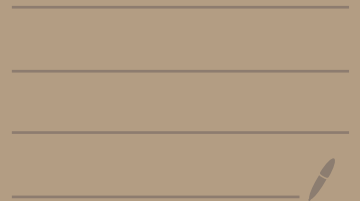


Principles of Digital Communication

April 28, 2021



Reading was 5. [1-4].

Motivation:

$$i: k \text{ bit} \xrightarrow{\quad} c_i \xrightarrow{\quad} r(t) = \sum_i c_i \psi_i(t) \xrightarrow{\quad}$$

$c_i \in \mathbb{C}^n$

$\{\psi_i\}$ an L -normal basis.

$R \in R1$

$$\xrightarrow{\quad} R(t)$$

$\langle R, \psi_1 \rangle$
 $\vdots$
 $\langle R, \psi_m \rangle$

requires $m = 2^k$ inner products

$R \in R2$

$$\xrightarrow{\quad} R(t)$$

$\langle R, \psi_1 \rangle$
 $\vdots$
 $\langle R, \psi_n \rangle$

requires n inner products

if it is also the case that

$\{\psi_i\}$ are not only L -normal but
time shifts of each other

$$\xrightarrow{\quad} R(t) \quad \boxed{\psi^*(-t)} \quad \xrightarrow{\quad}$$

sample at n time instants

$t_1, \dots, t_n$ s.t

$$\psi_j(t) = \psi(t - t_j), \quad j=1, \dots, n.$$

Based on these considerations we are led to the following question: is there a simple test which tells us if the integer shifts of $\psi(\cdot)$ form an L^2 -normal set?

$\equiv$ we want to test

$$\int \psi^*(t) \psi(t-n) dt = \begin{cases} 1 & n=0 \\ 0 & n=\pm 1, \pm 2, \pm 3, \dots \end{cases}$$

$$x(t) = \psi(t) \iff x_F(f) = \psi_F(f)$$

$$y(t) = \psi(t-n) \iff y_F(f) = \psi_F(f) \cdot e^{-j2\pi n f}$$

Aside $y_F(f) = \int y(t) e^{-j2\pi f t} dt$

$$= \int \psi(t-n) e^{-j2\pi f t} dt$$

$$= \int \psi(t-n) e^{-j2\pi f(t-n+n)} dt$$

$$= e^{-j2\pi f n} \int \psi(t-n) e^{-j2\pi f(t-n)} dt$$

$$= \int \psi(t) e^{-j2\pi f t} dt$$

$$= e^{-j2\pi f n} \cdot \psi_F(f)$$

L^2 -normality test

$$\equiv \int \psi_F^*(f) \psi_F(f) e^{-j2\pi f n} df = \begin{cases} 1 & n=0 \\ 0 & n=\pm 1, \pm 2, \pm 3, \dots \end{cases}$$

$$\equiv \int |\psi_F(t)|^2 e^{-j2\pi f n} df = \mathbb{I}\{n=0\}$$

$$\equiv \sum_{\substack{k \text{ integer} \\ -\frac{1}{2} + k \\ \frac{1}{2} + k}} \int |\psi_F(t)|^2 e^{-j2\pi f n} df = \mathbb{I}\{n=0\}$$

$$\equiv \sum_k \int_{-\frac{1}{2}}^{\frac{1}{2}} |\psi_F(t+k)|^2 e^{-j2\pi(f+k)n} df = "$$

$$\equiv \sum_k \int_{-\frac{1}{2}}^{\frac{1}{2}} |\psi_F(t+k)|^2 e^{-j2\pi f n} df = \mathbb{I}\{n=0\}$$

$$\equiv \int_{-\frac{1}{2}}^{\frac{1}{2}} \left(\sum_k |\psi_F(t+k)|^2 \right) e^{-j2\pi f n} df = \underline{\underline{\mathbb{I}\{n=0\}}}$$

$g(t)$

note that $f \mapsto g(t)$ is a periodic function with period 1.

so $g(t) = \sum_n g_n e^{j2\pi f n}$ (discrete Fourier series).

$$\text{with } g_n = \int_{-\frac{1}{2}}^{\frac{1}{2}} g(t) e^{-j2\pi f n} df$$

We now see that our condition for $\{\psi(t-n)\}$ to be 1-normal family $\Rightarrow$

