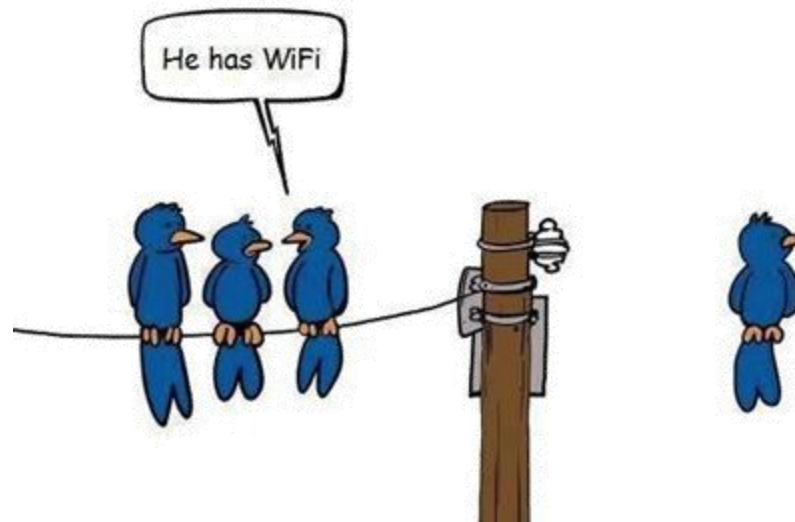


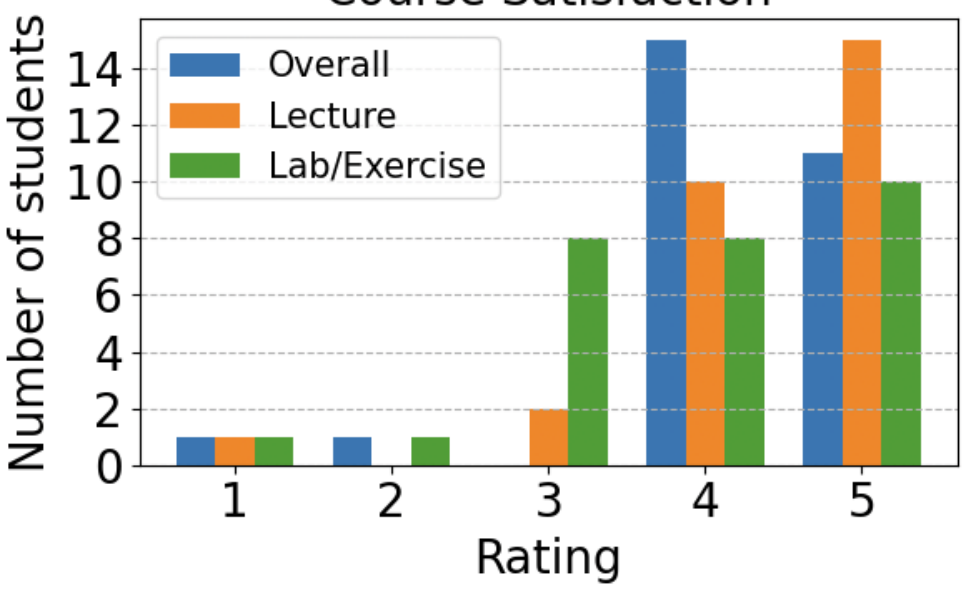
# COM-405: Mobile Networks

## Lecture 8.0: Radar Haitham Hassanieh

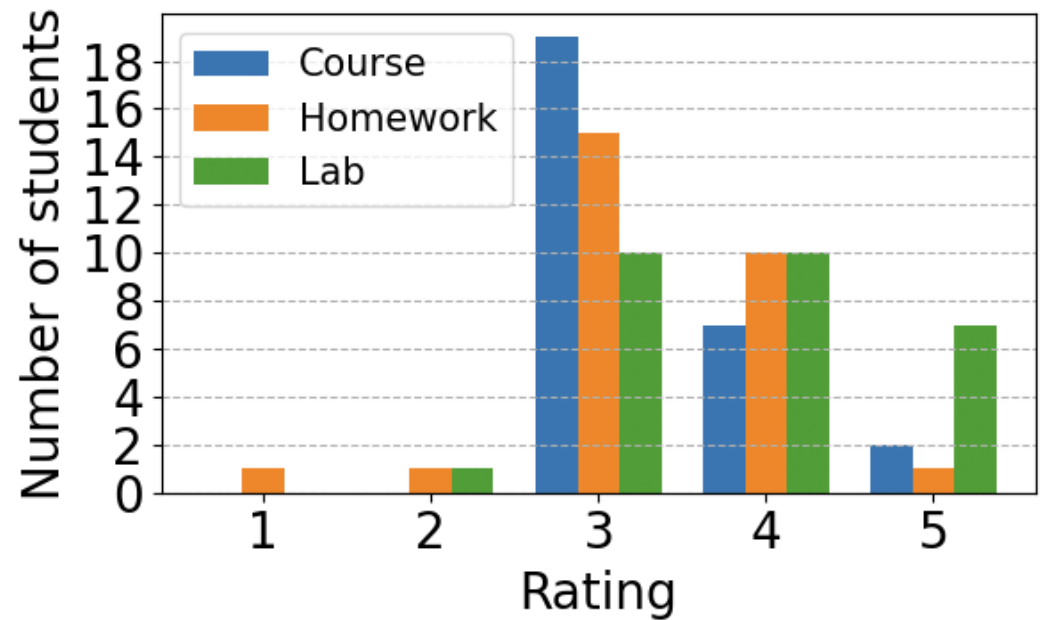


# Feedback Survey Results

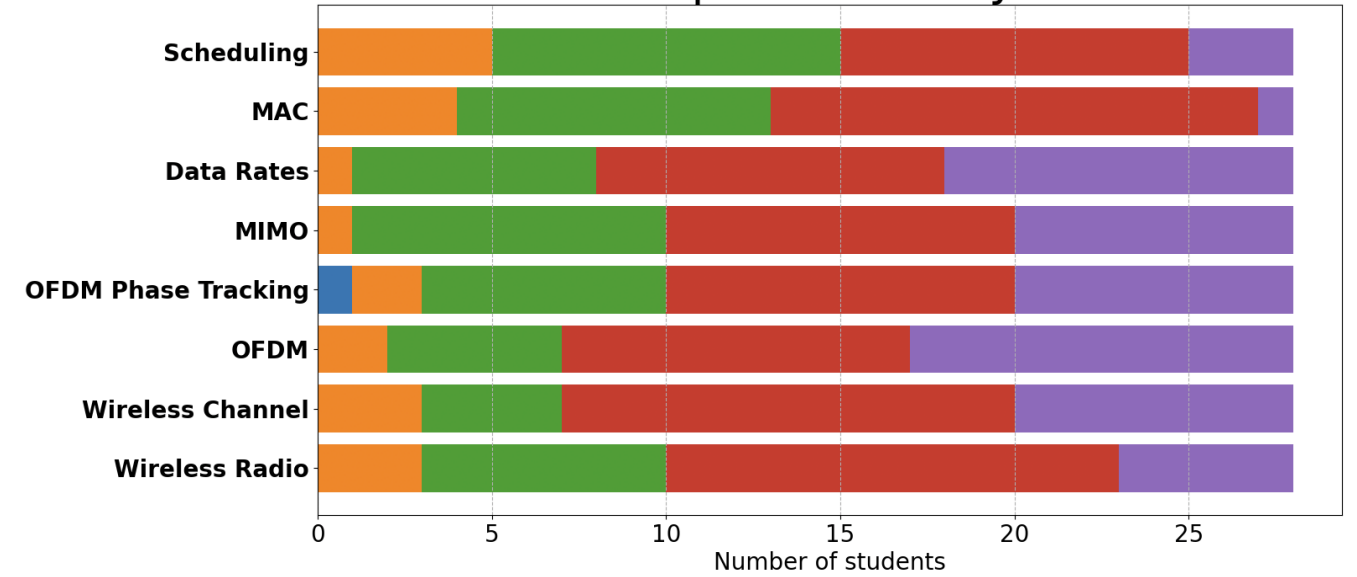
## Course Satisfaction



## Workload

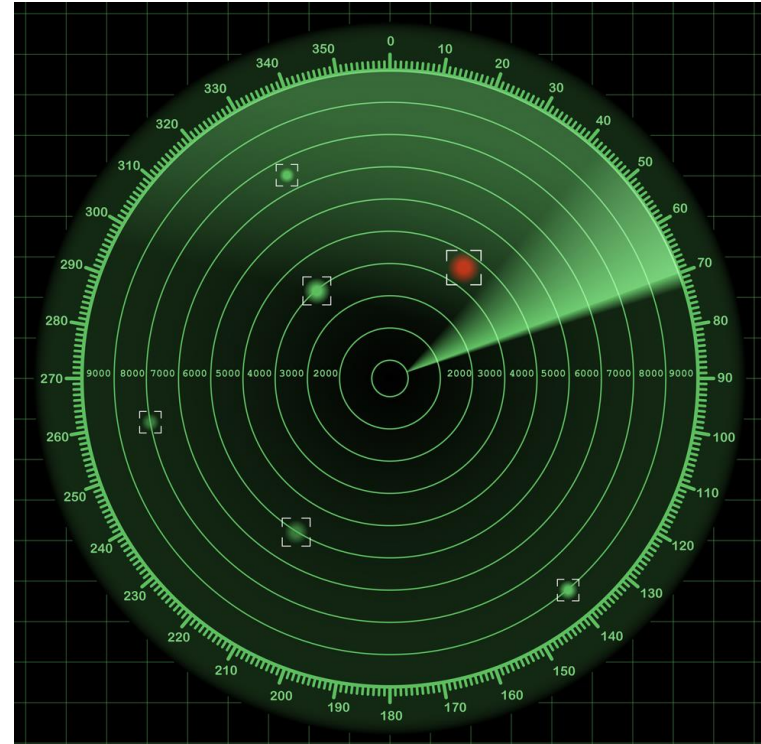


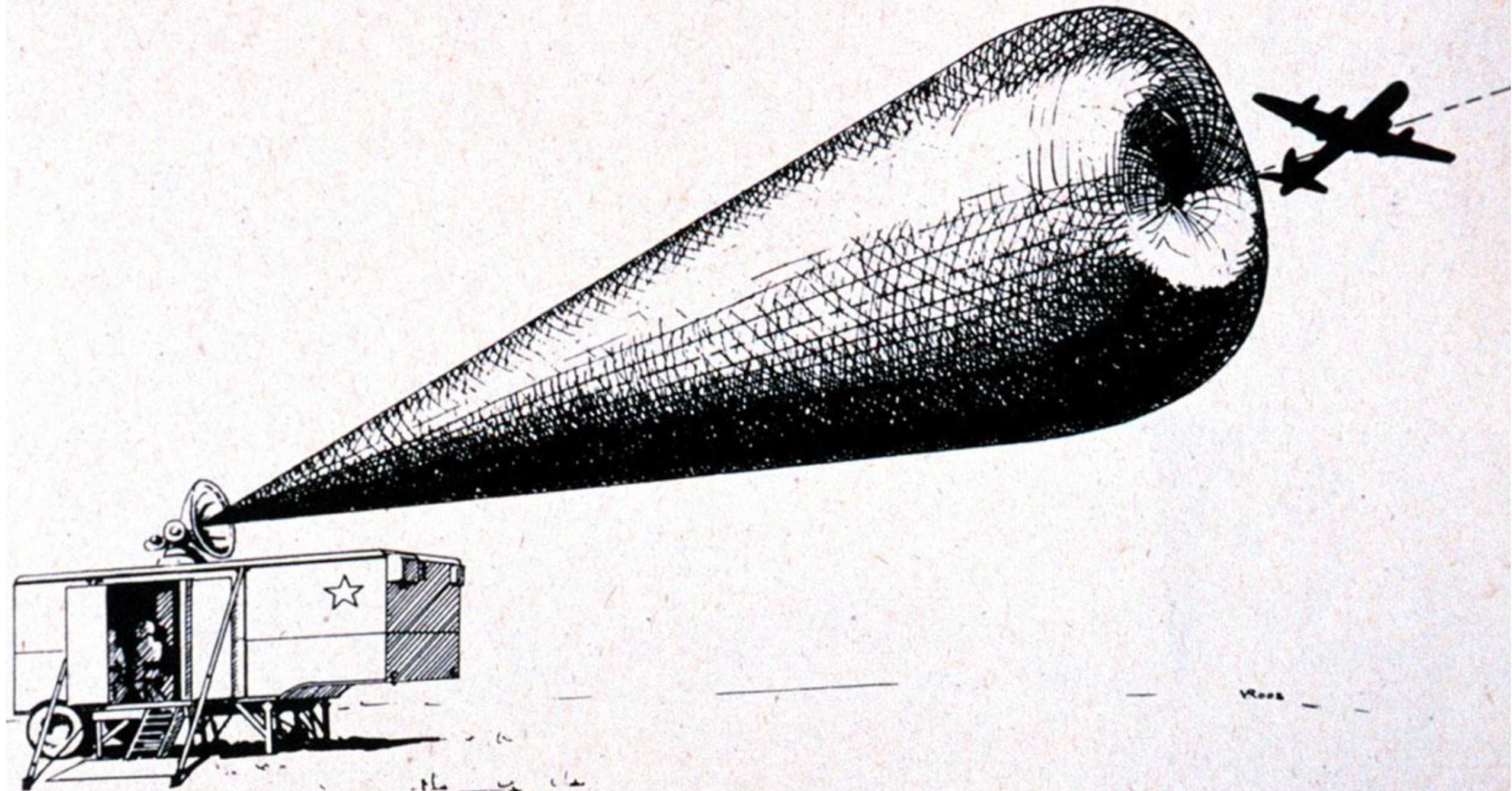
## Topic Familiarity



# RADAR: RAdio Detection And Ranging

- Invented during world war 2 to detect planes
- Changed the course of the war
- Transmit wireless signals, detect the reflections





the ...

1944

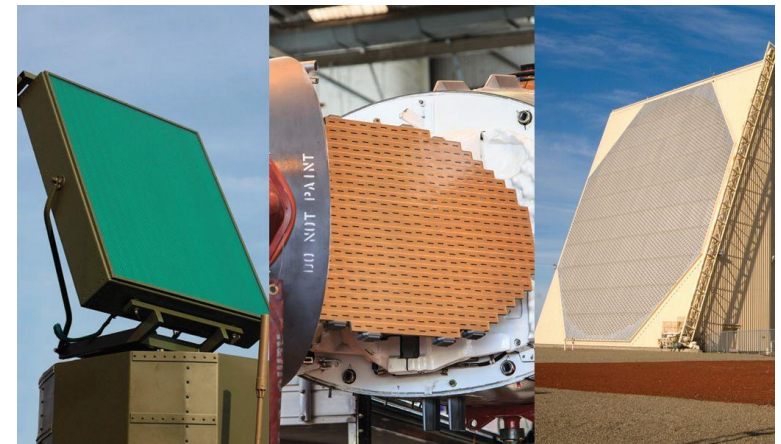
# Radar: Interested in Detecting

- Range: Distance from the radar
  - UWB: Ultra Wide Band
  - FMCW: Frequency Modulated Carrier Wave
- Direction: Angle from which the object is coming

Directional antenna that is rotating.

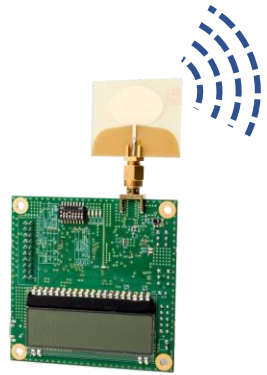


Phased Antenna Array that is always Scanning.

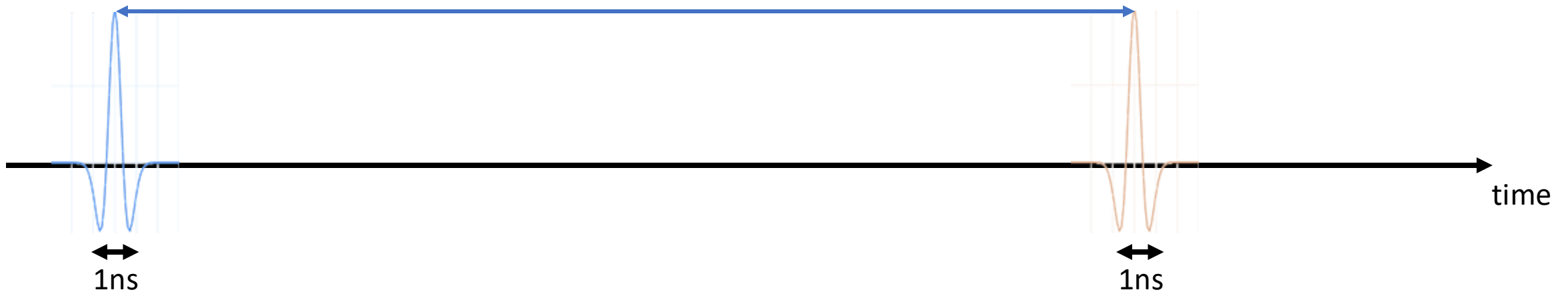


# UWB: Ultra Wide Band

- Transmit a Narrow Pulse in Time (1 ns)

