

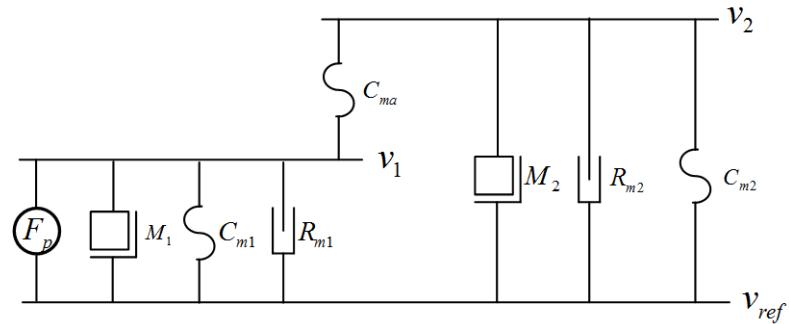
## Chapter IV - Electroacoustic systems and radiation

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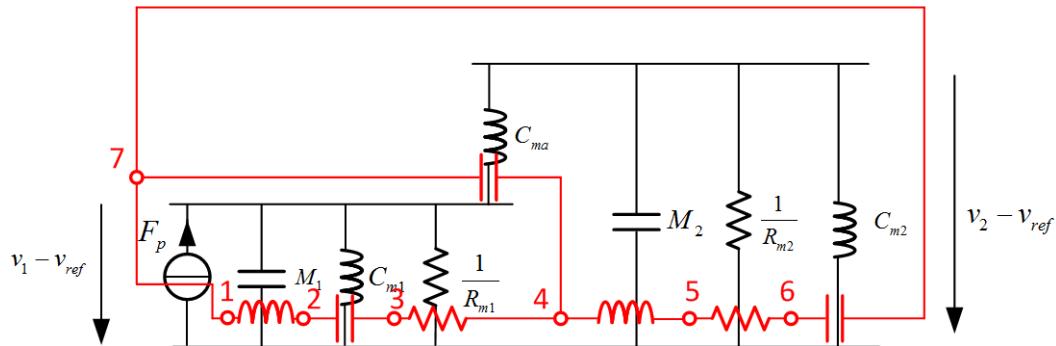
Fall semester 2017

### Exercise 1. Double panel partition

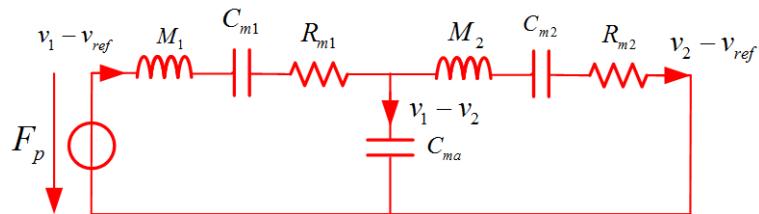
1. We see two velocities  $v_1$  and  $v_2$  corresponding to the two individual panels (masses  $M_1$  and  $M_2$ ), and we also add a line for the reference velocity  $v_{ref} = 0$ . We obtain the following symbolic scheme (velocities  $v_i$  corresponding to "potentials") :



2. We deduce the inverse scheme (in black) :



And finally the direct scheme (see the red scheme overlaying the black inverse scheme) :



3.  $H = \frac{v_2}{F_p}$ , since  $Z_{meq} = C_{ma} \parallel (R_{m2}, C_{m2}, M_2) = \frac{j\omega M_2 + R_{m2} + \frac{1}{j\omega C_{m2}}}{1 + j\omega C_{ma}(j\omega M_2 + R_{m2} + \frac{1}{j\omega C_{m2}})}$ , we get :  $\frac{Z_{meq}}{Z_{meq} + (j\omega M_1 + R_{m1} + \frac{1}{j\omega C_{m1}})} F_p = (j\omega M_2 + R_{m2} + \frac{1}{j\omega C_{m2}}) v_2$ .