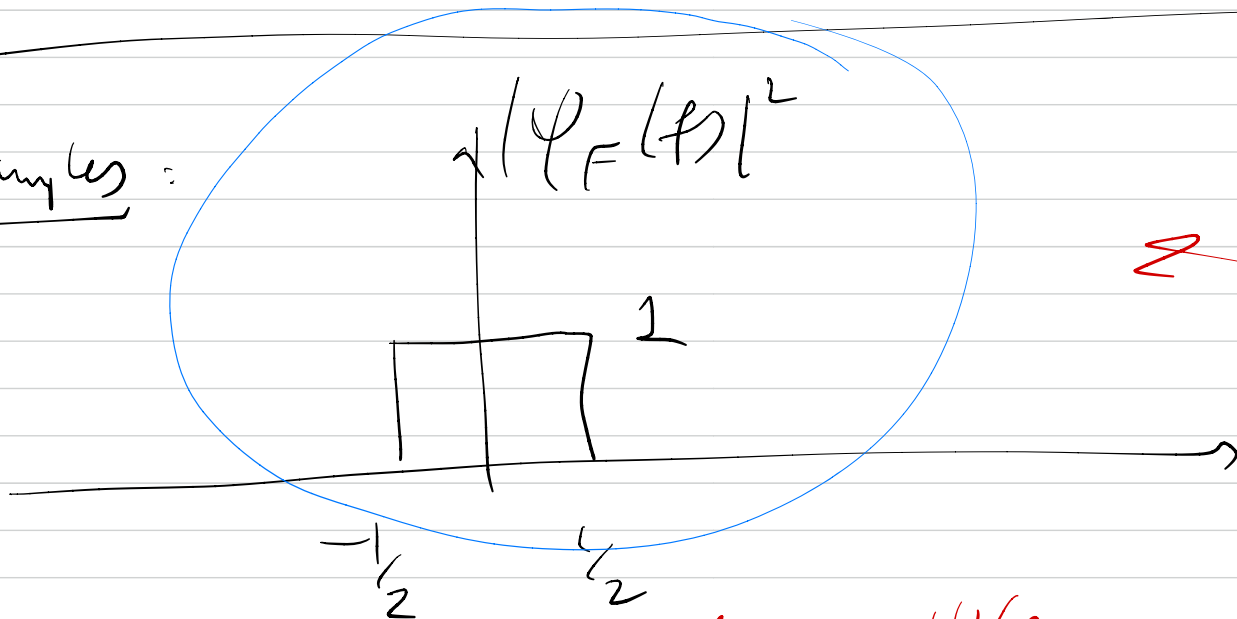
$$g_0 = 1 \quad g_1 = g_{-1} = g_2 = g_{-2} \dots = 0,$$

i.e. $g(t) = 1$. Consequently we have the following test:

Theorem (Nyquist Criterion): $\{\psi_n = \psi(t-n) : n \in \mathbb{Z}\}$ form an 1-normal collection if and only if

$$\sum_n |\psi_F(t-n)|^2 = 1.$$

Examples:

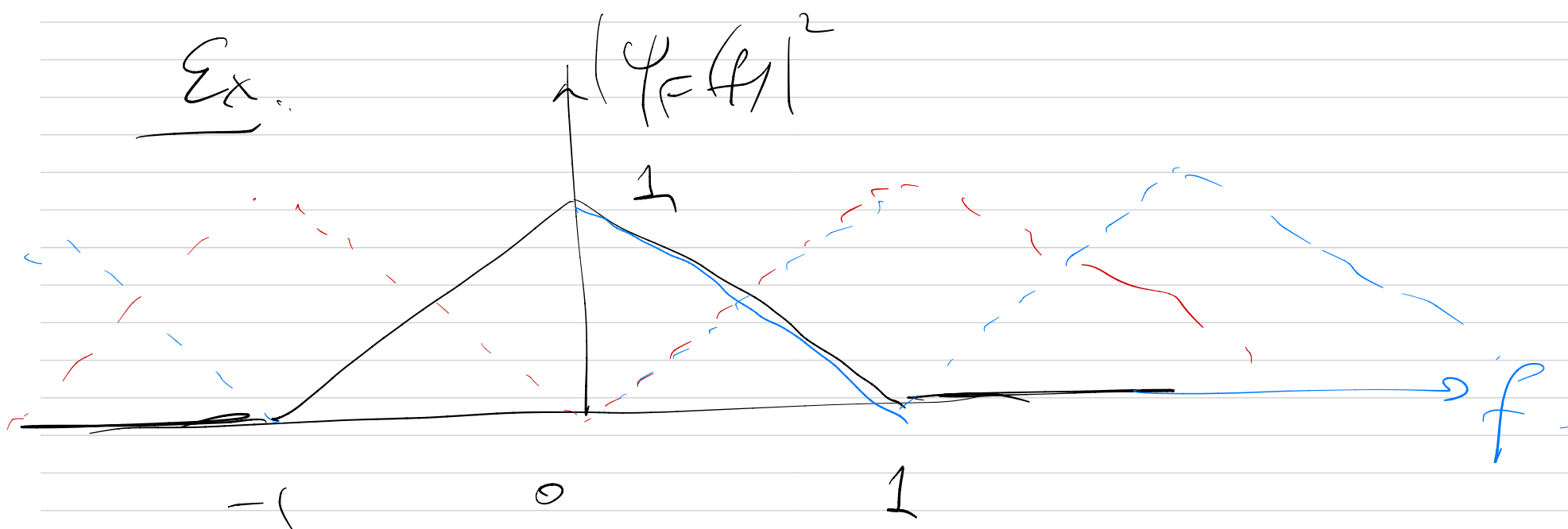


Indeed if we insist that $\psi_F(t) = 0$ when $|t| > 1/2$ then ψ_F needs to be equal to