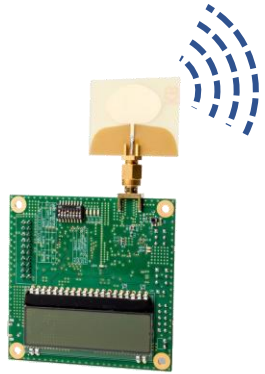


Reflection Time



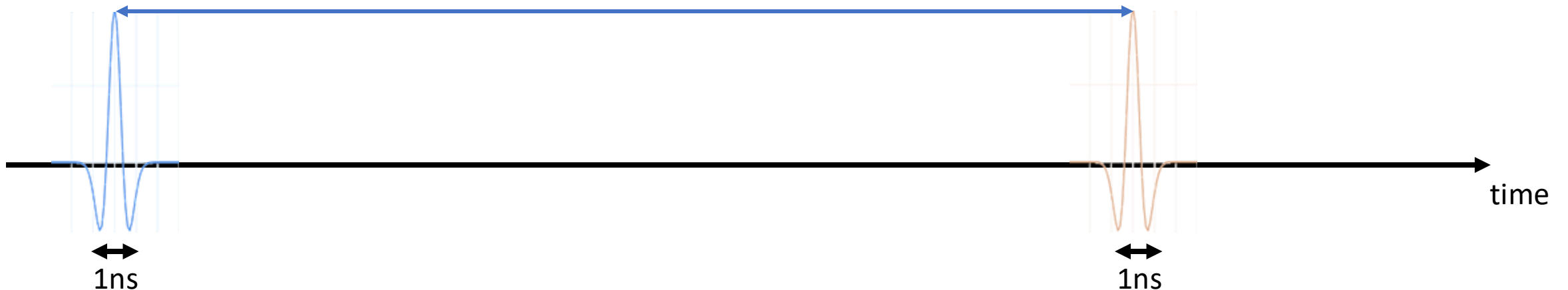
# UWB: Ultra Wide Band

- Transmit a Narrow Pulse in Time (1 ns)



$$\text{Distance} = (\text{Reflection Time} / 2) \times \text{Speed of Light}$$

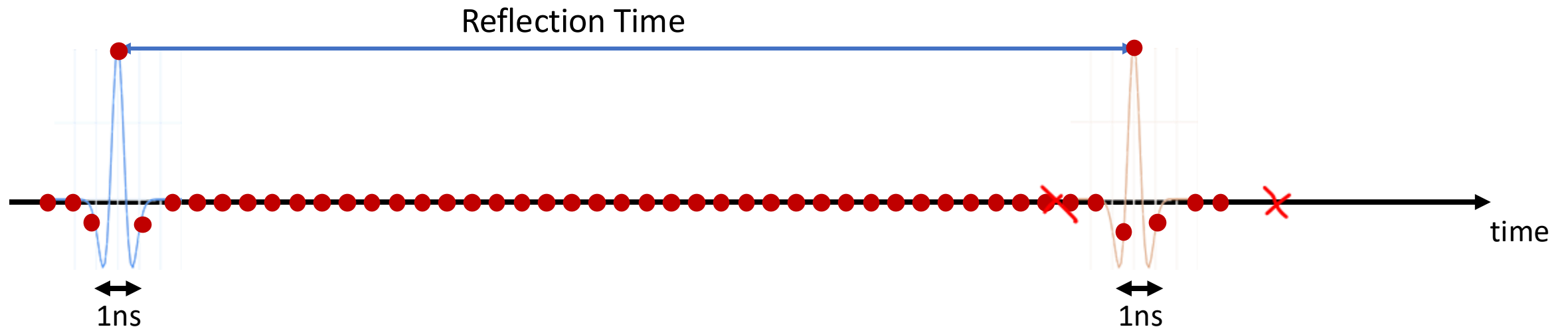
Reflection Time



# UWB: Ultra Wide Band

- Transmit a Narrow Pulse in Time (1 ns)
- Bandwidth =  $1/\text{pulse time} = 1 \text{ GHz} \rightarrow \text{UWB}$
- To detect pulse need to sample at Multi-GigaSample/sec

$$\text{Distance} = (\text{Reflection Time} / 2) \times \text{Speed of Light}$$



# UWB: Ultra Wide Band

## Pros:

- High resolution
- Simple idea
- Used also for communication (UWB 802.15.3a, 3 - 10 GHz)

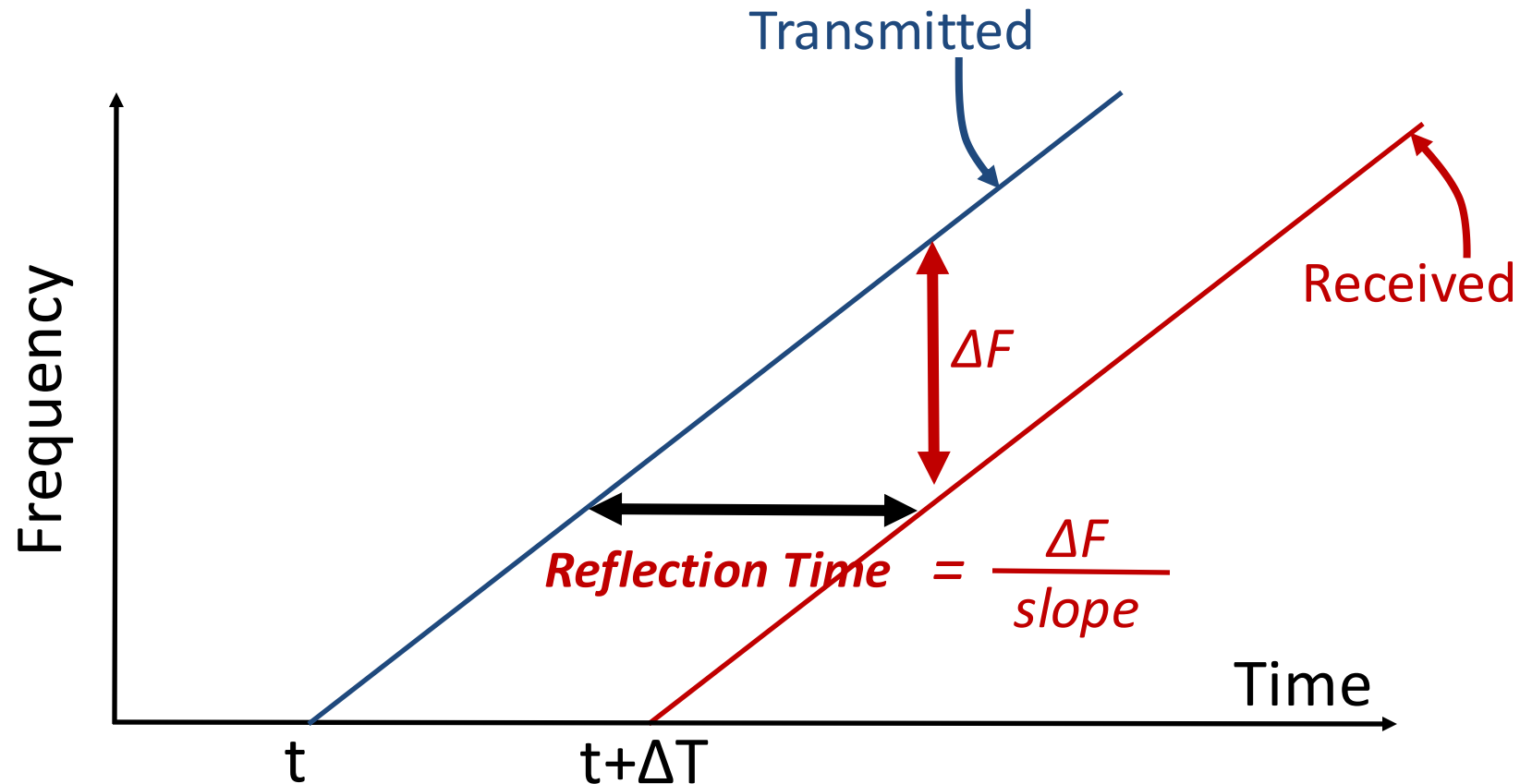
## Cons:

- Very High Speed Sampling Rate!



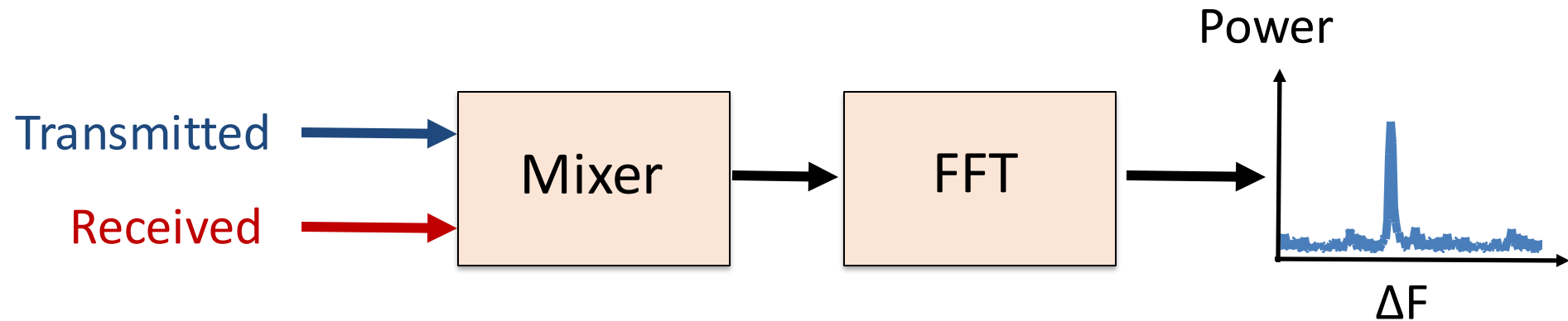
# FMCW: Frequency Modulated Carrier Wave

- Use chirps: signals where the frequency changes with time
- Measure time using a frequency change



# How do we measure $\Delta F$ ?

- Subtracting frequencies is easy (e.g., removing carrier in WiFi)
- Done using a mixer (low-power; cheap)




Signal whose frequency is  $\Delta F$


$\Delta F \rightarrow$  Reflection Time  $\rightarrow$  Distance

# FMCW: Frequency Modulated Carrier Wave



- FMCW Transmitted Signal:  $x(t) = e^{j2\pi\left(\frac{\alpha}{2}t^2 + f_0t + \phi_0\right)}$




- FMCW Received Signal:  $y(t) = Ae^{j2\pi\left(\frac{\alpha}{2}(t-\tau)^2 + f_0(t-\tau) + \phi_0\right)}$   
(Single reflector)

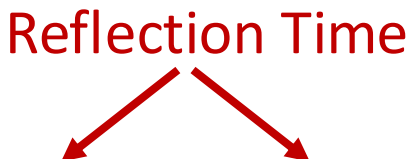
$$= Ae^{j2\pi\left(\frac{\alpha}{2}t^2 - \alpha\tau t + \frac{\alpha}{2}\tau^2 + f_0t - f_0\tau + \phi_0\right)}$$




- Mix TX and RX Signal:  $y_b(t) = y(t) \times x^*(t) = Ae^{-j2\pi(\alpha\tau t + f_0\tau)}$



# FMCW: Frequency Modulated Carrier Wave

- FMCW Transmitted Signal:  $x(t) = e^{j2\pi\left(\frac{\alpha}{2}t^2 + f_0t + \phi_0\right)}$   


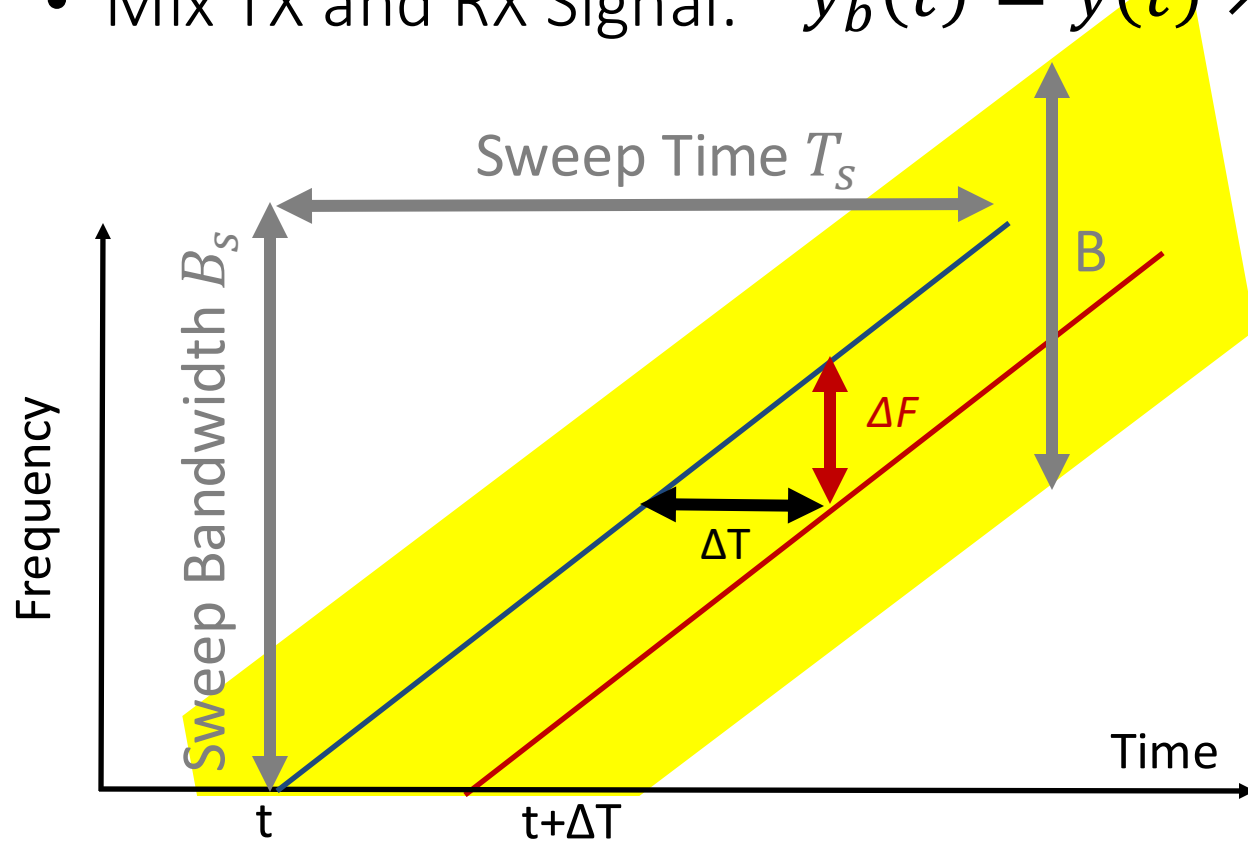
- FMCW Received Signal:  
(Multiple reflectors)  $y(t) = \sum_i A_i e^{j2\pi\left(\frac{\alpha}{2}(t-\tau_i)^2 + f_0(t-\tau_i) + \phi_0\right)}$   


- Mix TX and RX Signal:  $y_b(t) = y(t) \times x^*(t) = \sum_i A_i e^{-j2\pi(\alpha\tau_i t + f_0\tau_i)}$   
  


# FMCW: Frequency Modulated Carrier Wave

- FMCW Transmitted Signal:  $x(t) = e^{j2\pi\left(\frac{\alpha}{2}t^2 + f_0t + \phi_0\right)}$

- Mix TX and RX Signal:  $y_b(t) = y(t) \times x^*(t) = \sum_i A_i e^{-j2\pi(\alpha\tau_i t + f_0\tau_i)}$



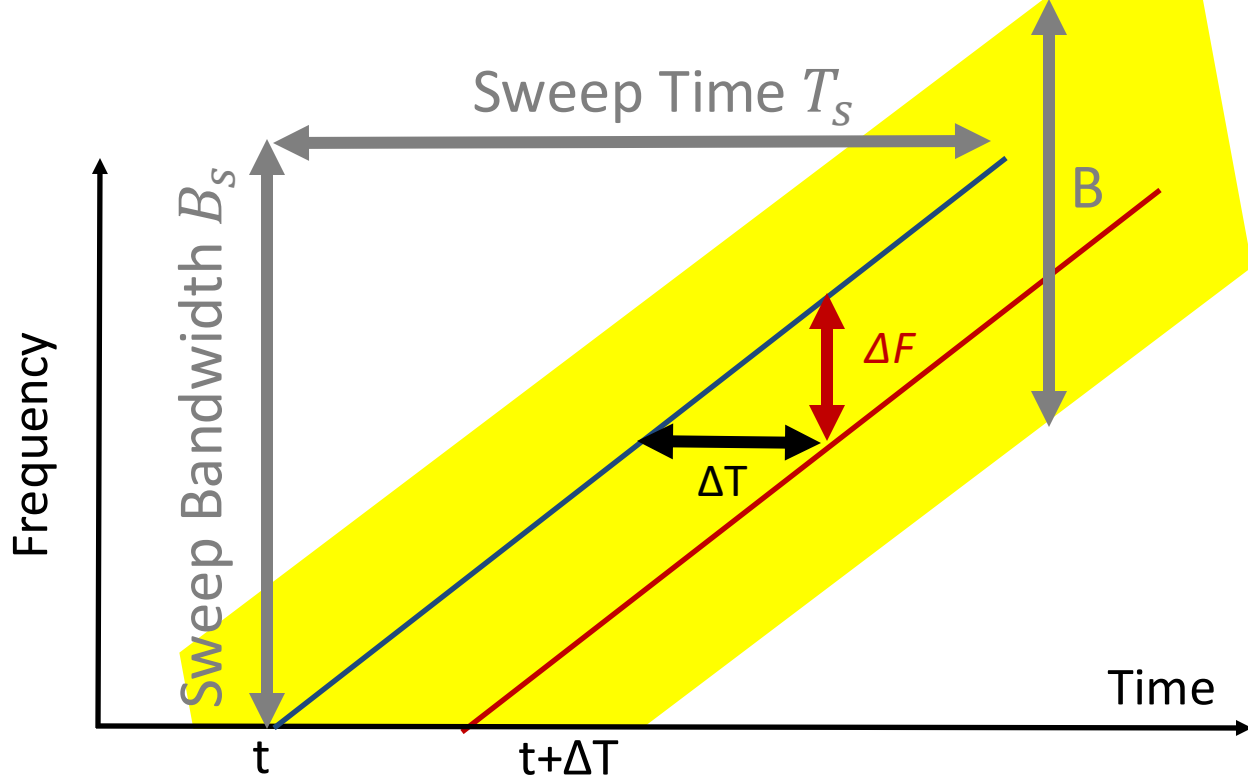
$$\text{Slope: } \alpha = \frac{B_s}{T_s}$$

$$\Delta F = \alpha \Delta T = \alpha \tau_i$$

$$\text{Sampling Rate: } B \ll B_s$$

$$\text{FFT Window: } T = T_s$$

# FMCW: Frequency Modulated Carrier Wave



Slope:  $\alpha = \frac{B_s}{T_s}$

$\Delta F = \alpha \Delta T = \alpha \tau_i$

Sampling Rate:  $B \ll B_s$

FFT Window:  $T = T_s$

**Max Range:**  $\Delta F < B \Rightarrow \tau_{max} < B/\alpha = B T_s/B_s \Rightarrow d_{max} < c B T_s/2B_s$

**Resolution:**  $\delta F = 1/T = 1/T_s \Rightarrow \tau_{min} = 1/\alpha T_s = 1/B_s \Rightarrow d_{min} = c/2B_s$

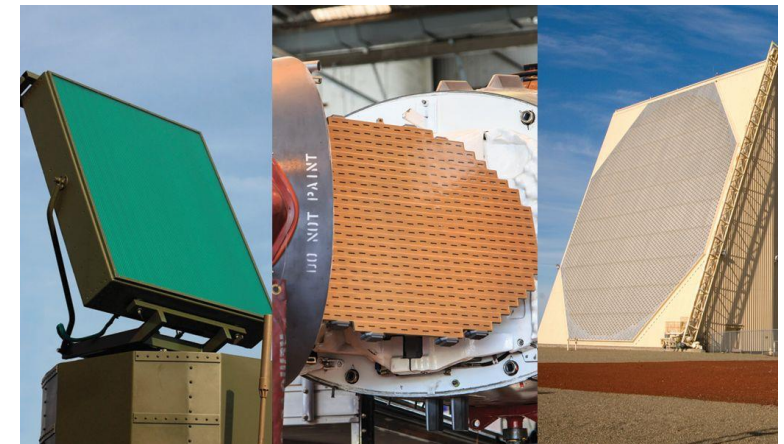
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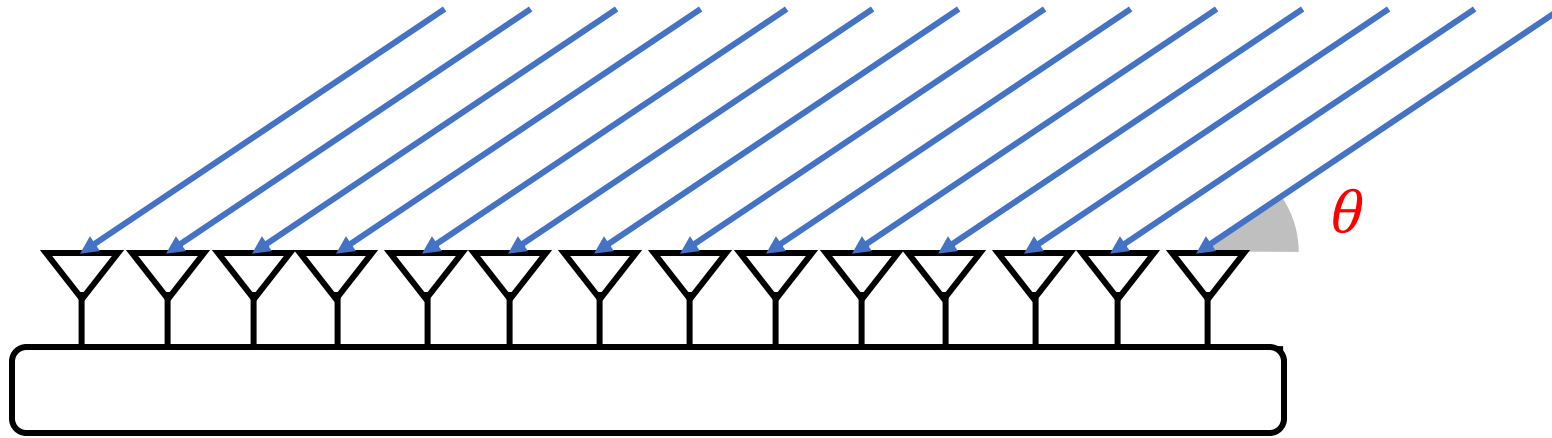
Directional antenna that is rotating.