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$$\text{Then, } H = \frac{1}{j\omega M_2 + R_{m2} + \frac{1}{j\omega C_{m2}}} \frac{Z_{meq}}{Z_{meq} + (j\omega M_1 + R_{m1} + \frac{1}{j\omega C_{m1}})}$$

$$\text{Finally : } H = \frac{1}{j\omega M_2 + R_{m2} + \frac{1}{j\omega C_{m2}} (1 + j\omega C_a (j\omega M_1 + R_{m1} + \frac{1}{j\omega C_{m1}})) + j\omega M_1 + R_{m1} + \frac{1}{j\omega C_{m1}}}.$$

## Exercise 2. Helmholtz resonators

$f_s = \frac{1}{2\pi\sqrt{m_a C_a}} = \frac{c}{2\pi} \sqrt{\frac{S}{VL}}$  The resonance frequencies can be arranged in the following order :  
 $f_b < f_d < f_a < f_c$

## Exercise 3. Silencer

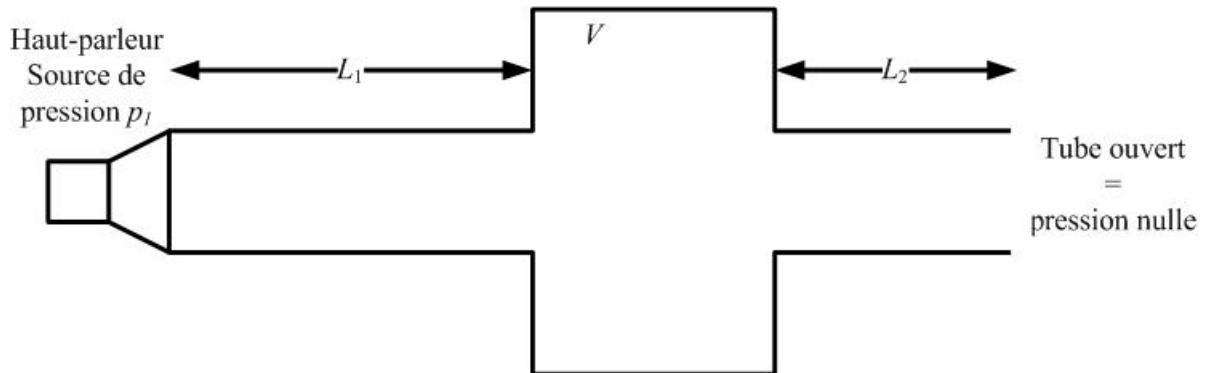


Figure 1 – Schematic representation of a silencer.

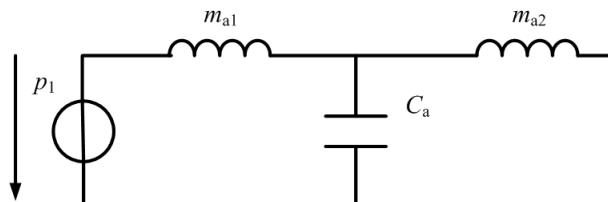


Figure 2 – Equivalent circuit of the silencer.

The acoustic mass in the duct of length  $L_1$  is given by  $m_{a1} = \rho L_1 / (\pi r_d^2)$ , and that the duct of length  $L_2$  by  $m_{a2} = \rho L_2 / (\pi r_d^2)$ . The effective acoustic compliance in the coupling cavity is given by  $C_a = V / (\rho c^2)$ . The analog acoustic scheme is illustrated in Fig. 7.

The input impedance of the silencer can be derived as

$$Z_a = \frac{p_1}{q_1} = \frac{(j\omega^2)m_{a1}C_a + \frac{m_{a1}}{m_{a2}} + 1}{j\omega C_a + \frac{1}{j\omega m_{a2}}}$$

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**Exercise 4. Boomwhacker**

$$1. \begin{cases} \frac{\partial p}{\partial x} = -\frac{\rho_0}{S} \frac{\partial q}{\partial t} \\ \frac{\partial q}{\partial x} = -\chi_s S \frac{\partial p}{\partial t} \end{cases}$$

$$\text{Then } \frac{\partial^2 p}{\partial x^2} - \frac{1}{c_0^2} \frac{\partial^2 p}{\partial t^2} = 0, \text{ where } c_0 = \frac{1}{\sqrt{\rho_0 \chi_s}}$$