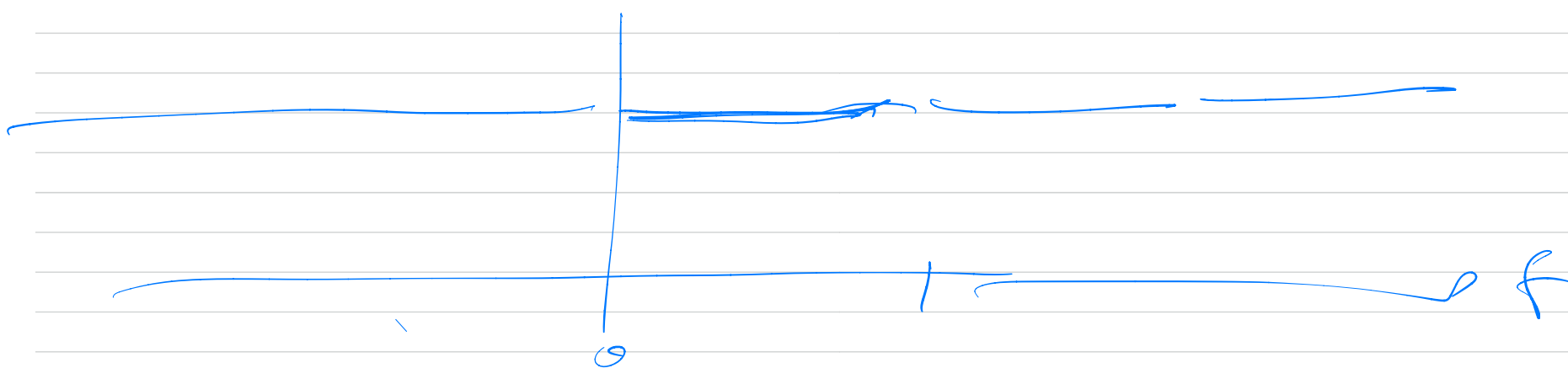


$\Rightarrow$ the example above gives the minimal bandwidth function that satisfies the Nyquist condition.

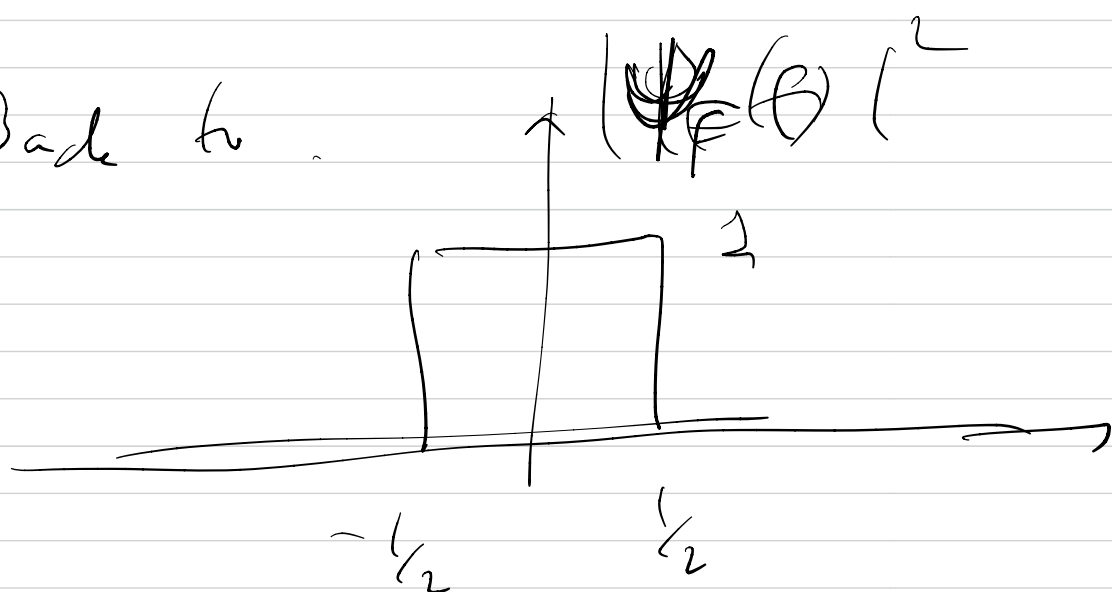
Ex.