Phased Antenna Array that is always Scanning.



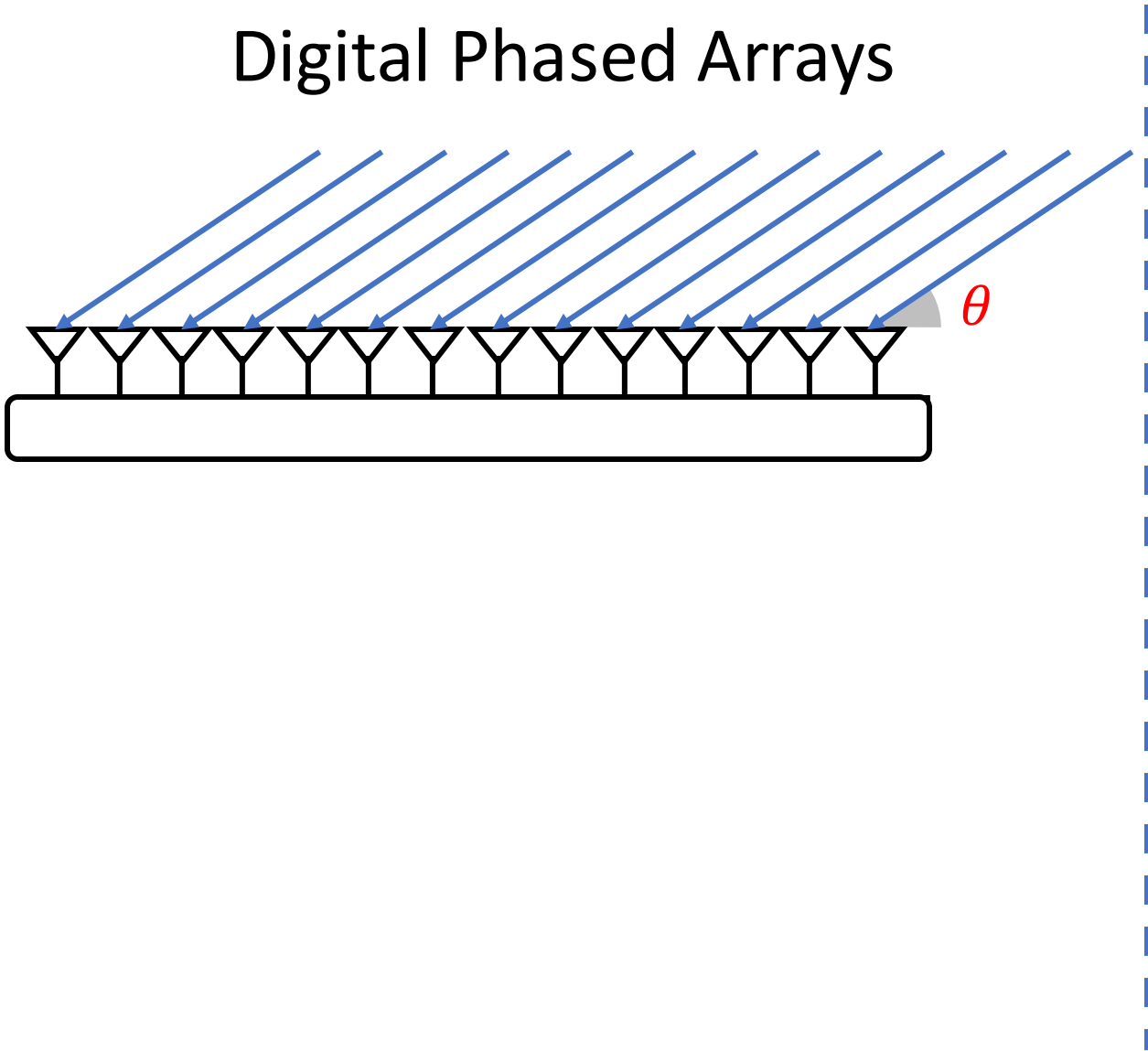
# Phased Arrays Primer



$$h_k = \alpha_1 e^{-j2\pi \frac{d_1 - k s \cos \theta_1}{\lambda}}$$

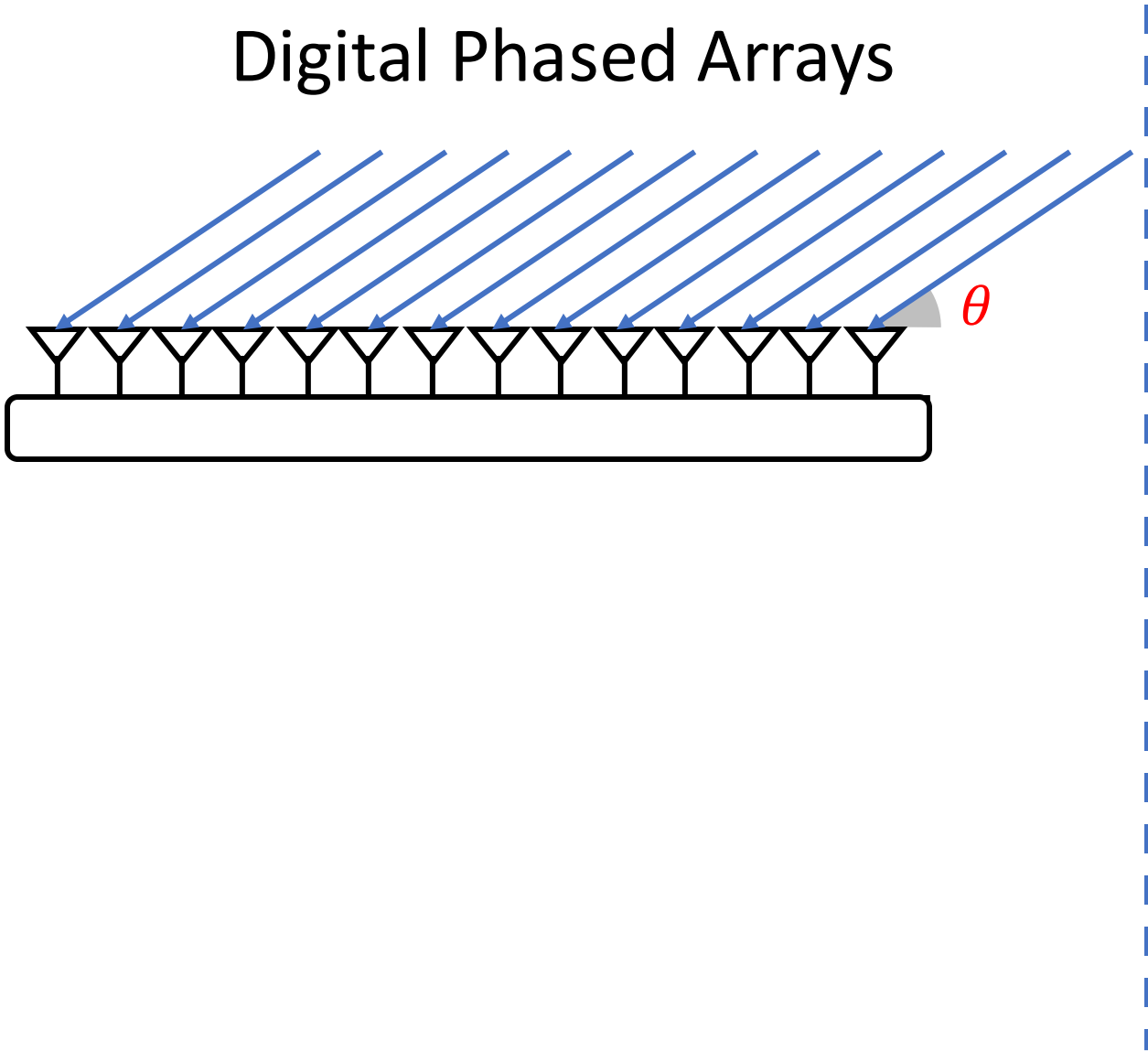
# Phased Arrays Primer

## Digital Phased Arrays



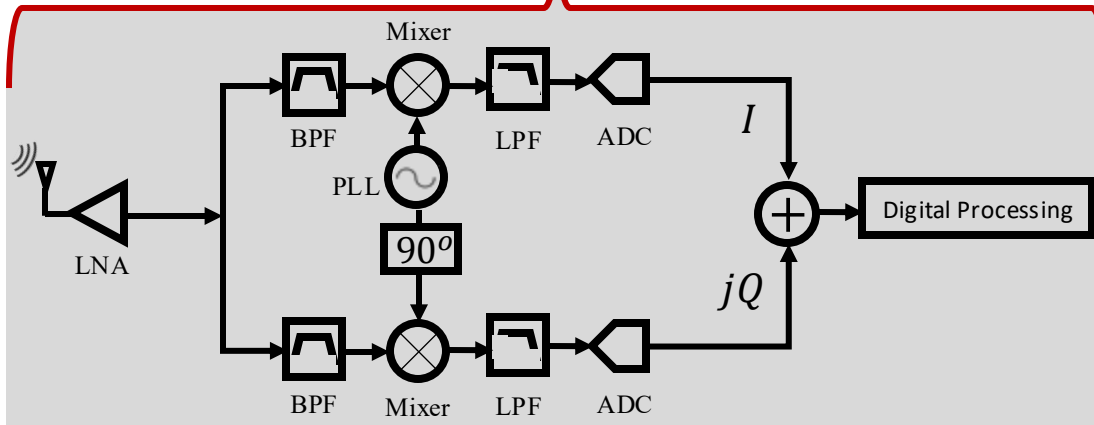
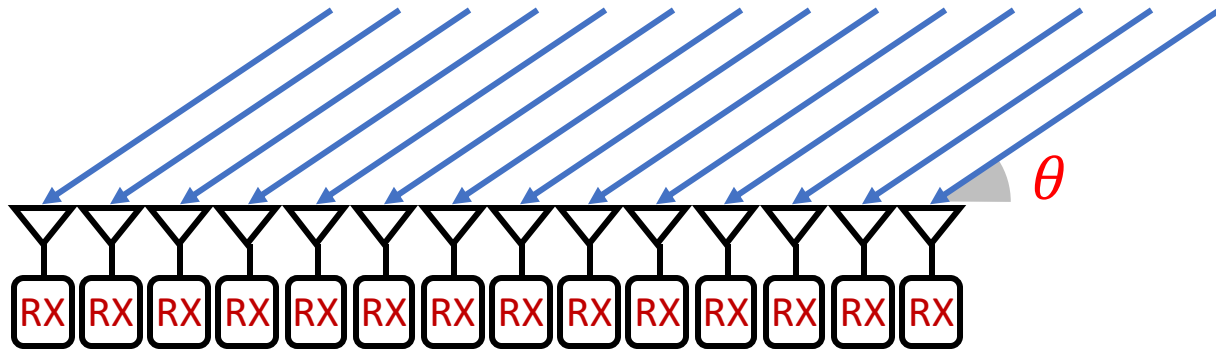
# Phased Arrays Primer

## Digital Phased Arrays



# Phased Arrays Primer

## Digital Phased Arrays

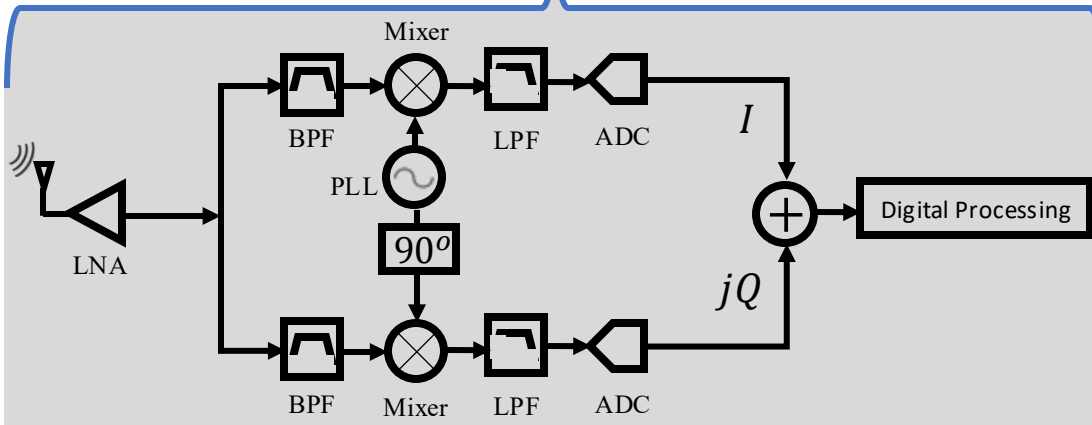
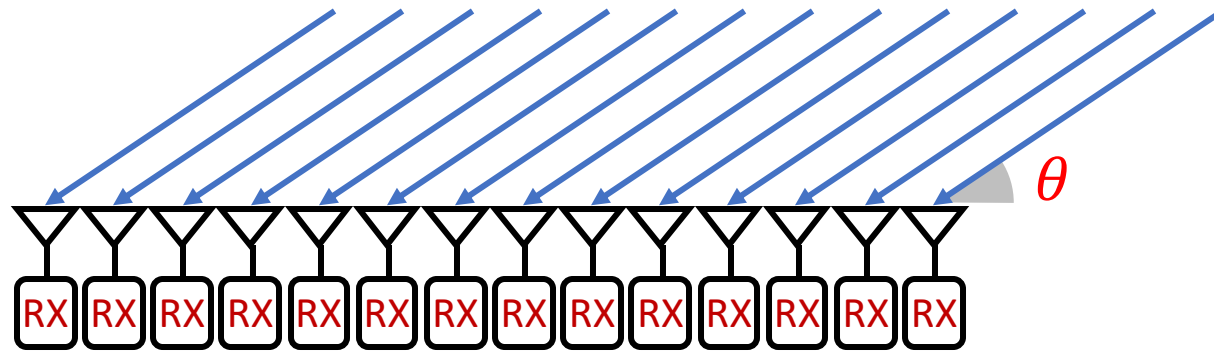


**Every antenna connected to full fledged RX**

**Sample & Process Signals in Digital**

# Phased Arrays Primer

## Digital Phased Arrays

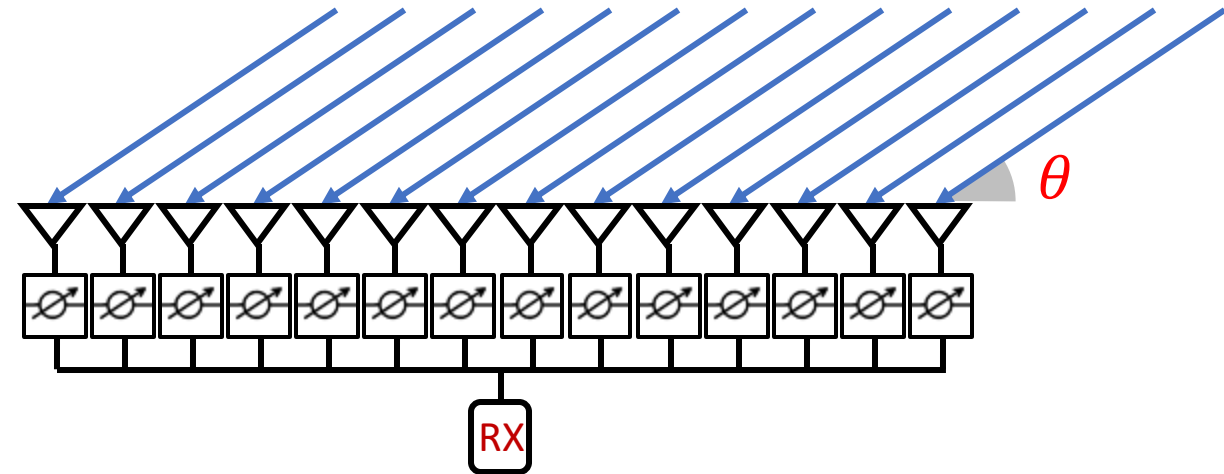


**Every antenna connected to full fledged RX**

**Sample & Process Signals in Digital**

**Very expensive and high power for large arrays & mmWave**

## Analog Phased Arrays

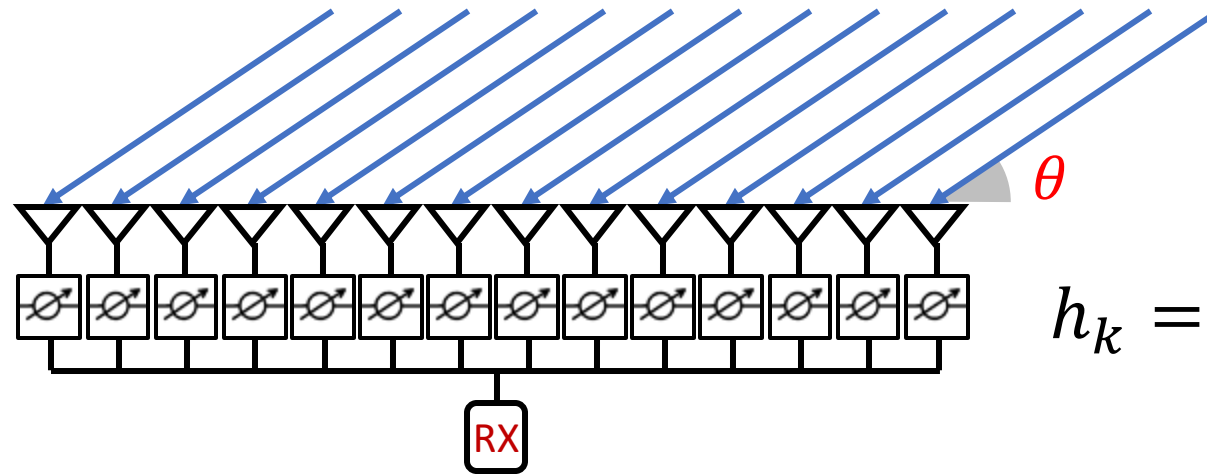


**All antennas connected to a single receiver**

- Each antenna connected to a phase shifter.
- Phase shifter changes the phase of the signal on each antenna by multiplying with  $e^{j\phi}$ .
- Steer the beam electronically by changing the phases of the signals.
- Get the sum along a certain direction.