2. If we derive  $p(x, t) = P(x)e^{j\omega t}$  in the wave equation, we obtain the Helmholtz equation  $\frac{d^2 P}{dx^2} + k^2 P = 0$ , where  $k = \frac{\omega}{c_0}$

The solution of the wave equations are then of the form :

$$P(x) = P_{0+}e^{-jkx} + P_{0-}e^{+jkx}.$$

Taking the first equation (Euler's generalized equation), and since  $\frac{\partial q}{\partial t} = j\omega Q(x)e^{j\omega t}$ , we deduce :

$$Q(x) = \frac{1}{Z_{ac}} (P_{0+}e^{-jkx} - P_{0-}e^{+jkx}).$$

3.

$$\forall x, \begin{cases} P(x) = P_{0+}e^{-jkx} + P_{0-}e^{jkx} \\ Q(x) = \frac{S}{\rho_0 c_0} (e^{-jkx} - P_{-}e^{jkx}) \end{cases}$$

If we denote  $P_+(x) = P_{0+}e^{-jkx}$  and  $P_-(x) = P_{0-}e^{jkx}$ , we can write the diffusion relationship :

$$\begin{pmatrix} P(x) \\ Q(x) \end{pmatrix} = \begin{bmatrix} 1 & 1 \\ \frac{1}{Z_{ac}} & -\frac{1}{Z_{ac}} \end{bmatrix} \begin{pmatrix} P_+(x) \\ P_-(x) \end{pmatrix}$$

$$\text{Then : } \begin{pmatrix} P(x-L) \\ Q(x-L) \end{pmatrix} = \begin{bmatrix} e^{jkL} & e^{-jkL} \\ \frac{e^{jkL}}{Z_{ac}} & -\frac{e^{-jkL}}{Z_{ac}} \end{bmatrix} \begin{pmatrix} P_+(x) \\ P_-(x) \end{pmatrix}$$

We can also invert the diffusion matrix and it yields :

$$\begin{pmatrix} P_+(x) \\ P_-(x) \end{pmatrix} = \frac{Z_{ac}}{2} \begin{bmatrix} -\frac{1}{Z_{ac}} & -1 \\ -\frac{1}{Z_{ac}} & 1 \end{bmatrix} \begin{pmatrix} P(x) \\ Q(x) \end{pmatrix}$$

And finally :

$$\begin{pmatrix} P(x-L) \\ Q(x-L) \end{pmatrix} = \begin{bmatrix} e^{jkL} & e^{-jkL} \\ \frac{e^{jkL}}{Z_{ac}} & -\frac{e^{-jkL}}{Z_{ac}} \end{bmatrix} \cdot \frac{Z_{ac}}{2} \begin{bmatrix} -\frac{1}{Z_{ac}} & -1 \\ -\frac{1}{Z_{ac}} & 1 \end{bmatrix} \begin{pmatrix} P(x) \\ Q(x) \end{pmatrix} = \begin{bmatrix} \cos kL & jZ_{ac} \sin kL \\ \frac{j}{Z_{ac}} \sin kL & \cos kL \end{bmatrix} \begin{pmatrix} P(x) \\ Q(x) \end{pmatrix}$$

which also holds for  $x = L$ .

4. The duct is closed at the right termination, then  $Q(L) = 0$ .

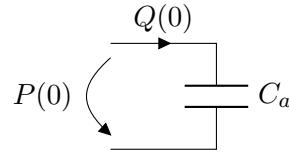
$$\text{Then } Z_a(0) = \frac{P(0)}{Q(0)} = -jZ_{ac} \cot kL$$

5. If the left termination is open ( $P(0) = 0$ ), then we should have  $Z_a(0) = 0$ . The resonance frequencies correspond then to  $\cot kL = 0$ , then  $f_n = \frac{(2n+1)c_0}{4L}$

For a length of 19 cm, the first resonance frequency occurs at  $f_1 = 447$  Hz (almost  $A_{440}$ ).