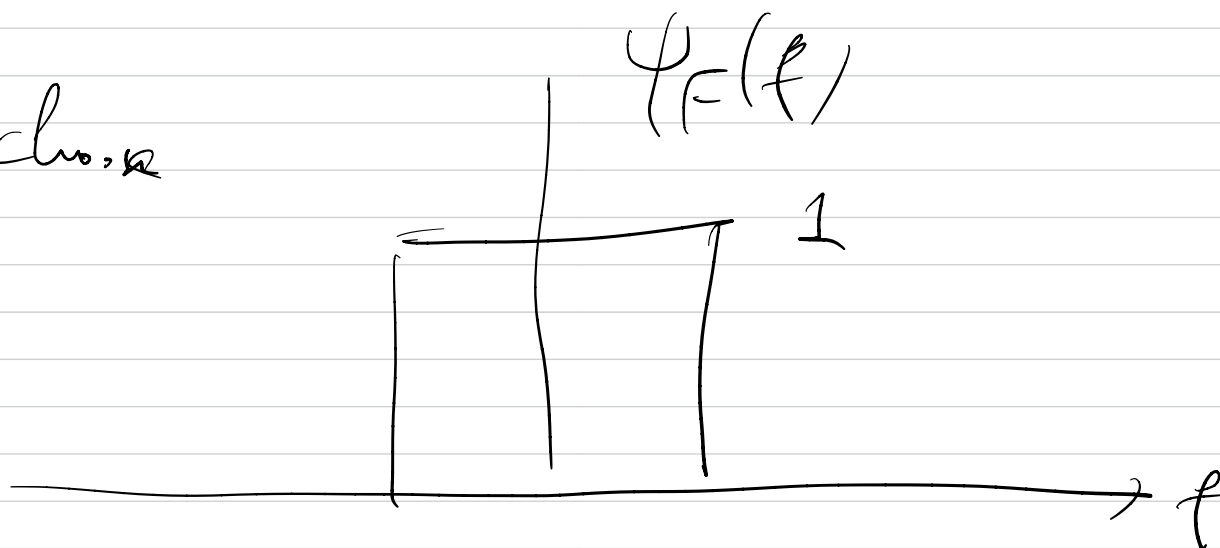
$$\sum_n |\psi_F(f-n)|^2$$



Back to $|\psi_F(f)|^2$

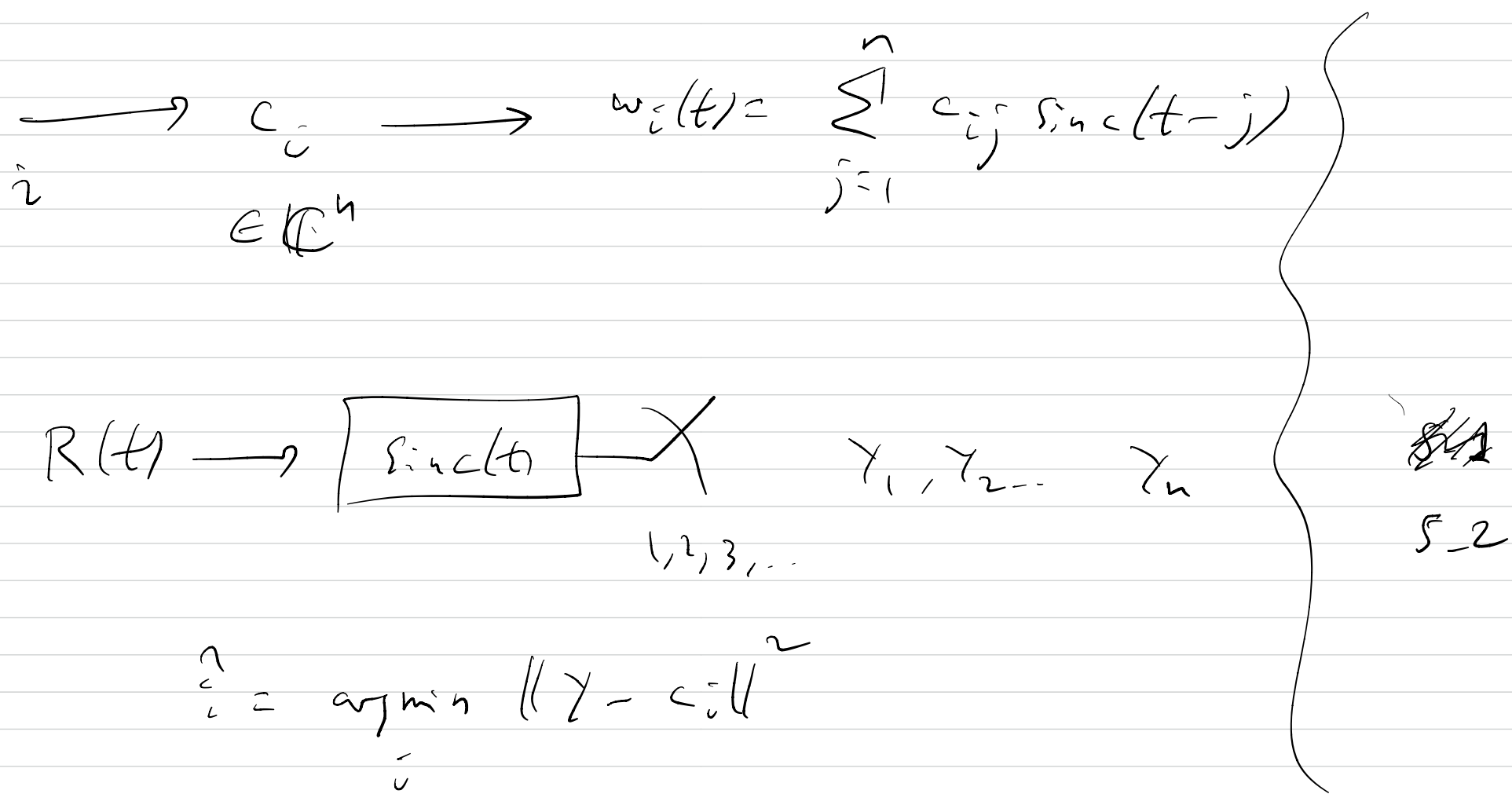


we can choose



$$\equiv \psi(t) = \text{sinc}(t).$$

So, one can build a transmitter & RCDR as



$$\hat{c}_i = \arg \min_i \| \gamma - c_i \|^2$$

5-3 : Power Spectral Density.

We saw last Friday: if $\dots, X_{-2}, X_{-1}, X_0, X_1, \dots$