**Cheap & low power but requires scanning**

# Analog Phased Arrays

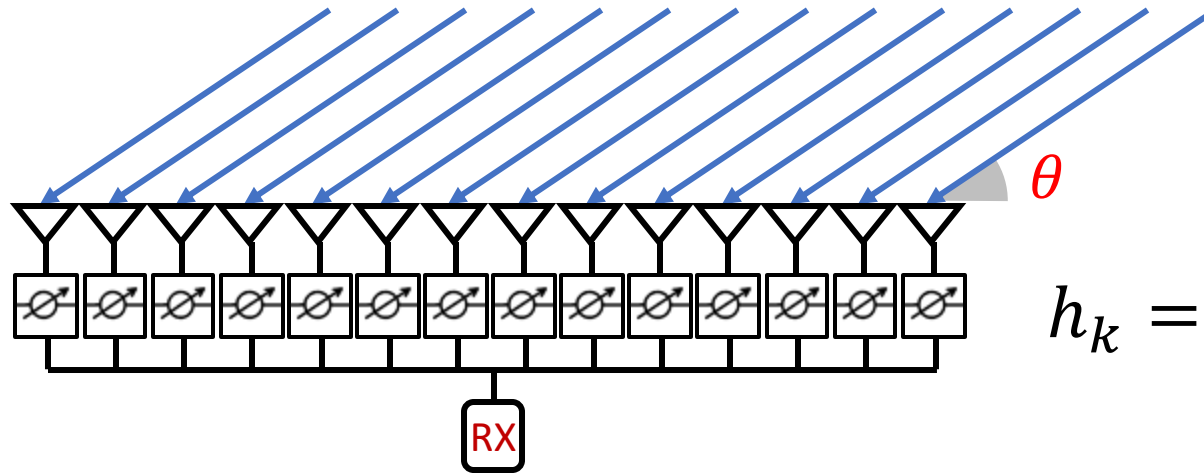


$$h_k = \alpha e^{-j2\pi \frac{d-k s \cos \theta}{\lambda}}$$

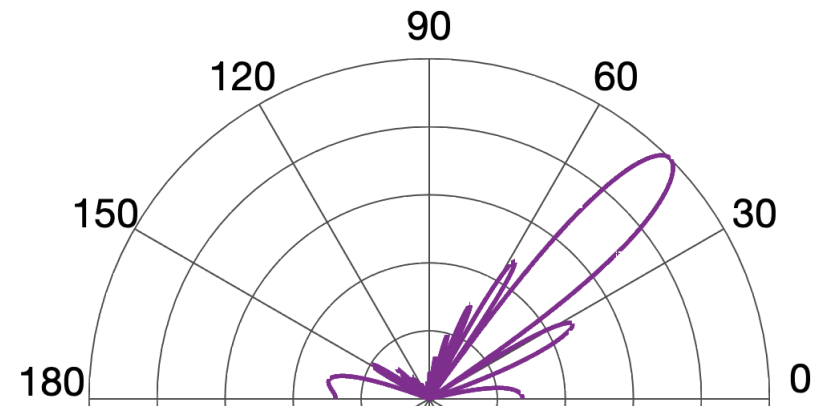
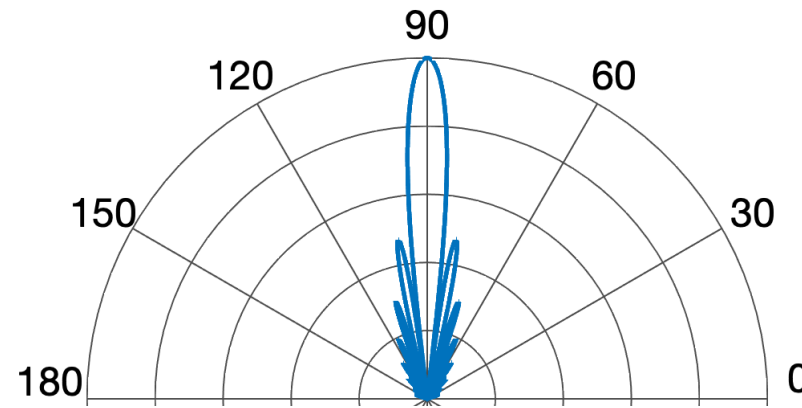
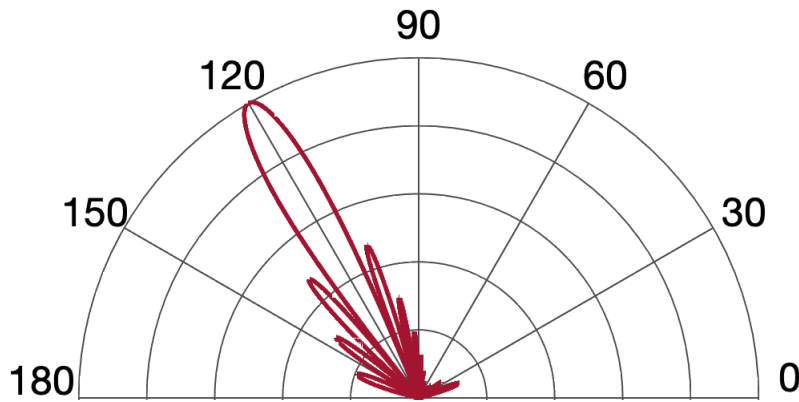
$$\begin{aligned}
 y(t) &= \sum_1^N y_k(t) e^{j\phi_k} = \sum_1^N h_k x(t) e^{j\phi_k} = \sum_1^N \alpha e^{-j2\pi \frac{d-k s \cos \theta}{\lambda}} x(t) e^{j\phi_k} \\
 &= x(t) \alpha e^{-j2\pi \frac{d}{\lambda}} \sum_1^N e^{j\pi k \cos \theta} e^{j\phi_k}
 \end{aligned}$$

To get signal along direction  $\theta_1$ , set the phases on the phase shifters to  $\phi_k = -\pi k \cos \theta$

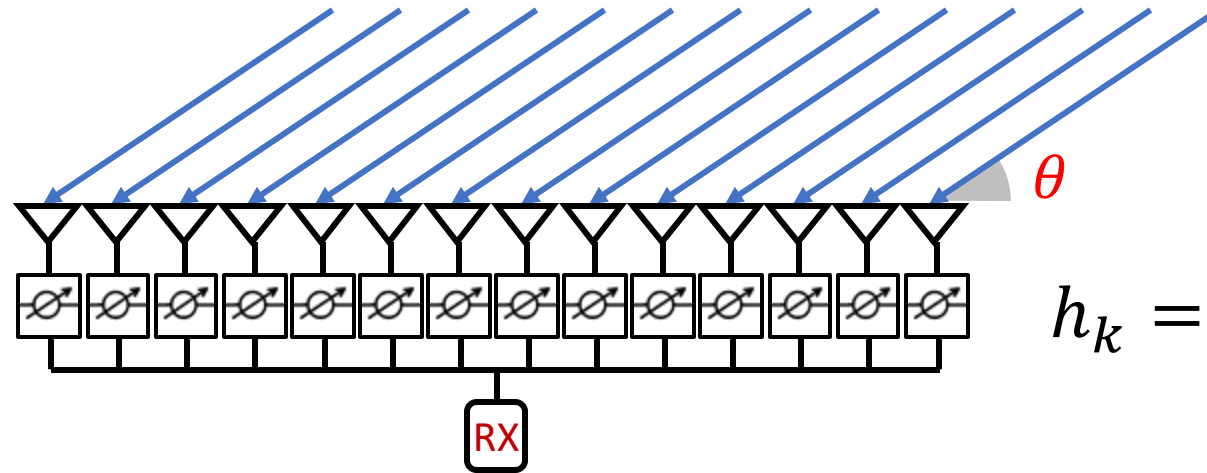
# Analog Phased Arrays



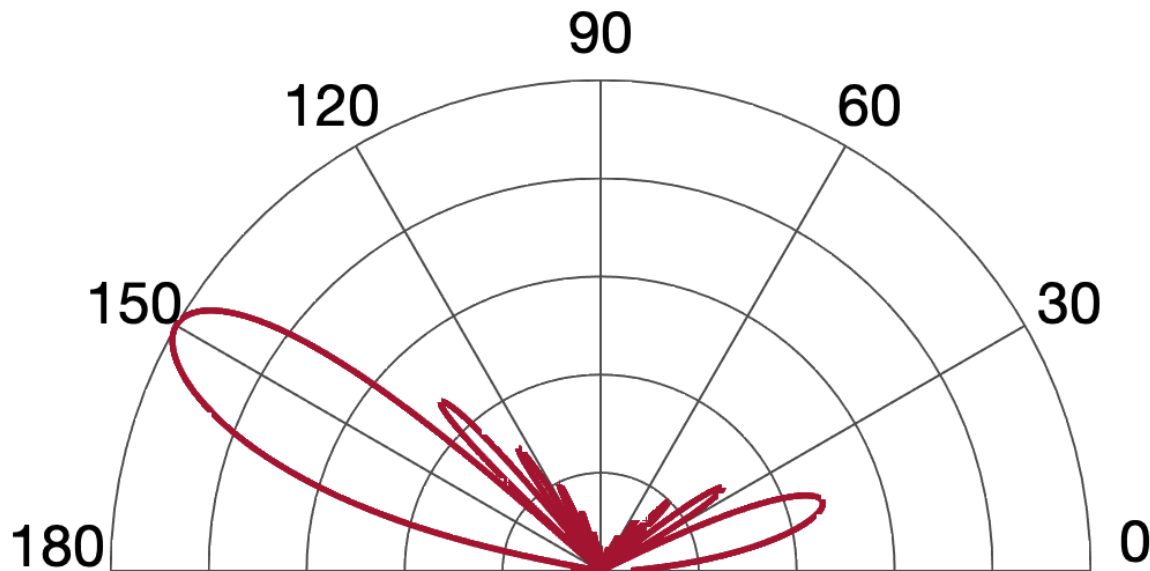
$$h_k = \alpha e^{-j2\pi \frac{d - k s \cos \theta}{\lambda}}$$



# Analog Phased Arrays

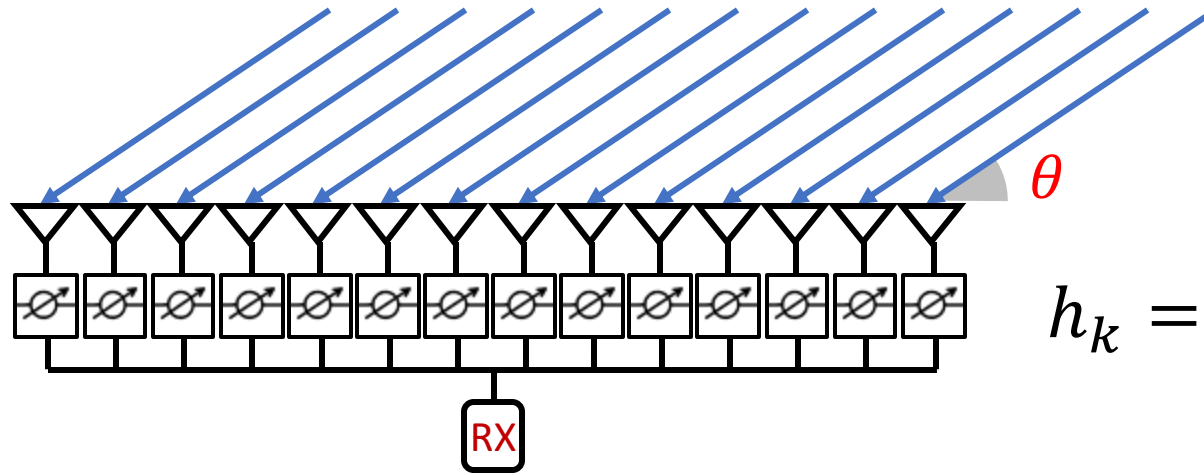


$$h_k = \alpha e^{-j2\pi \frac{d - k s \cos \theta}{\lambda}}$$

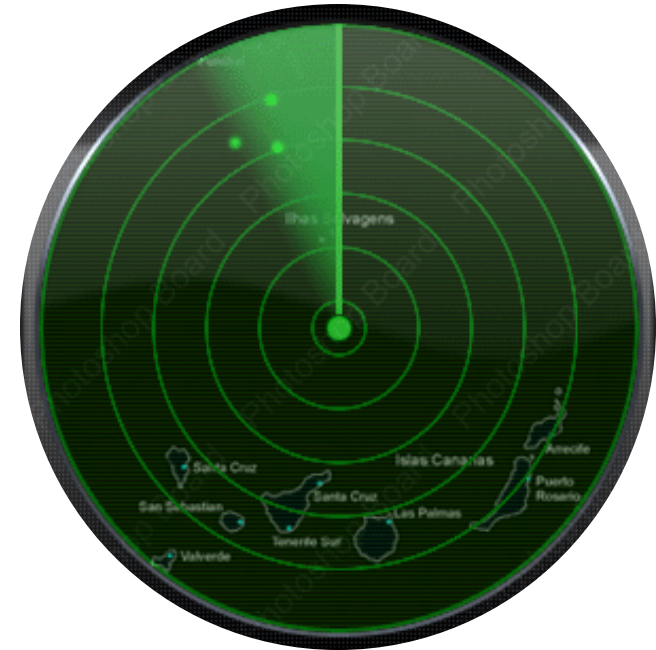
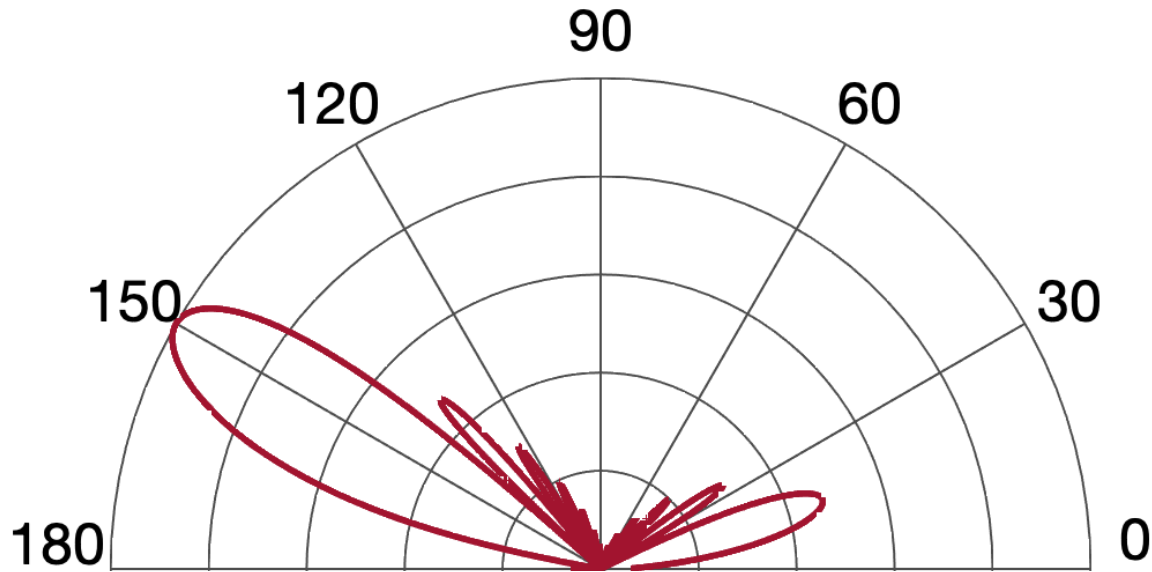


- (1) Set the phase shifters to receive signals from a given direction.
- (2) Send FMCW signals.
- (3) Receiver FMCW reflections, down convert, sample and compute range FFT.
- (4) Repeat until you get 2D range image: AoA + Range