6. at low frequencies ( $kL \ll 1$ ),  $Z_a(0) \approx -j \frac{Z_{ac}}{kL} = \frac{1}{j\omega \frac{V}{\rho_0 c_0^2}}$ . The low-frequency behavior is

an acoustic compliance  $C_a = \frac{V}{\rho_0 c_0^2}$ , and the acoustical scheme is the following :



### Exercise 5. Bi-directional and cardioid sources

The directivity factor is computed as :  $\Delta = \frac{4\pi}{\int_{\phi=0}^{2\pi} \int_{\theta=0}^{\pi} D^2(\theta, \phi) \sin \theta d\theta d\phi}$  (referred to an omnidirectional source with solid angle  $4\pi$ ).

1. Here  $D(\theta, \phi) = \cos \theta$ , then  $\Delta = \frac{4\pi}{2\pi \int_{\theta=0}^{\pi} \cos^2(\theta) \sin \theta d\theta}$ .

Integrating by parts,  $u = \cos(\theta)$  and  $du = -\sin(\theta)d\theta$ , one gets :  $\Delta = 3$  or  $L_\Delta = 4.8$  dB. Half-power beamwidth :  $D^2(\theta) = \frac{1}{2} \rightarrow \theta_{-3dB} = 45^\circ$ , the total aperture (beamwidth) is then  $90^\circ$  (symmetry over the axis)

2. Same as before :  $\Delta = 3$  or  $L_\Delta = 4.8$  dB, and  $\theta_{-3dB} = 45^\circ$ .

### Exercise 6. Directivity of a loudspeaker

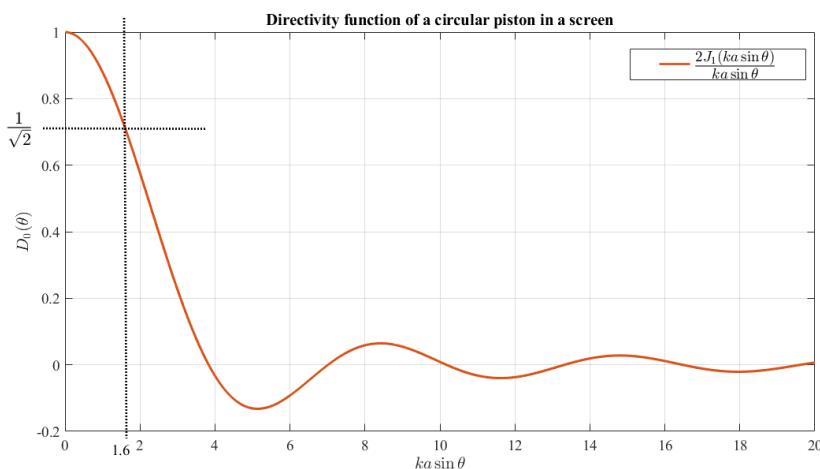


Figure 3 – Directivity function of a piston on screen

We assume that a source remains "omnidirectional" (with a tolerance of 3 dB) if its directivity function remains higher than  $D_0(\theta) \geq \frac{1}{\sqrt{2}}$ ,  $\forall \theta \in [-\frac{\pi}{2}, \frac{\pi}{2}]$  (since  $20 \cdot \log_{10} \frac{1}{\sqrt{2}} = -3$  dB(re.  $D_0(0) = 1$ )). According to the figure above, that means that  $ka \sin \theta_{min} = 1.6$  (where  $\theta_{min}$  designates the angle at which the directivity reaches  $\frac{1}{\sqrt{2}}$ ).

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1. For  $f = 500$  Hz and  $\theta_{min} = 90^\circ$ ,  $a \leq \frac{c}{f \sin \theta_{min}} = 21,8$  cm (here we consider  $\theta_{min} = \pi/2$  so that the half-bandwidth angle is  $\pm 90^\circ$ , ie. omnidirectional)
2. For  $f = 2000$  Hz and  $\theta_{min} = 90^\circ$ ,  $a \leq 5,46$  cm (same assumption)
3. For  $f = 1000$  Hz and  $\theta_{min} = 30^\circ$ ,  $a \leq 21,8$  cm (here the half-bandwidth should be  $\pm 30^\circ$ ).