a stationary discrete time stochastic process with zero-mean

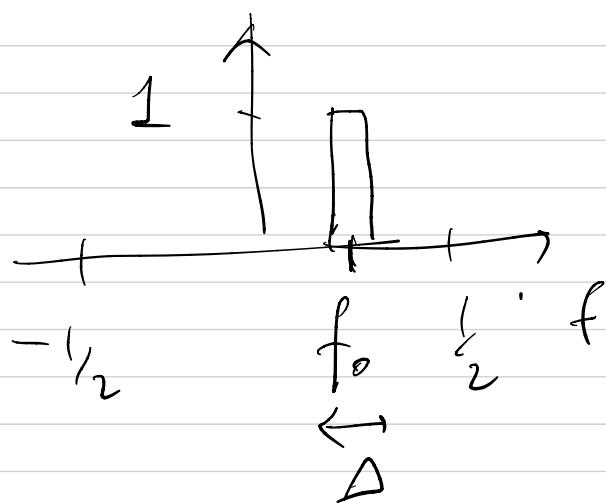
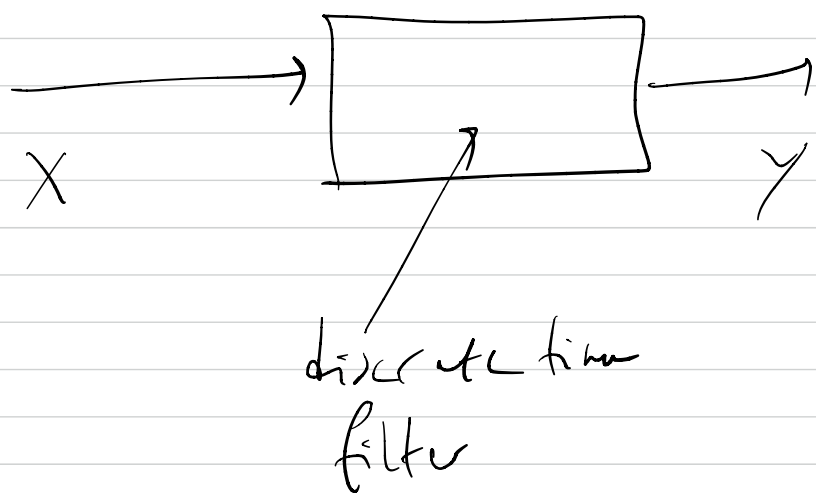
the quantity $S_X(f) \triangleq \sum_k R_X(k) e^{-j2\pi f k}$ ← DFT of R_X

$R_X(k) = E[X_0^* X_k] = E[X_i^* X_{i+k}]$

↑ autocorrelation function

tells us how much power the process has at frequency f .