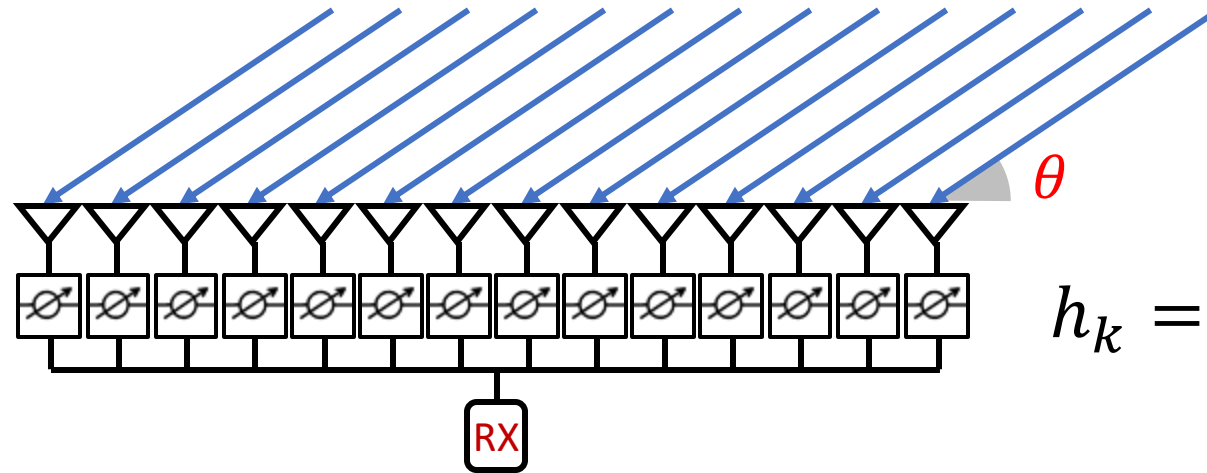
# Analog Phased Arrays



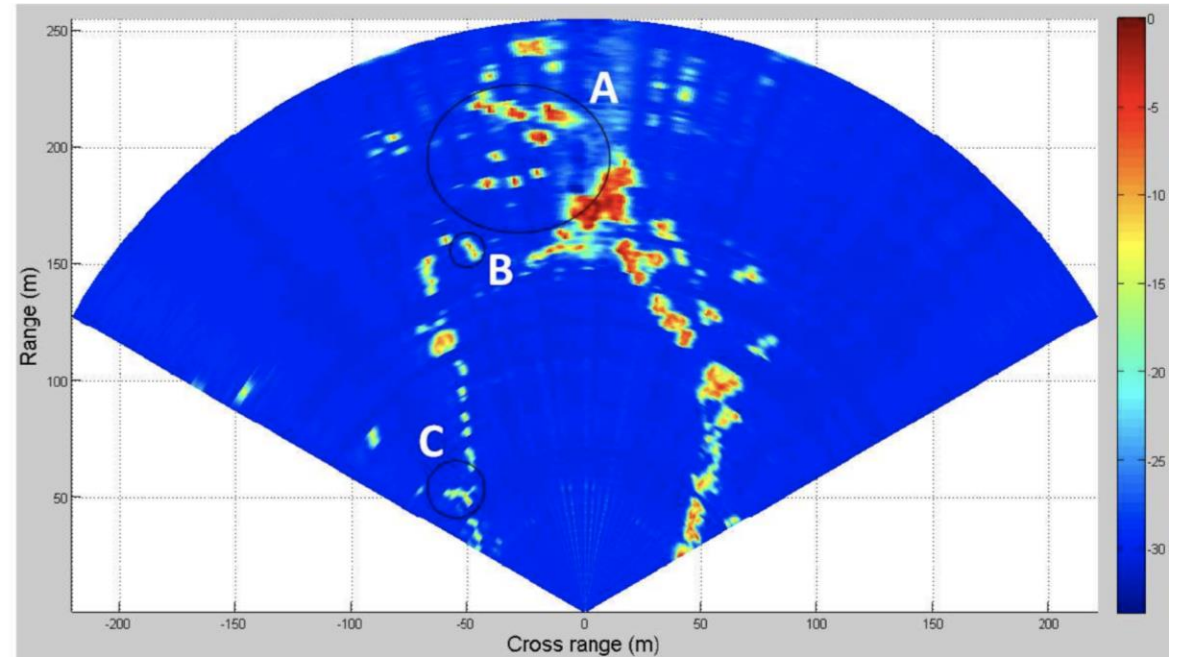
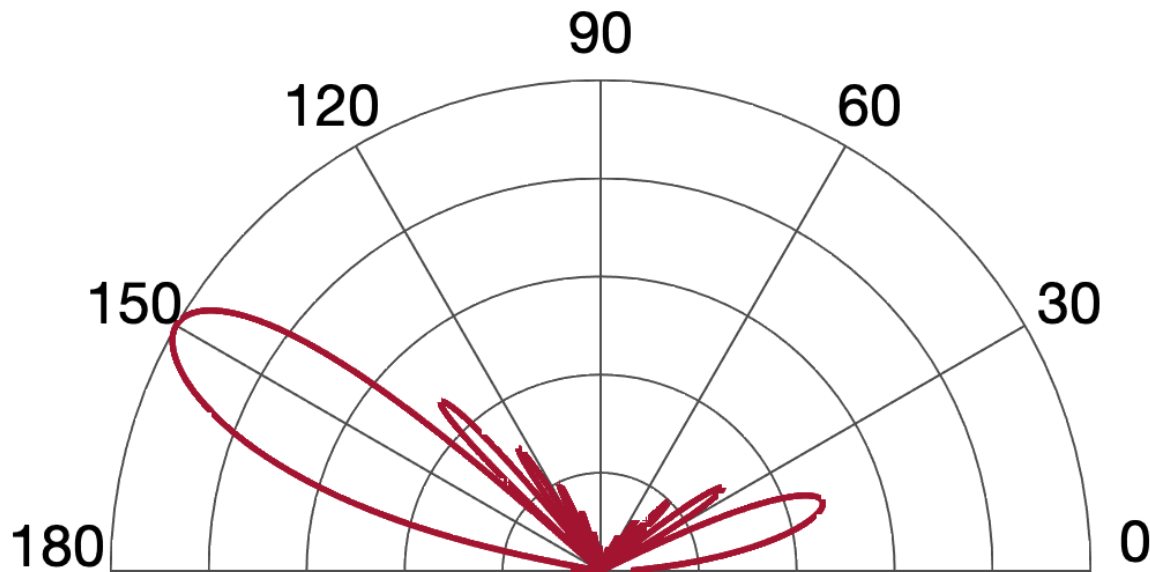
$$h_k = \alpha e^{-j2\pi \frac{d - k s \cos \theta}{\lambda}}$$



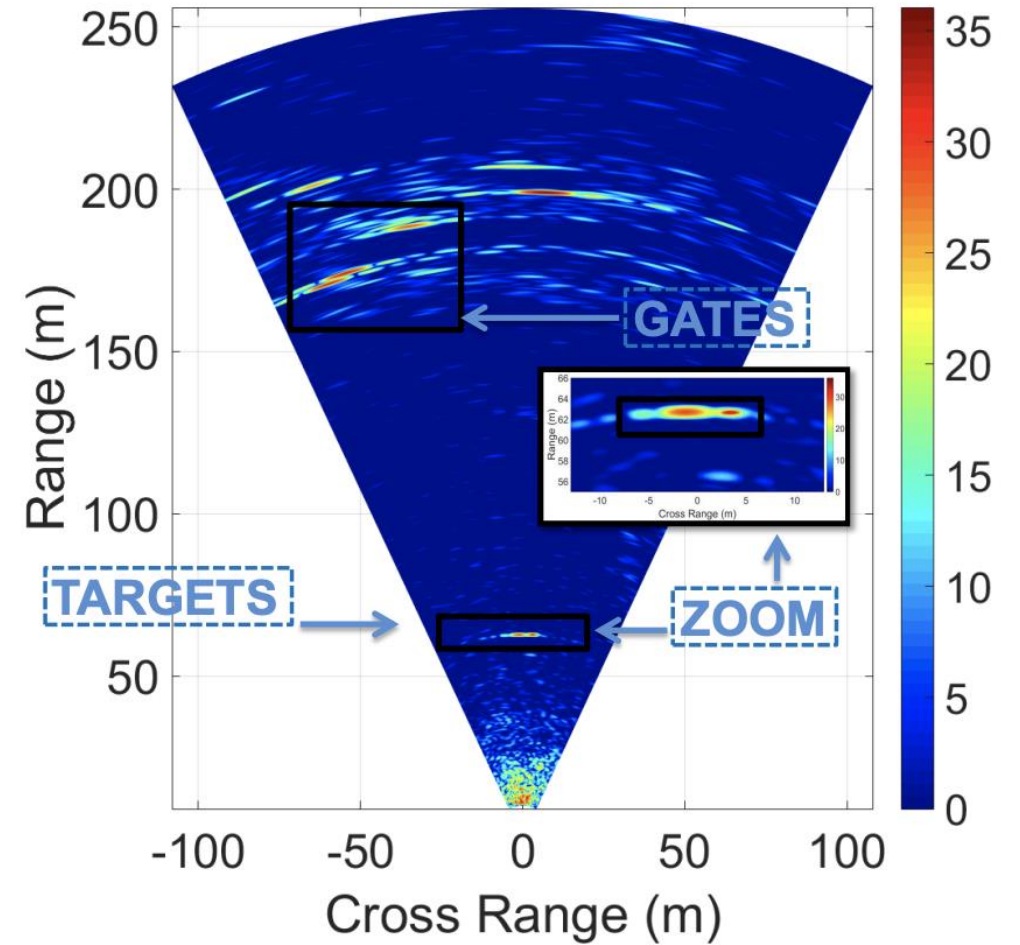
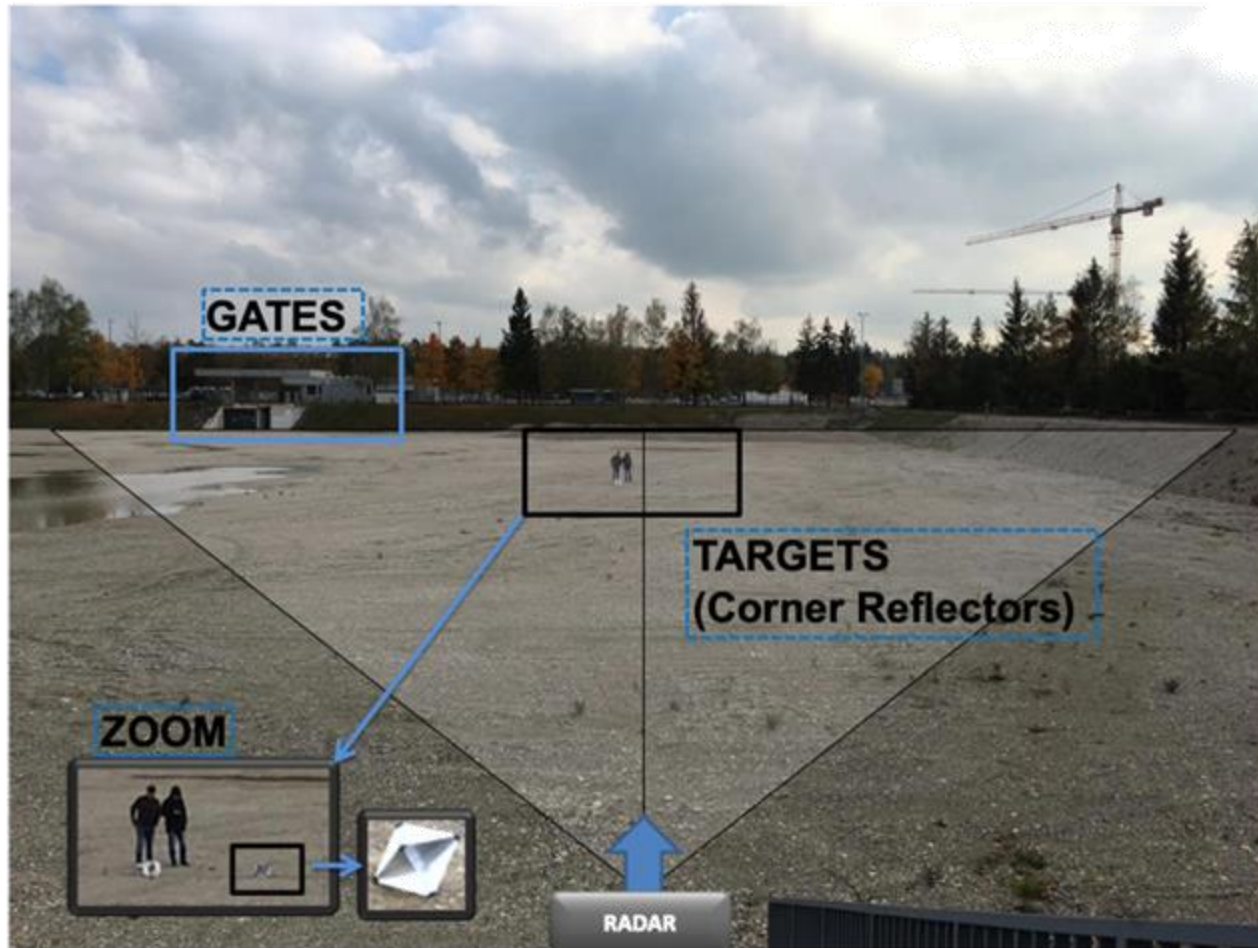
# Analog Phased Arrays



$$h_k = \alpha e^{-j2\pi \frac{d - k s \cos \theta}{\lambda}}$$

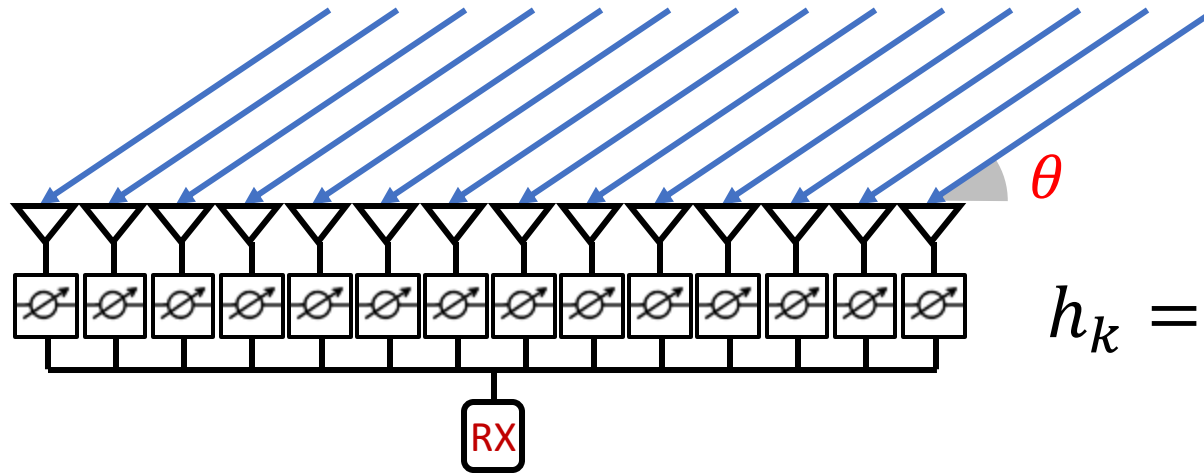


# Analog Phased Arrays

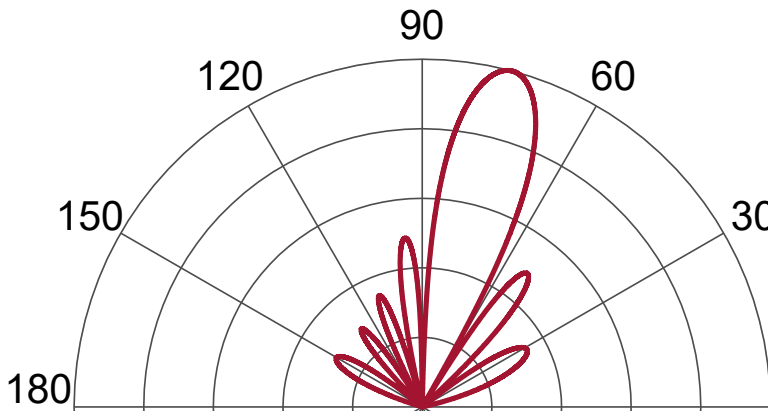


Ganis, A.; Miralles-Navarro, E.; Schoenlinner, B.; Prechtel, U.; Meusling, A.; Heller, C.; Spreng, T.... (2018). A portable 3D Imaging FMCW MIMO Radar Demonstrator with a 24x24 Antenna Array for Medium Range Applications. IEEE Transactions on Geoscience and Remote Sensing. 56(1):298-312. <https://doi.org/10.1109/TGRS.2017.2746739>

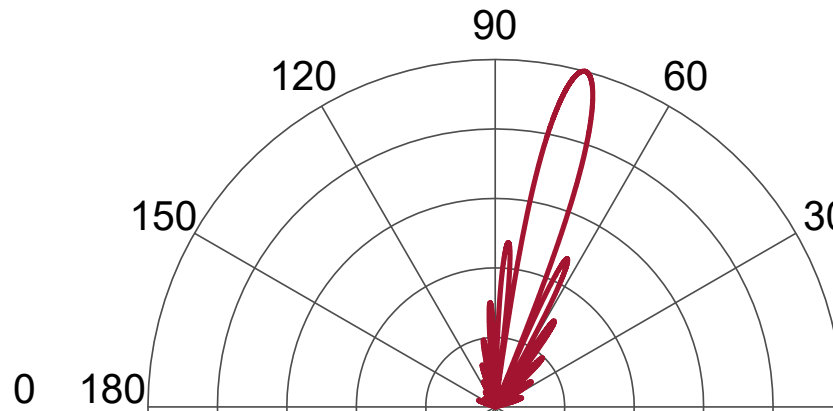
# Analog Phased Arrays



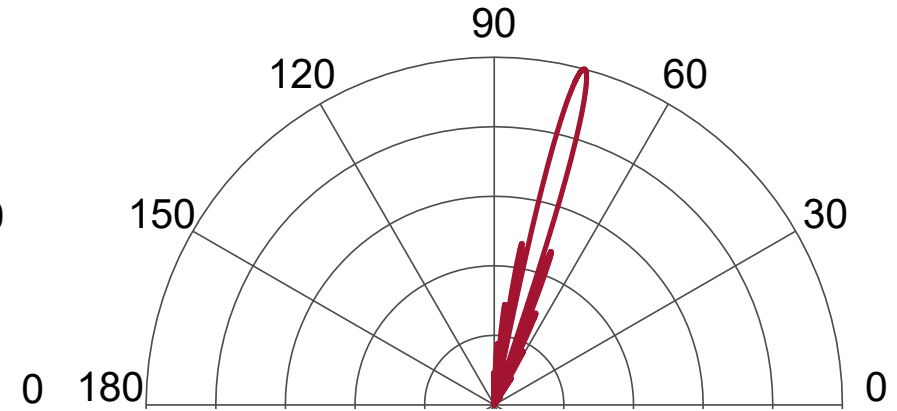
$$h_k = \alpha e^{-j2\pi \frac{d - k s \cos \theta}{\lambda}}$$



