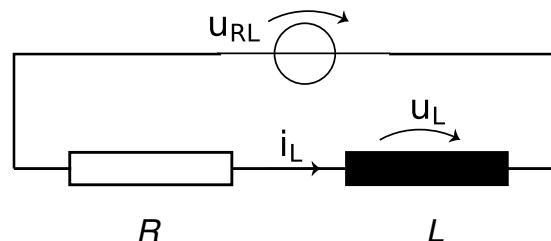
$$i_L = \frac{u_0}{R} \left(1 - e^{-\frac{tR}{L}}\right)$$

$$u_L = L \frac{di_L}{dt} = u_0 e^{-\frac{tR}{L}}$$



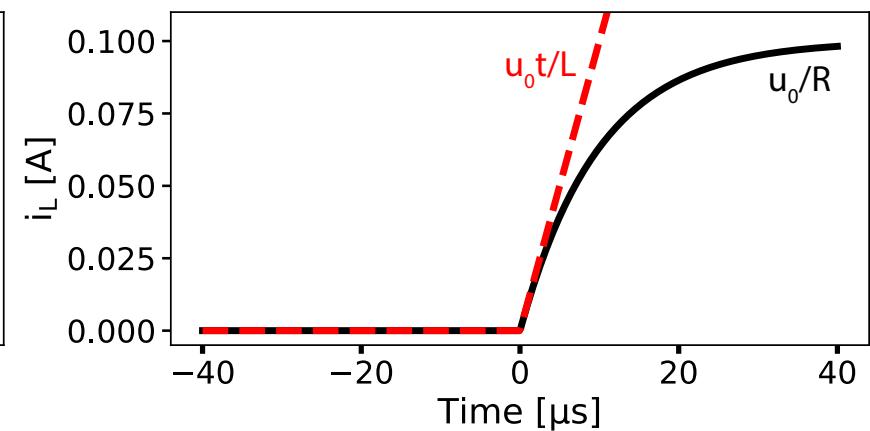
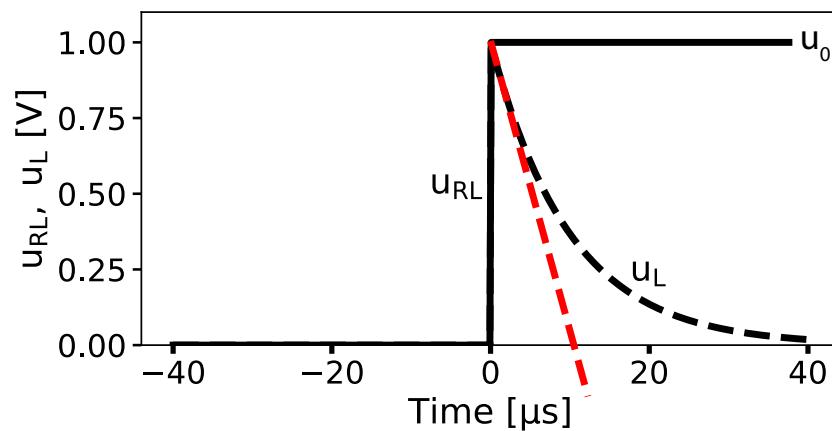
# RL circuit (Ch. 6.3.2)

- Rule 2: In real systems (with non-zero resistance) – the current flowing through an inductor can not change suddenly (no current jump)  $t \geq 0$



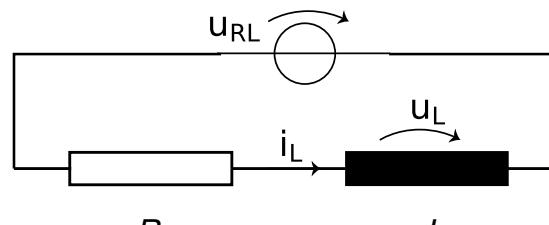
$$i_L = \frac{u_0}{R} (1 - e^{-\frac{tR}{L}})$$

$$u_L = L \frac{di_L}{dt} = u_0 e^{-\frac{tR}{L}}$$



# RL circuit (Ch. 6.3.2)

- Rule 2: In real systems (with non-zero resistance) – **the current flowing through an inductor can not change suddenly (no current jump)**



$$t \geq 0$$

$$i_L = \frac{u_0}{R} \left(1 - e^{-\frac{tR}{L}}\right)$$

$$u_L = L \frac{di_L}{dt} = u_0 e^{-\frac{tR}{L}}$$

We find that the current ramps up with a characteristic time constant  $\tau_{RL} = \frac{L}{R}$ . The RL time constant can be reduced for example by decreasing L.

At  $t = 0$ , the voltage spikes to a value of  $u_L(t = 0) = u_0$  after which it decreases again with the same slope as the current. We note that for  $L \rightarrow 0$  we find back the case of the resistor connected to a voltage source which is switched on.