What I mean by this:



$$E[|Y|^2] = \int_{f_0 - \frac{\Delta}{2}}^{f_0 + \frac{\Delta}{2}} S_X(f) df$$

$$\approx \underline{\underline{S_X(f_0) \Delta}}$$

This gives meaning to the name "Power Spectral Density" of the process X .

The continuous time analog is: $(X(t) : t \in \mathbb{R})$

a continuous time stochastic process. Assume $X(t)$ to be 0-mean. Define

$$R(\tau) = E[X(t)^* X(t+\tau)] \quad \left(\begin{array}{l} \text{assume } X(t) \\ \text{is stationary} \end{array} \right)$$

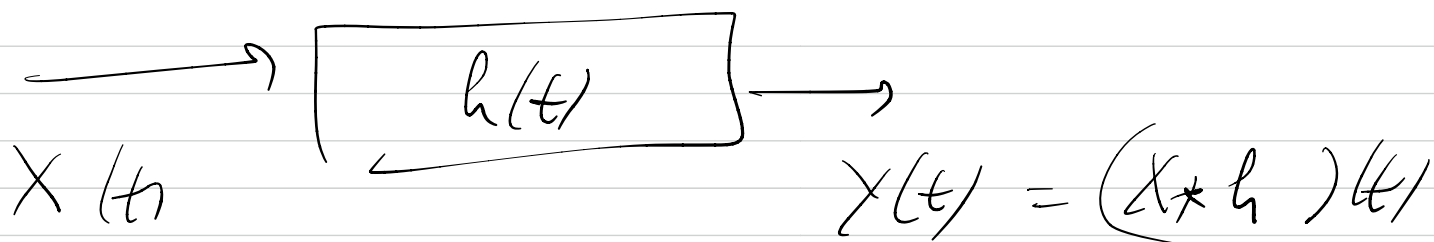
is the auto correlation function.

$(R(0) = E[|X(t)|^2])$ is the power of X

Let us define

$$S_x(f) = \text{F. Transform of } R_x(\tau) \\ = \int R_x(\tau) e^{-j2\pi f\tau} d\tau.$$

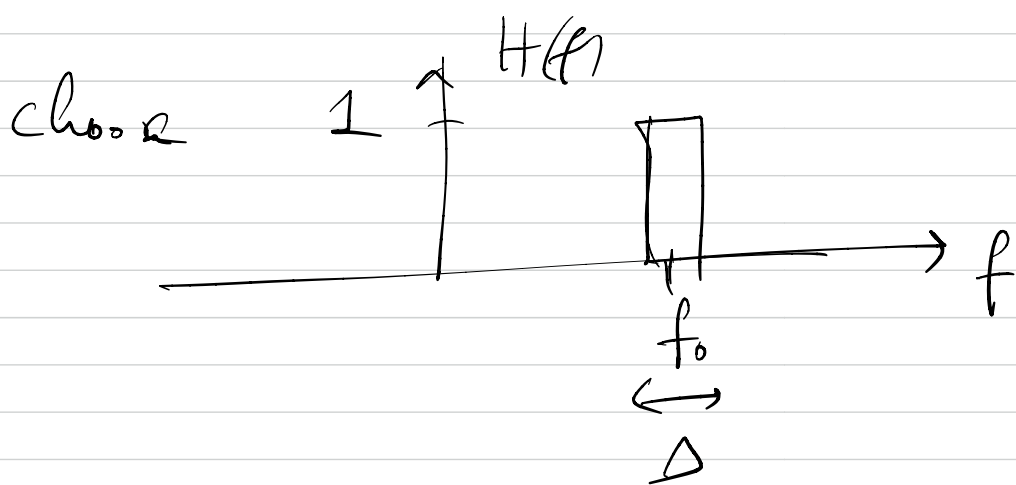
One can show:



$$S_y(f) = S_x(f) |H(f)|^2.$$

$$R_y(0) = \int S_y(f) df = \int S_x(f) |H(f)|^2 df //$$

↑
power in Y



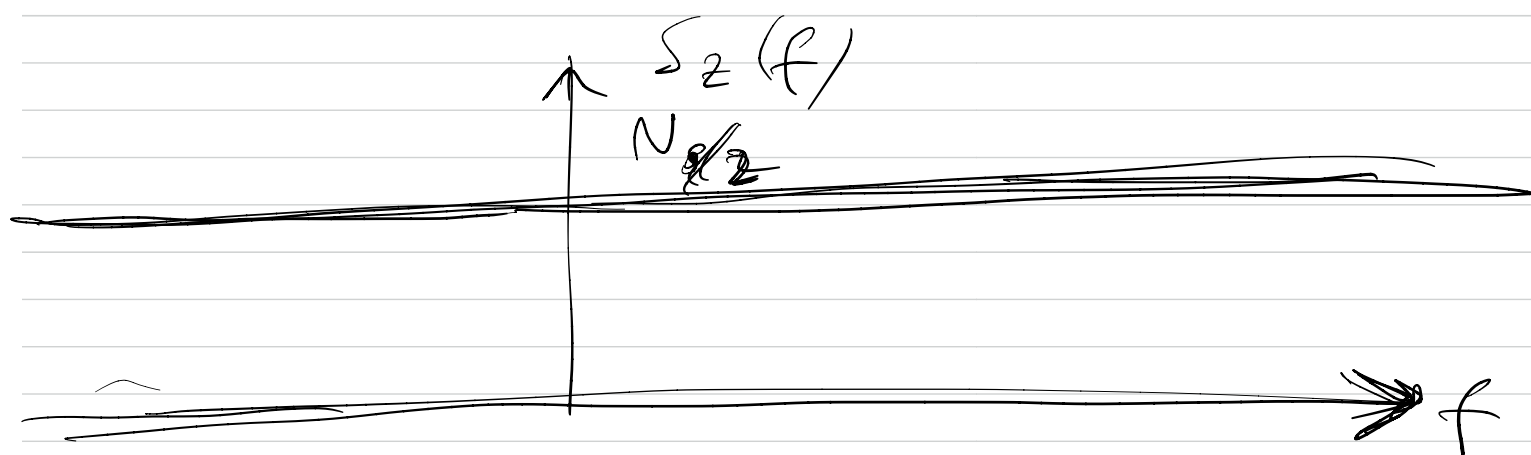
then $Y(t)$ will be the part of X that lives
in the freq. interval $(f_0 - \frac{\Delta}{2}, f_0 + \frac{\Delta}{2})$

and will have power $\approx S_x(f_0) \Delta$

This justifies the name "power spectral density" for $S_x(f)$.

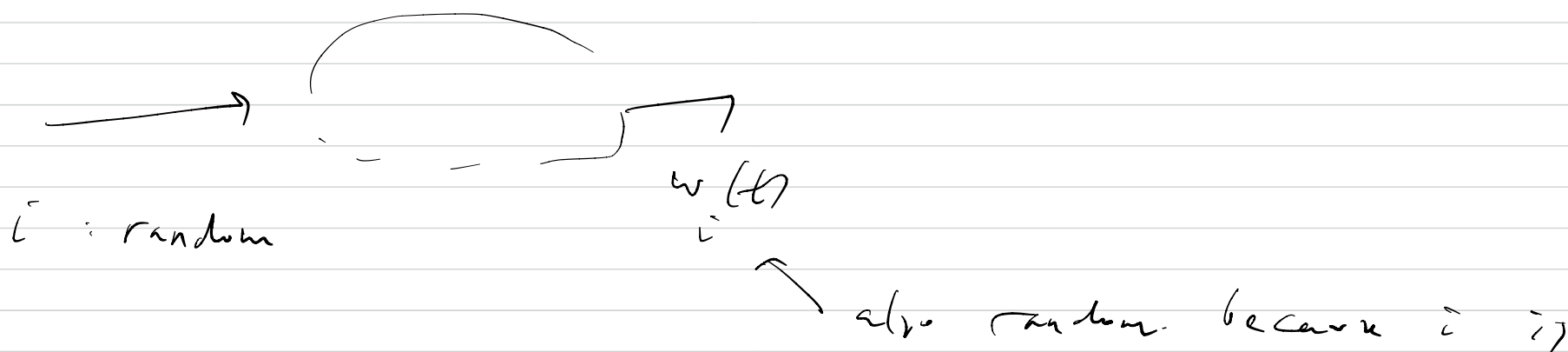
Ex: the "white noise" process $z(t)$ has