8 Antenna



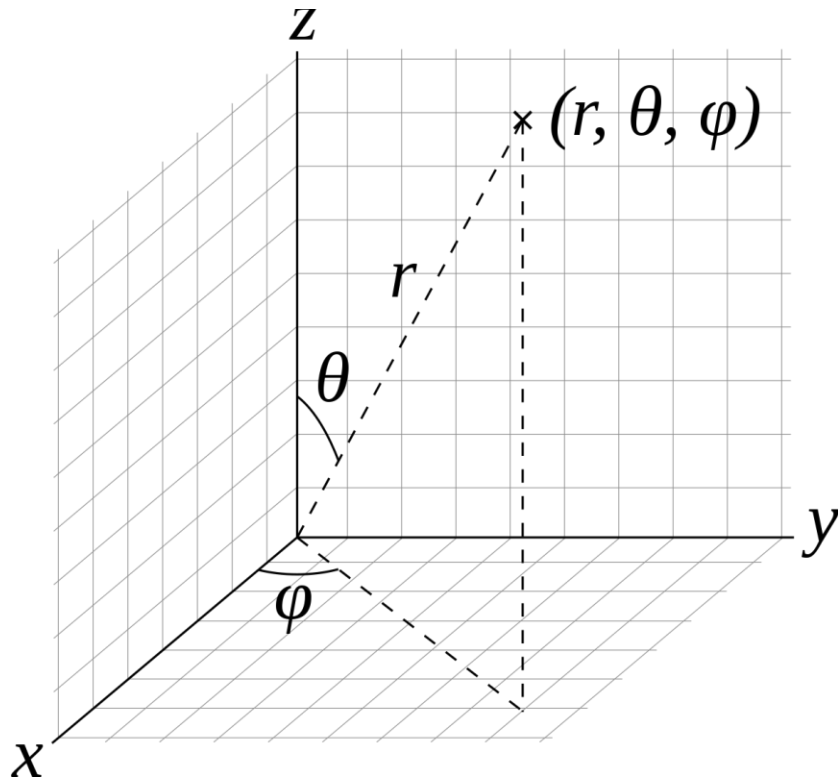
16 Antenna



32 Antenna

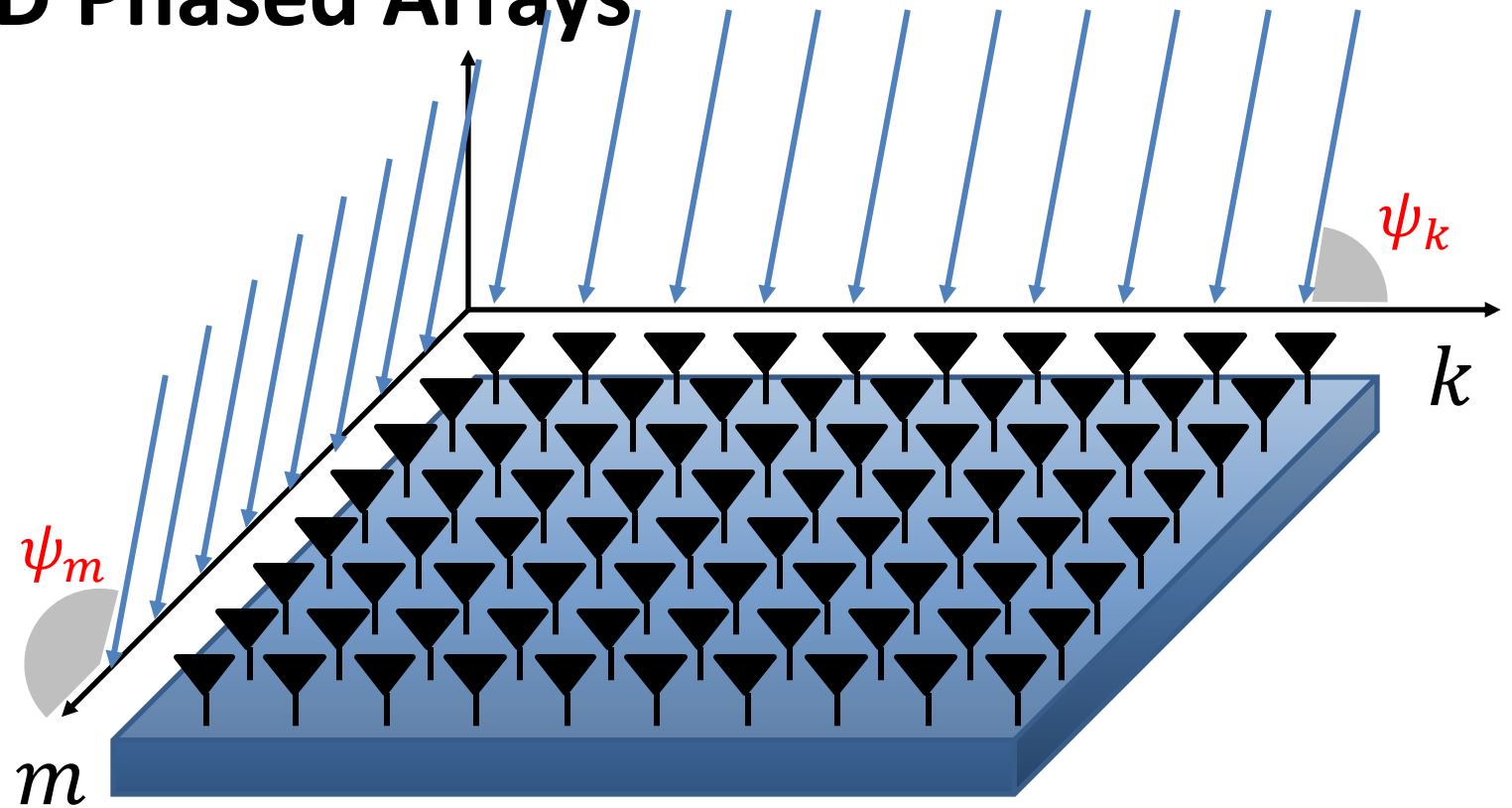
Larger Array → Narrower Beams → Higher Resolution

# 2D Phased Arrays



Can recover:

- Range:  $r$
- Azimuth AoA:  $\varphi$
- Elevation AoA:  $\theta$

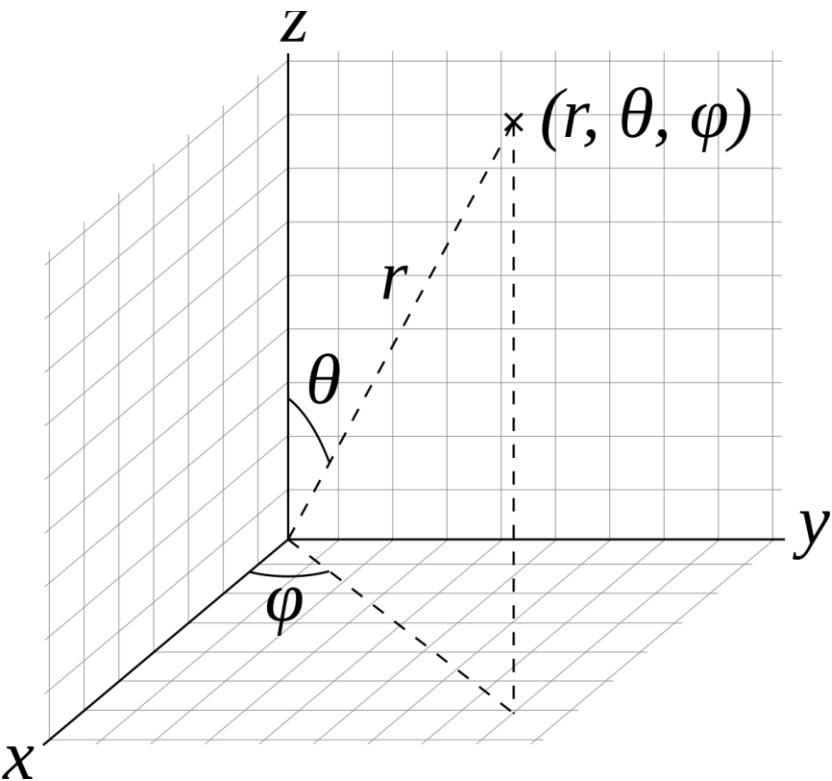


$$h_{m,k} = \alpha e^{-j\frac{2\pi}{\lambda} (r + f(m) + f(k))}$$

**Fix  $m$ :**  $h_{m,k} = \alpha e^{-j\frac{2\pi}{\lambda} (r + f(m) + ks \cos \psi_k)}$

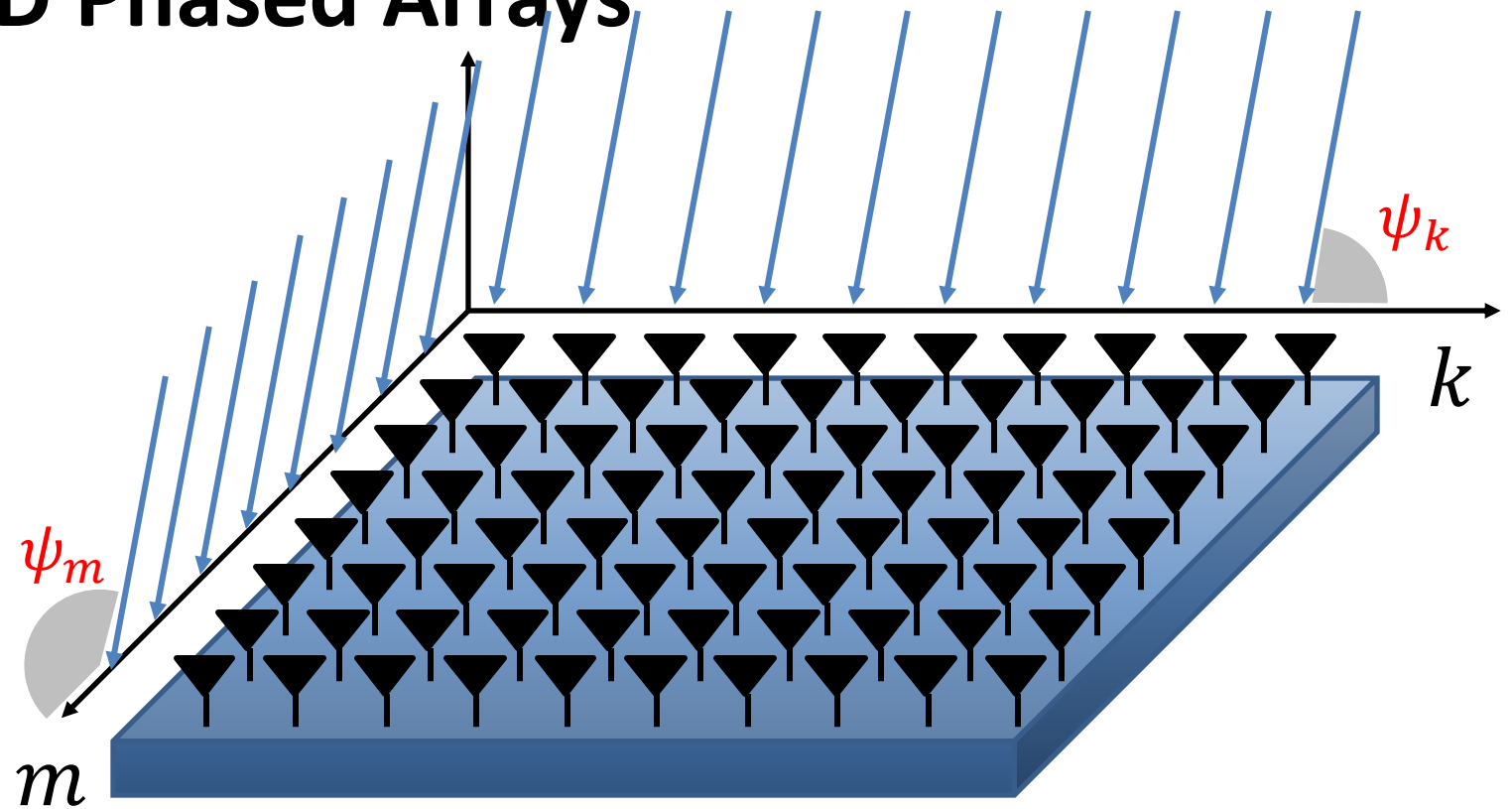
**Fix  $k$ :**  $h_{m,k} = \alpha e^{-j\frac{2\pi}{\lambda} (r + ms \cos \psi_m + ks \cos \psi_k)}$

# 2D Phased Arrays



Can recover:

- Range:  $r$
- Azimuth AoA:  $\varphi$
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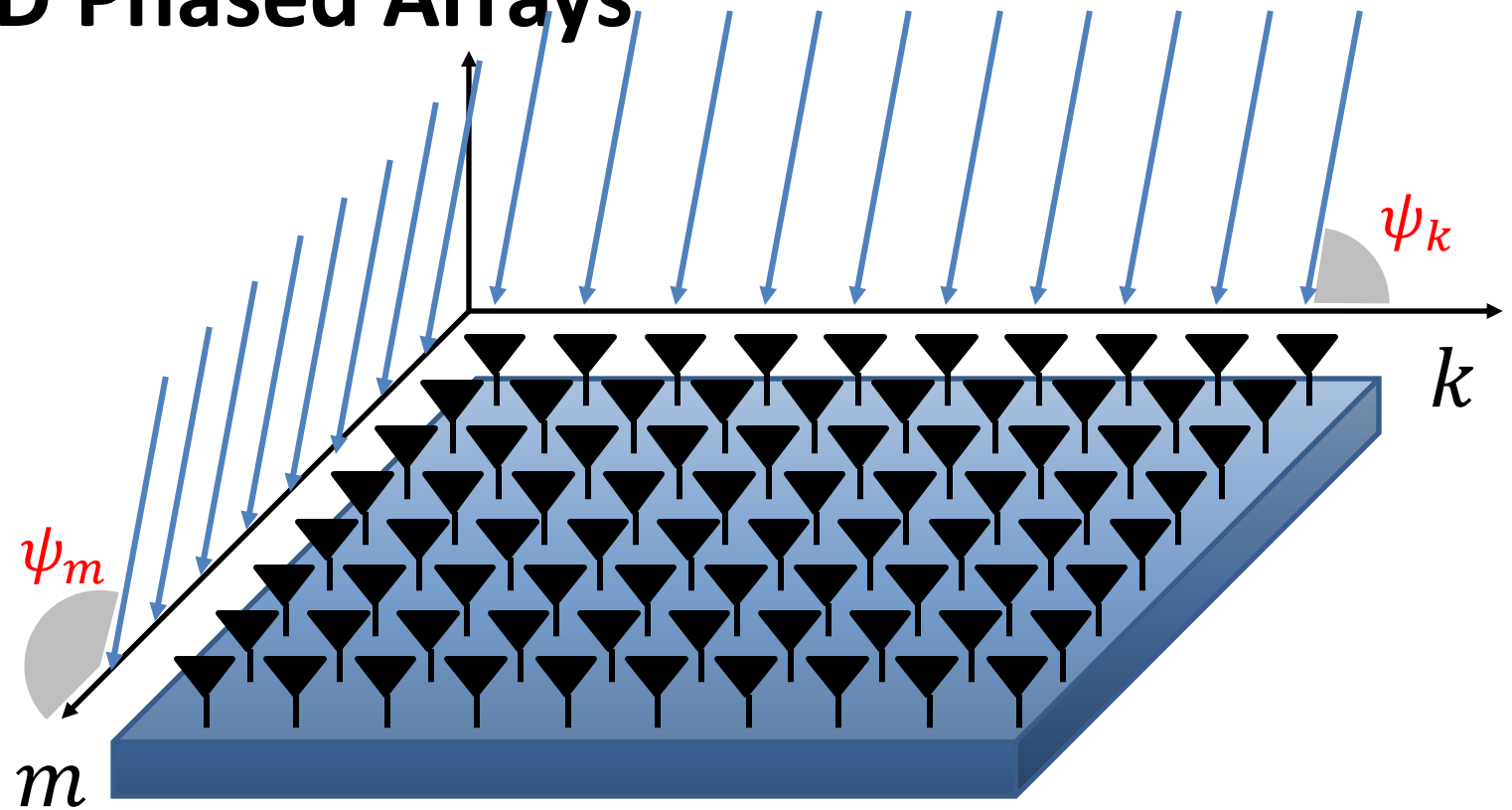
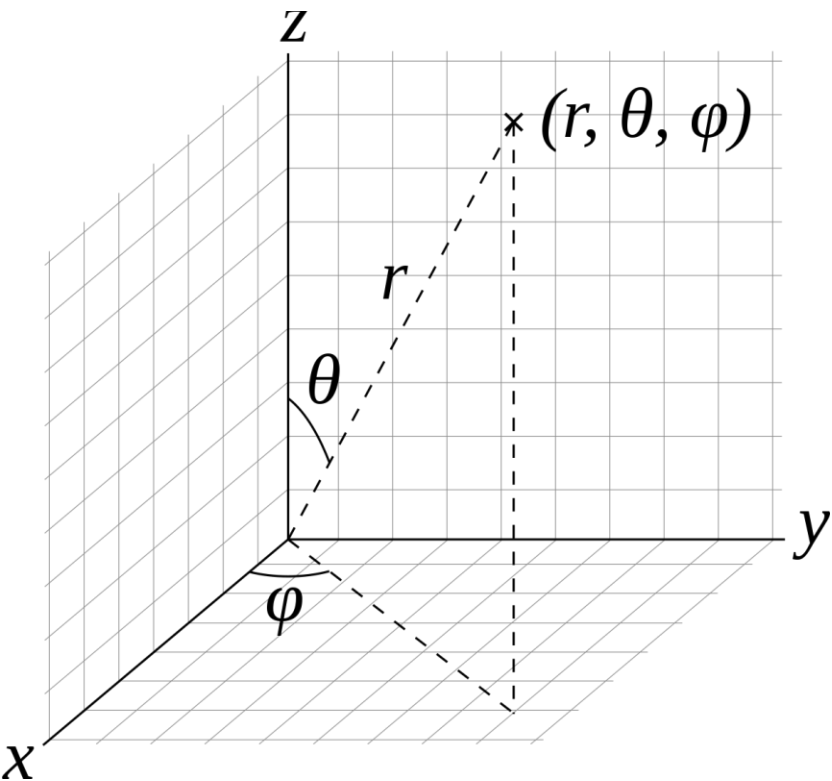


$$h_{m,k} = \alpha e^{-j\frac{2\pi}{\lambda} (r + ms \cos \psi_m + ks \cos \psi_k)}$$