## Exercise 7. Radiation of a 2-way loudspeaker - monopole hypothesis

Reminder :

Field radiated by a monopole :

$$\underline{p}_M(r) = j \frac{\rho c}{4\pi r} k \underline{q} e^{-jkr}$$

Field radiated by a monopole on a closed box (semi-monopole) :

$$\underline{p}_E(r) = j \frac{\rho c}{2\pi r} k \underline{q} e^{-jkr}$$

1. Monopole on a closed box with flow velocity  $q_1$  or  $q_2$  :

$$q_1 = \frac{2\pi r_1}{\rho c k} p_{M1}(r_1) = \frac{2\pi r_1}{\rho c 2\pi f / c} \cdot (20 \cdot 10^{-6} \cdot 10^{\frac{L_{p1}}{20}}) = 4,9 \cdot 10^{-3} \text{ m.s}^{-3}$$

$$q_2 = 3,9 \text{ m.s}^{-3}$$

If synchronous source, microphone  $\sim$  in the loudspeaker's main axis, with wavelength  $\lambda \sim 1\text{m}$

$\Rightarrow$  the pressure waves are in phase and therefore :

$$p_{1+2} = p_1 + p_2 \Rightarrow L_{p_{1+2}} = 20 \cdot \log \left[ 10^{\frac{L_{p1}}{20}} + 10^{\frac{L_{p2}}{20}} \right] = 93 \text{ dB}$$

2. At 350 Hz, the boomer becomes slightly directive  $\Rightarrow$  attenuation of the off-axis measurements. For the medium however, no changes.

If measurement points are off-axis, the phase difference varies between the two signals  $\Rightarrow$  a quadratic summation must be done (of the energy) and the expression before changes to be :

$$L_{p_{1+2}} = 10 \cdot \log \left[ 10^{\frac{L_{p1}}{10}} + 10^{\frac{L_{p2}}{10}} \right] = 89,6 \text{ dB}$$

## Exercise 8. Radiation of a small speaker

1. At low-frequencies,  $R_{ar} \approx \frac{\rho c}{\pi a^2} \frac{(ka)^2}{2}$ . It is valid until  $ka \approx 1$  then for  $f < \frac{c}{2\pi a} = 909$  Hz.
2. On axis, the sound pressure reads  $p(r, \theta = 0) = j \frac{\rho c}{2\pi r} k \underline{q} e^{-jkr}$ , where  $q$  is the volume velocity, linked to the excursion  $\xi$  as  $q = j\omega(\pi a^2)\xi$ .

Then (if we denote  $\tilde{\nu}$  the rms value of quantity  $\nu$ ) :  $\tilde{p}(r, 0) = \frac{\rho \omega^2(\pi a^2)}{2\pi r} \tilde{\xi} = \frac{\sqrt{2}\rho\omega^2 a^2}{4r} \xi_{\max}$  (here the peak-to-peak elongation is given).

- at 250 Hz,  $\tilde{p}(r = 10\text{m}) = 0.158$  Pa or  $L_p \approx 78$  dB (re. 20  $\mu\text{Pa}$ )
- at 500 Hz,  $\tilde{p}(r = 10\text{m}) = 0.633$  Pa or  $L_p \approx 90$  dB (re. 20  $\mu\text{Pa}$ )
- at 1000 Hz, we should (theoretically) not consider the low-frequency approximation. However, the actual limitation corresponds to  $ka = \sqrt{2}$  (instead of  $ka = 1$ ), then the frequency bound is  $f_{\max} \approx 1287\text{Hz}$ . In this case we can still consider the low-frequency approximation.

$$\tilde{p}(r = 10\text{m}) = 2.5 \text{ Pa or } L_p \approx 102 \text{ dB (re. 20 } \mu\text{Pa)}$$