$$S_z(f) = \frac{N_0}{2} \text{ for every } f$$



Having told about what is Power-Spect-Dens. &

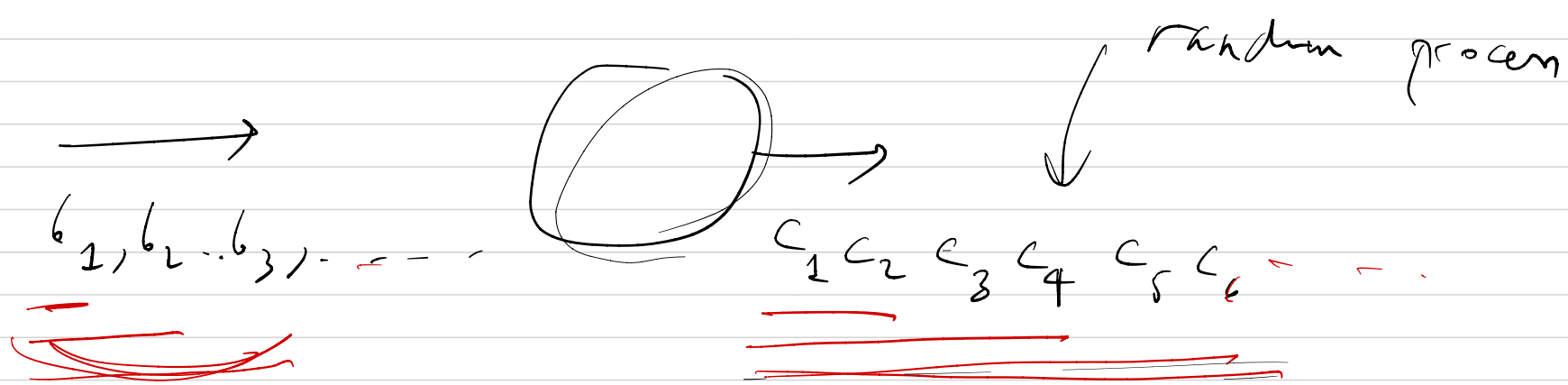
An "Engineering meaning", a real-life use is



$$\bar{i} = k \text{ bit integer} = b_1, \dots, b_k \quad b_i \in \{0, 1\}$$



$$c_{\bar{i}} = \text{vector of length } n = c_{i1}, \dots, c_{in}$$



maintain $n = 2k$

random process, e.g.

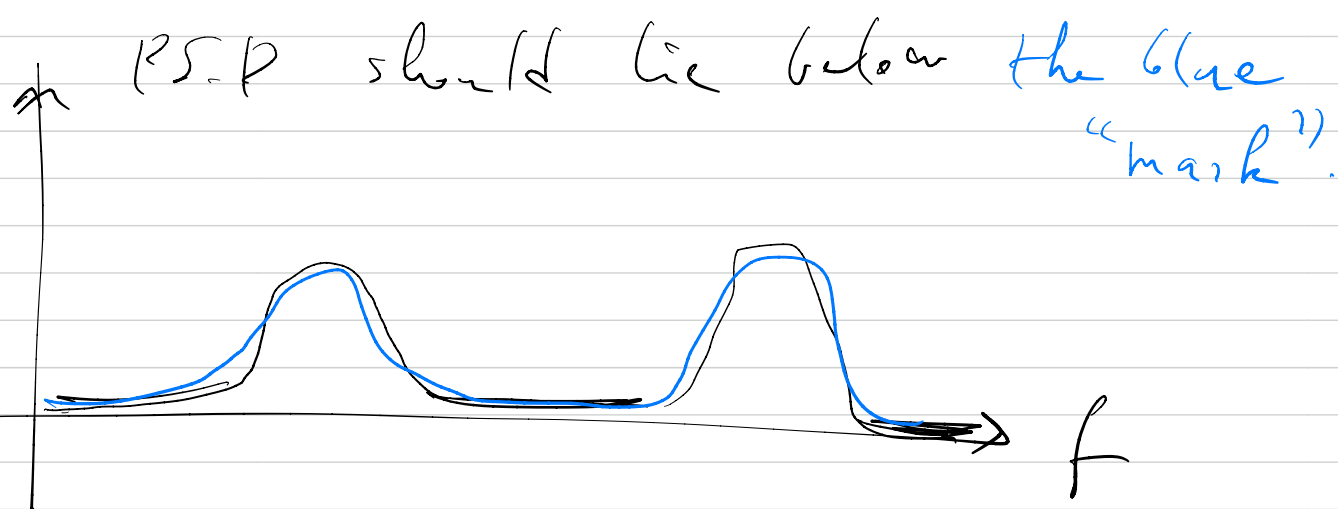
$$b_i \text{ i.i.d. } \Pr(b_i = 0) = \Pr(b_i = 1) = \frac{1}{2}$$

$$w(t) = \sum_{j=-\infty}^{\infty} c_{i,j} \psi(t-jT)$$

So the transmitter

emits a random waveform with this structure

We will be subject to constraints on the P.S.D of $w(t)$:



This suggests: try to compute the P.S.D of

$w(t)$ in terms of $\psi(\cdot)$ and the statistics of

$$\{c_{i,j}\}$$

We see in 5.3 $\int$

$$\underline{X(t)} = \sum_{i=1}^N \underline{X_i} \psi(t-i) \quad \text{then}$$

$$S_x(f) = \underbrace{|\psi_f(f)|^2}_{\text{}} \sum_k e^{-j2\pi k f} K_x(k)$$

$\nearrow$
 $E(X_i^* X_{i+k})$