$$\cos \psi_m = \sin \theta \cos \varphi$$

$$\cos \psi_k = \sin \theta \sin \varphi$$

# 2D Phased Arrays

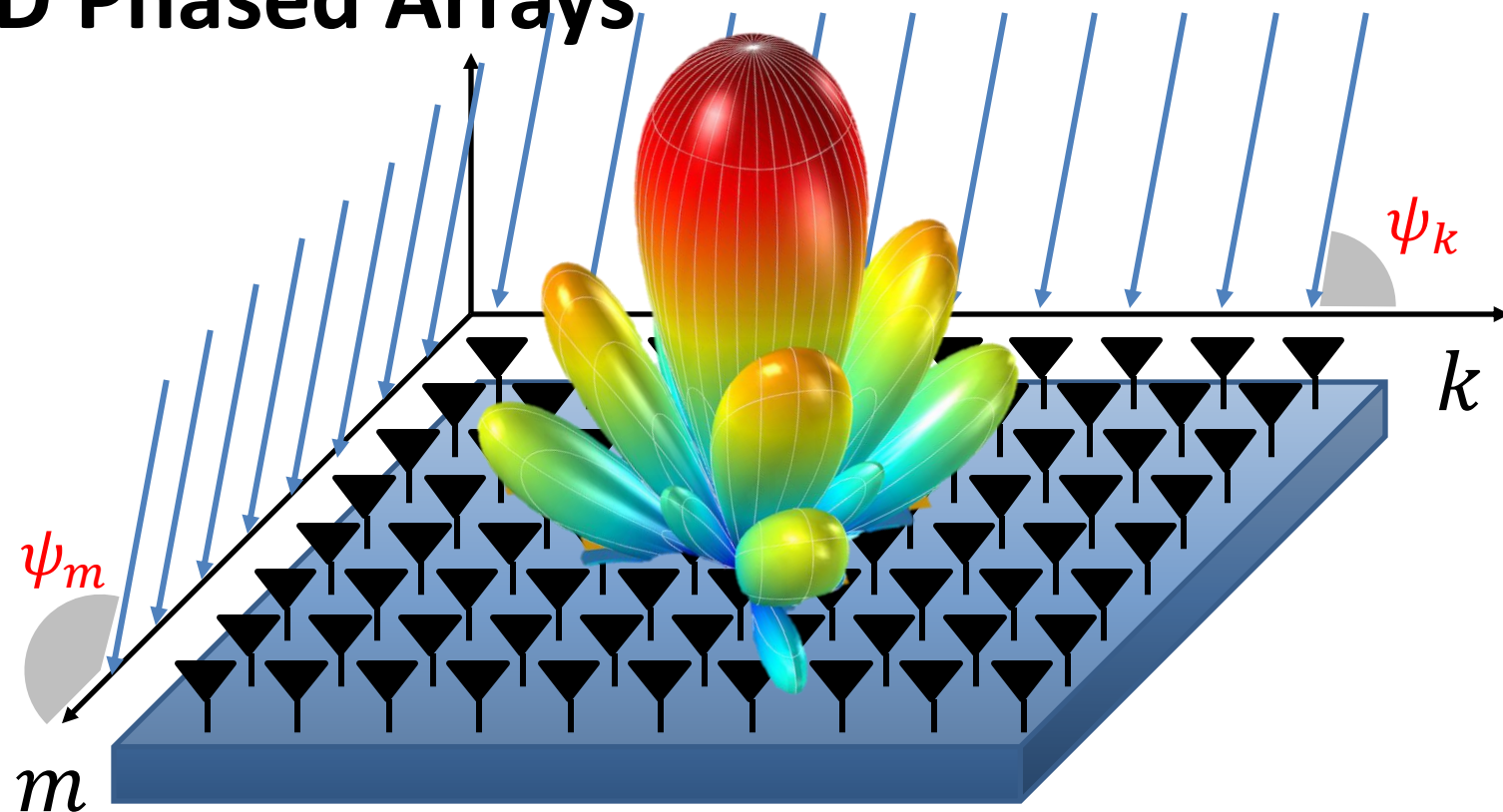
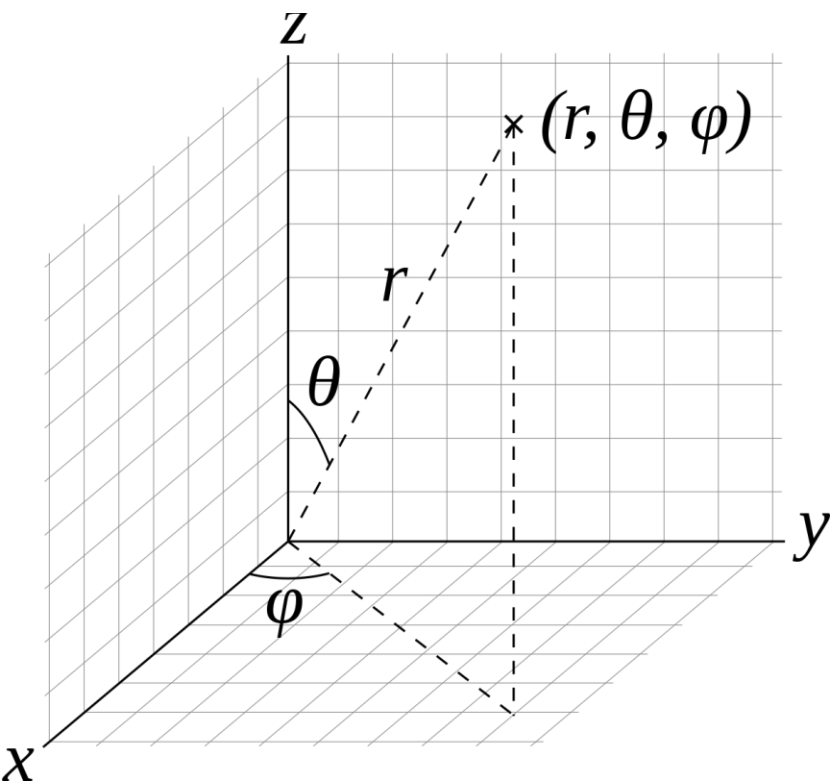


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# 2D Phased Arrays



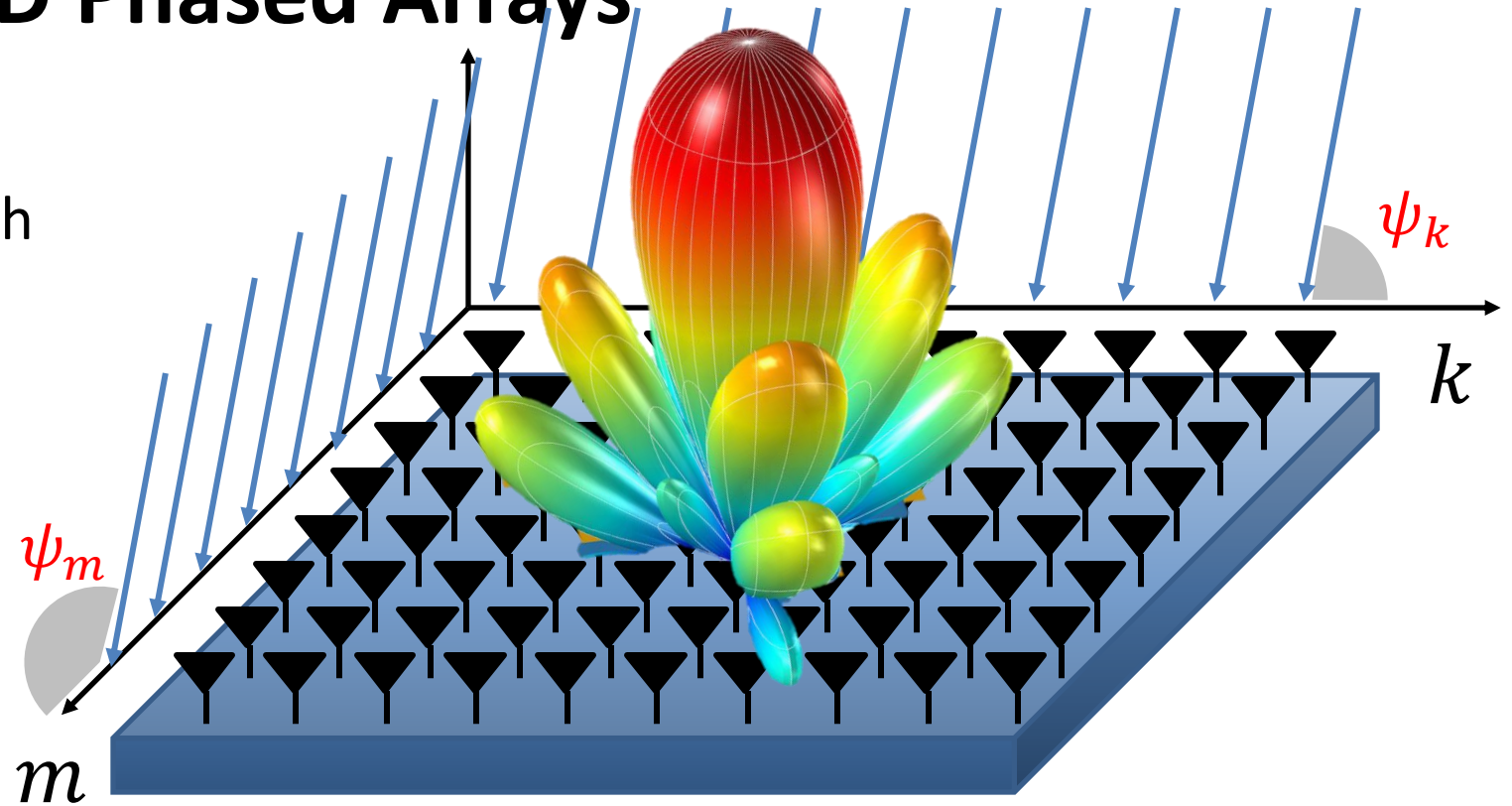
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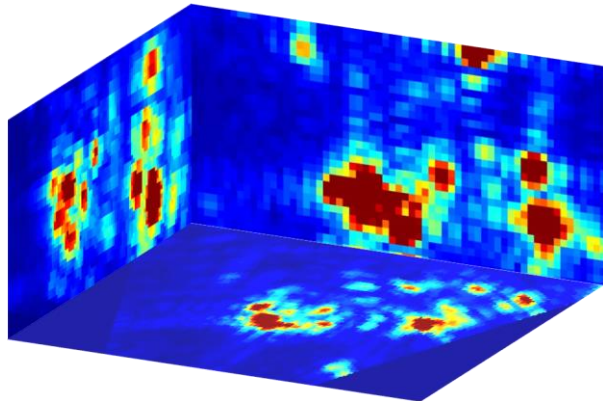
$$h_{m,k} = \alpha e^{-j\frac{2\pi}{\lambda} (r + ms \sin \theta \cos \varphi + ks \sin \theta \sin \varphi)}$$

# 2D Phased Arrays

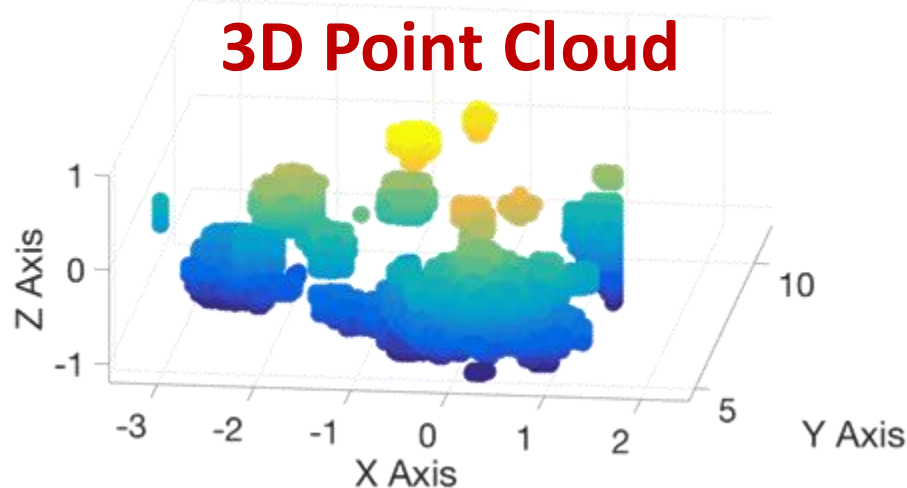
- 1) Pick the phase shift on each antenna to create a beam in each 3D direction.
- 2) Transmit FMCW signals and receive reflections.
- 3) Mix RX signal with TX and take range FFT .
- 4) Repeat in every direction.



**3D Heatmap Image**

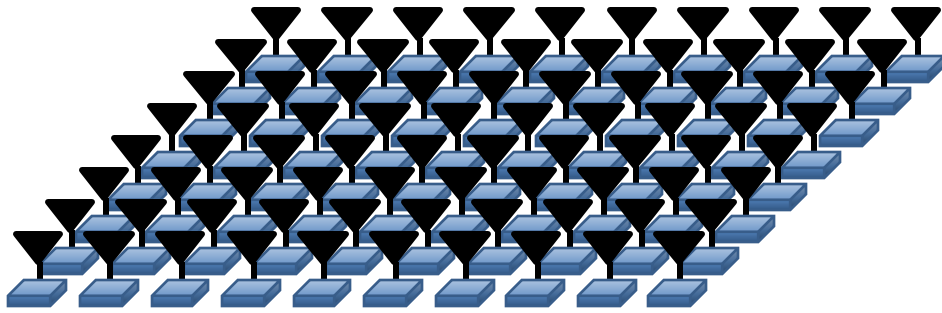


**3D Point Cloud**



# Phased Arrays Primer

## Digital Phased Arrays



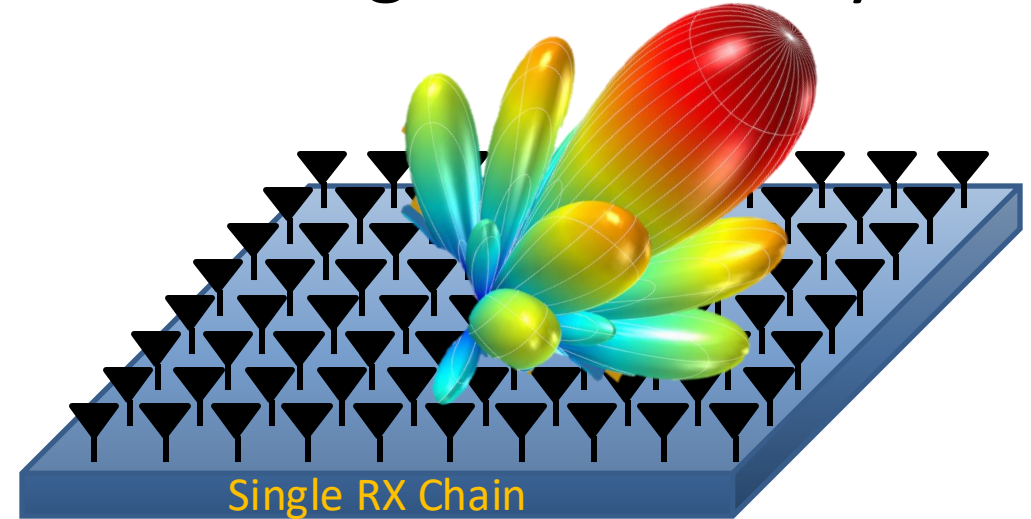
$N \times N$  RX Chains

**Could Potentially Do the Same thing:**

- 1) Mix the RX signal with TX.
- 2) Multiply the resulting signal on each antenna with  $e^{j\phi_{m,k}}$  and sum the signals.
- 3) Compute Range FFT.
- 4) Repeat in every direction.

$$(N \times N \times \mathbb{F}T \log T) \times N \times N = O(N^4T + N^2T \log T)$$

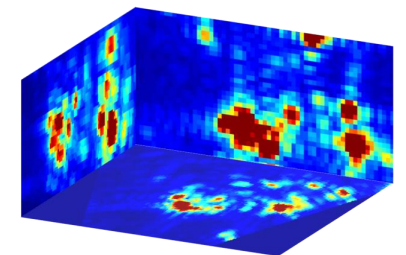
## Analog Phased Arrays



- 1) Pick the phase shift on each antenna to create a beam in each 3D direction.
- 2) Transmit FMCW signals and receive reflections.
- 3) Mix RX signal with TX and take range FFT .
- 4) Repeat in every direction.

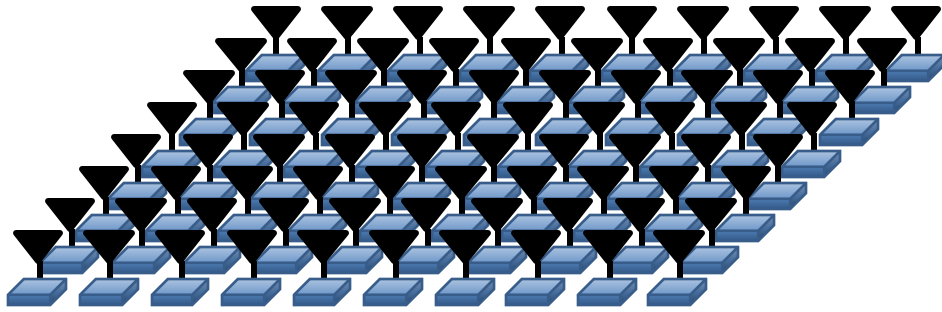


**3D Heatmap Image**



# Phased Arrays Primer

## Digital Phased Arrays

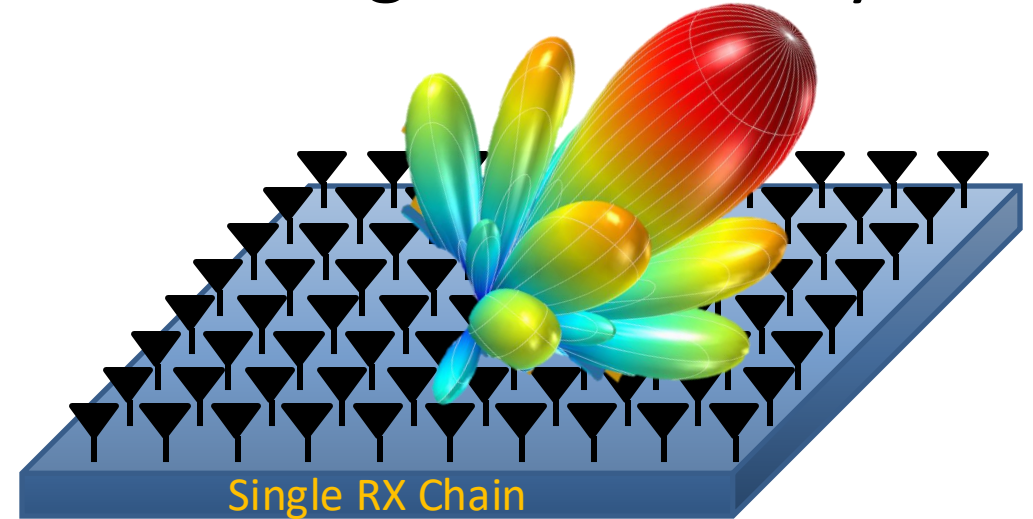


$N \times N$  RX Chains

**Could Potentially Do the Same thing:**

- 1) Mix the RX signal with TX.
- 3) Compute Range FFT.
- 2) Multiply the resulting signal on each antenna with  $e^{j\phi_{m,k}}$  and sum the signals.
- 4) Repeat in every direction.

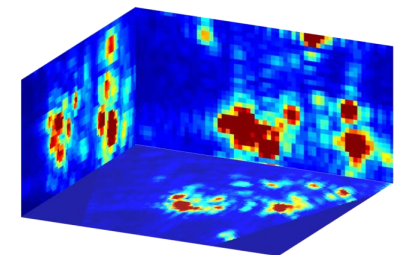
## Analog Phased Arrays



- 1) Pick the phase shift on each antenna to create a beam in each 3D direction.
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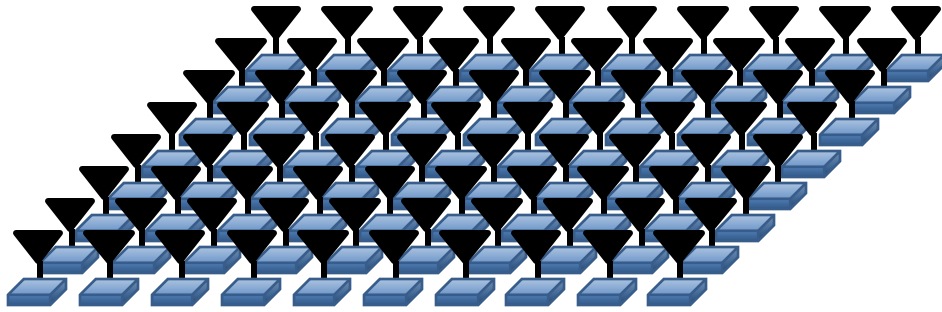
**3D Heatmap Image**



$$N \times N \times T \log T \neq N \times N \times \cancel{T} N \times N \neq O(N^4 T + N^2 T \log T)$$

# Phased Arrays Primer

## Digital Phased Arrays



$N \times N$  RX Chains

**Algorithm 1:**  $O(N^4T + N^2T \log T)$

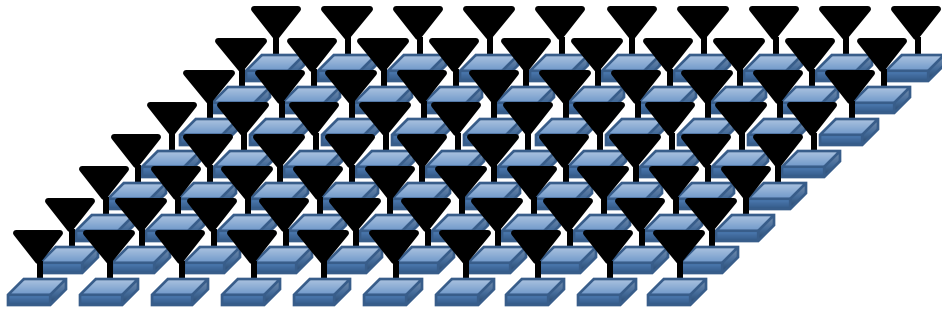
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- 3) Compute Range FFT.
- 4) Repeat in every direction.

**Algorithm 2: (Faster)** IDEA: antenna arrays are Fourier Transforms

$$h_{m,k} = \alpha e^{-j\frac{2\pi}{\lambda} (r+ms \sin \theta \cos \varphi + ks \sin \theta \sin \varphi)}$$

# Phased Arrays Primer

## Digital Phased Arrays



$N \times N$  RX Chains

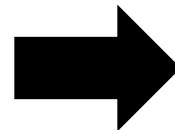
**Algorithm 1:**  $O(N^4T + N^2T \log T)$

- 1) Mix the RX signal with TX.
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- 4) Repeat in every direction.

**Algorithm 2: (Faster)** IDEA: antenna arrays are Fourier Transforms

$$h_{m,k} = \sum_{\theta_l} \sum_{\varphi_l} \alpha_l e^{-j\frac{2\pi}{\lambda} (r_l + ms \sin \theta_l \cos \varphi_l + ks \sin \theta_l \sin \varphi_l)}$$

$$h(x, y) = \sum_{f_x} \sum_{f_y} P(f_x, f_y) e^{-j2\pi (xf_x + yf_y)}$$

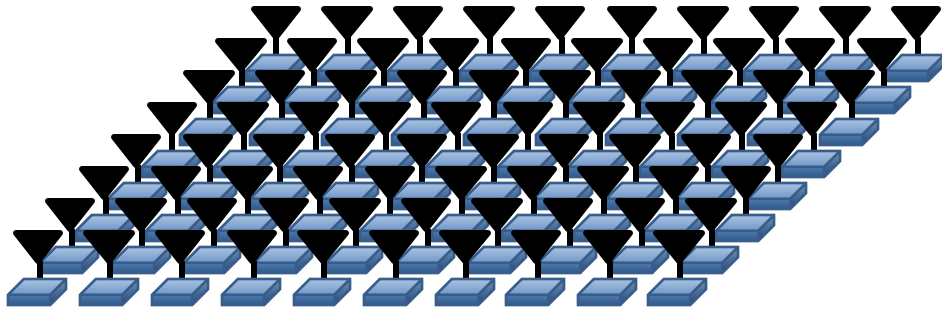


**Use 2D FFT**

$$\begin{aligned} f_x &= \sin(\theta_l) \cos(\varphi_l), \\ f_y &= \sin(\theta_l) \sin(\varphi_l), \\ x &= ms/\lambda, \\ y &= ks/\lambda \\ P(f_x, f_y) &= \alpha_l e^{-j\frac{2\pi r_l}{\lambda}} \end{aligned}$$

# Phased Arrays Primer

## Digital Phased Arrays



$N \times N$  RX Chains

**Algorithm 1:**  $O(N^4T + N^2T \log T)$

- 1) Mix the RX signal with TX.
- 2) Multiply the resulting signal on each antenna with  $e^{j\phi_{m,k}}$  and sum the signals.
- 3) Compute Range FFT.
- 4) Repeat in every direction.

**Algorithm 2: (Faster)** IDEA: antenna arrays are Fourier Transforms

1) Mix the RX signal with TX.

2) Compute 2D FFT across antennas

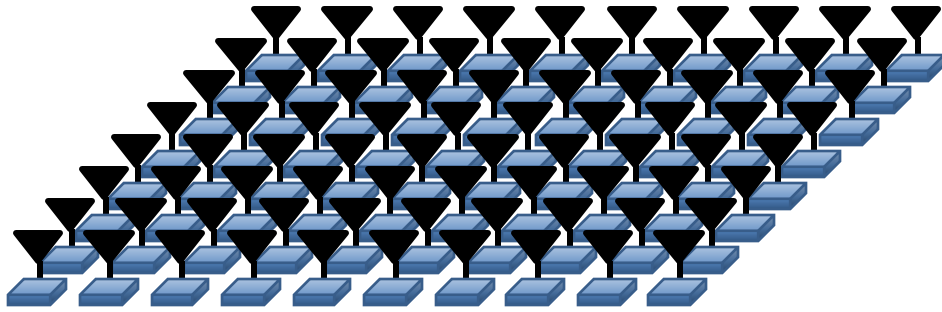
3) Compute Range FFT.

} **3D FFT**

$$N^2 \log N^2 \times T + N \times N \times T \log T = O(N^2T \log N^2T)$$

# Phased Arrays Primer

## Digital Phased Arrays



$N \times N$  RX Chains

### Algorithm 3: (More Accurate)

**Algorithm 1:**  $O(N^4T + N^2T \log T)$

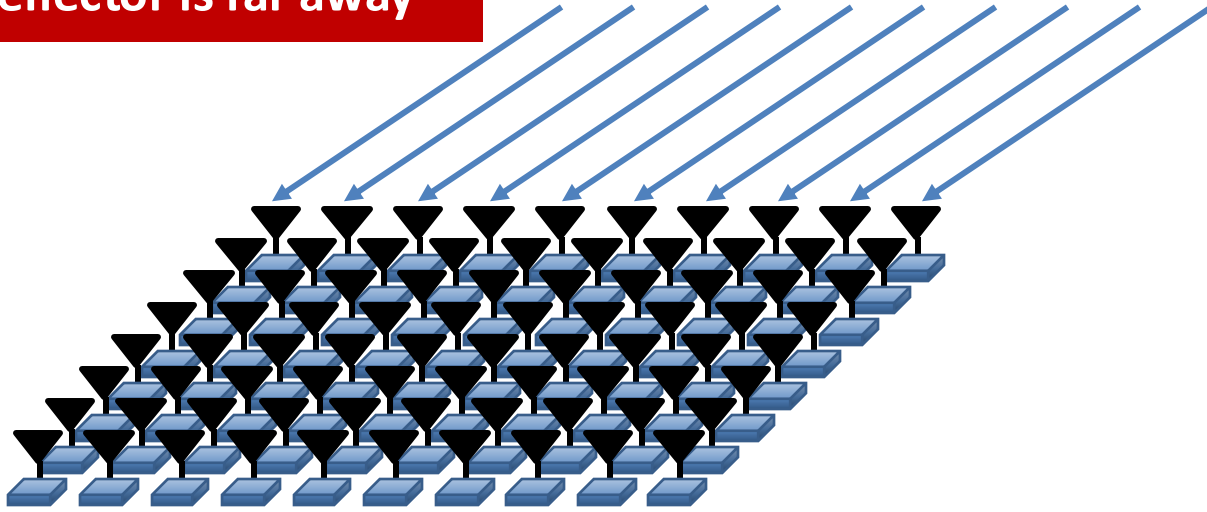
- 1) Mix the RX signal with TX.
- 3) Compute Range FFT.
- 2) Multiply the resulting signal on each antenna with  $e^{j\phi_{m,k}}$  and sum the signals.
- 4) Repeat in every direction.

**Algorithm 2: (Faster)**  $O(N^2T \log NT)$  **3D FFT**

- 1) Mix the RX signal with TX.
- 2) Compute 2D FFT across antennas
- 3) Compute Range FFT.

# Phased Arrays Primer

Assumes parallel waves  
Reflector is far away

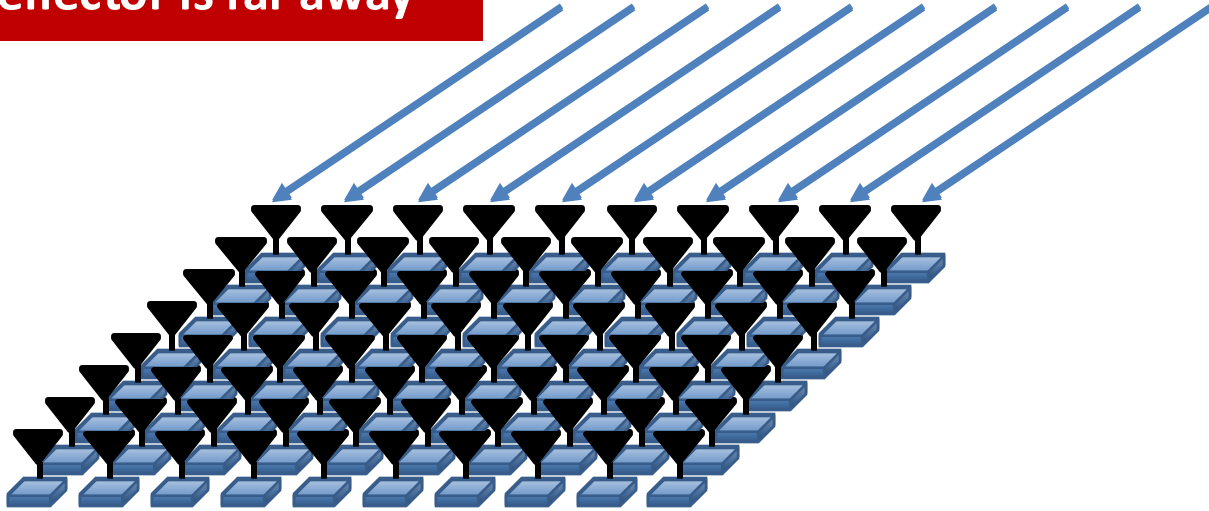


$N \times N$  RX Chains

Algorithm 3: (More Accurate)

# Phased Arrays Primer

Assumes parallel waves  
Reflector is far away

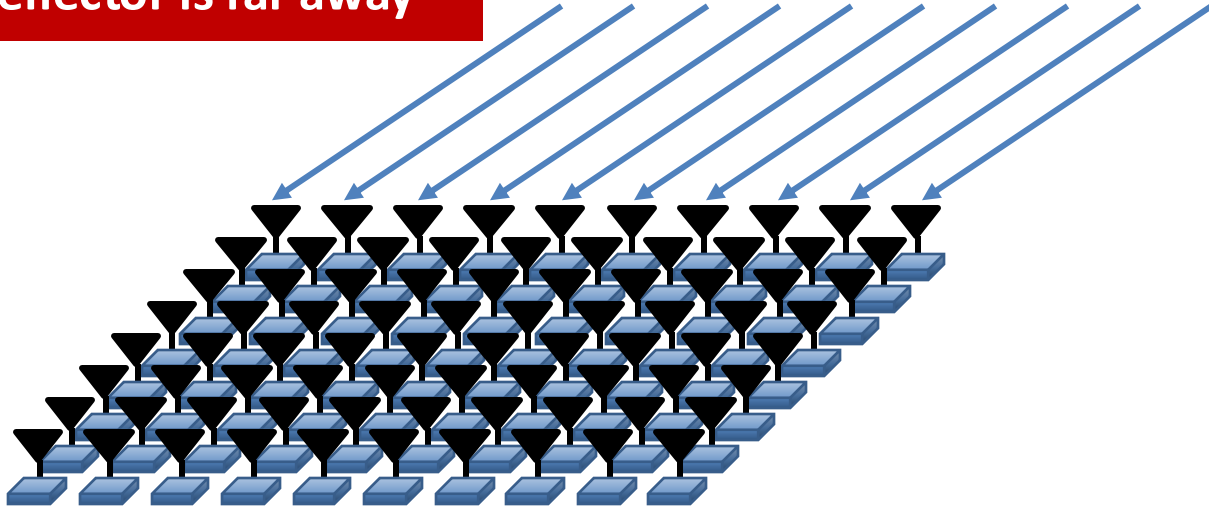


$N \times N$  RX Chains

Algorithm 3: (More Accurate)

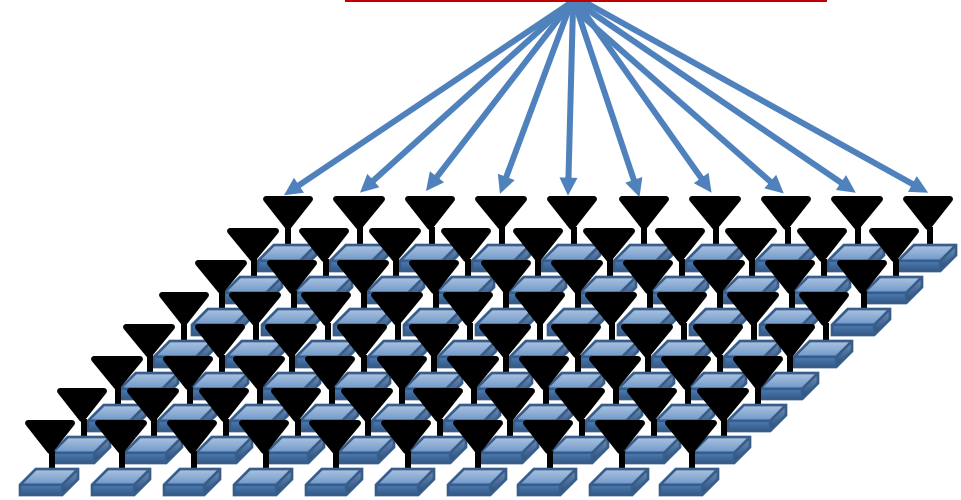
# Phased Arrays Primer

Assumes parallel waves  
Reflector is far away



$N \times N$  RX Chains

Reflector is not far



$N \times N$  RX Chains

**Algorithm 3: (More Accurate)** IDEA: use the exact equation

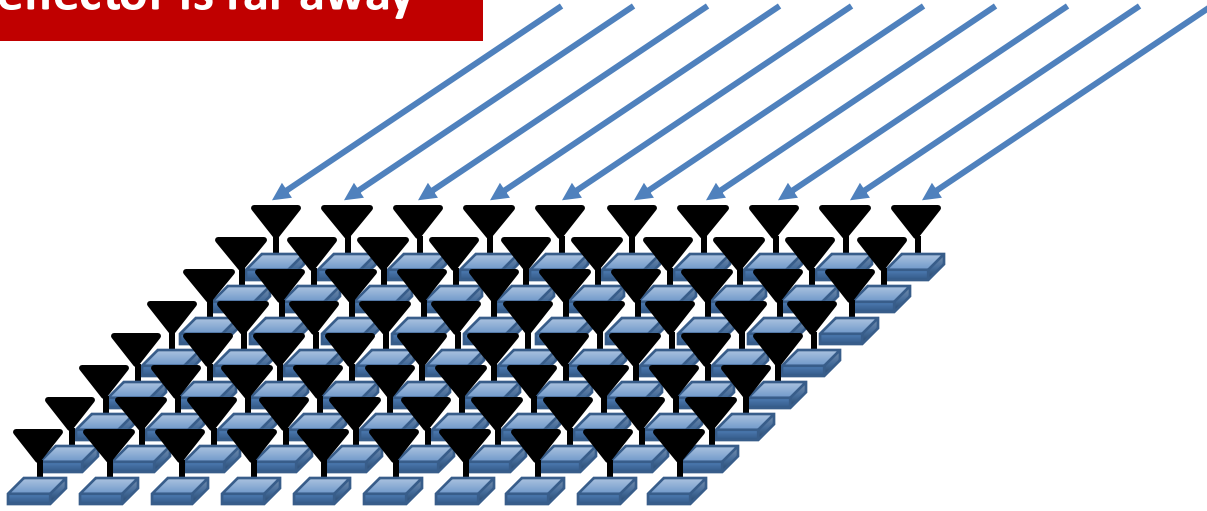
$p$ : FMCW Slope

$$s_{m,k}(t) = \alpha_l e^{-j2\pi(p\tau_l t + f_0\tau_l)} = \alpha_l e^{-j2\pi(p2d_l t/c + 2d_l/\lambda)}$$

Received FMCW signal from reflector  $l$  after mixing with TX

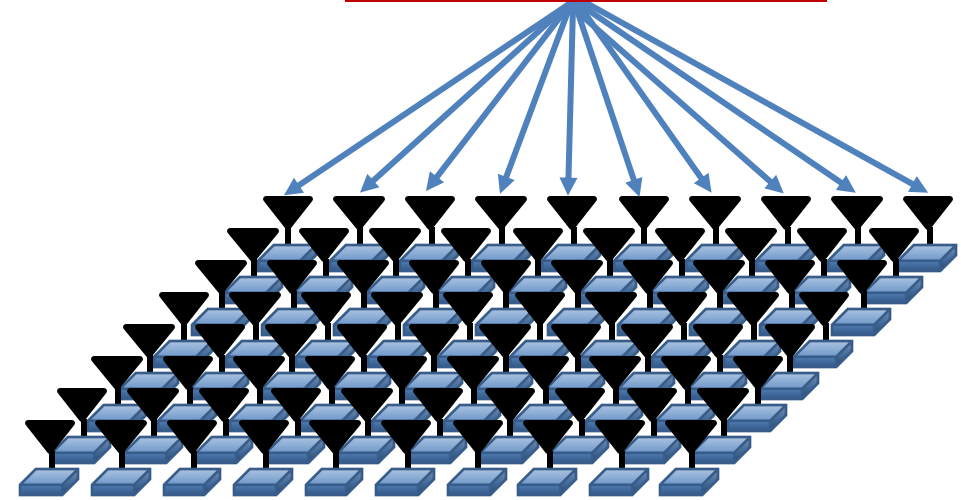
# Phased Arrays Primer

Assumes parallel waves  
Reflector is far away



$N \times N$  RX Chains

Reflector is not far



$N \times N$  RX Chains

**Algorithm 3: (More Accurate)** IDEA: use the exact equation

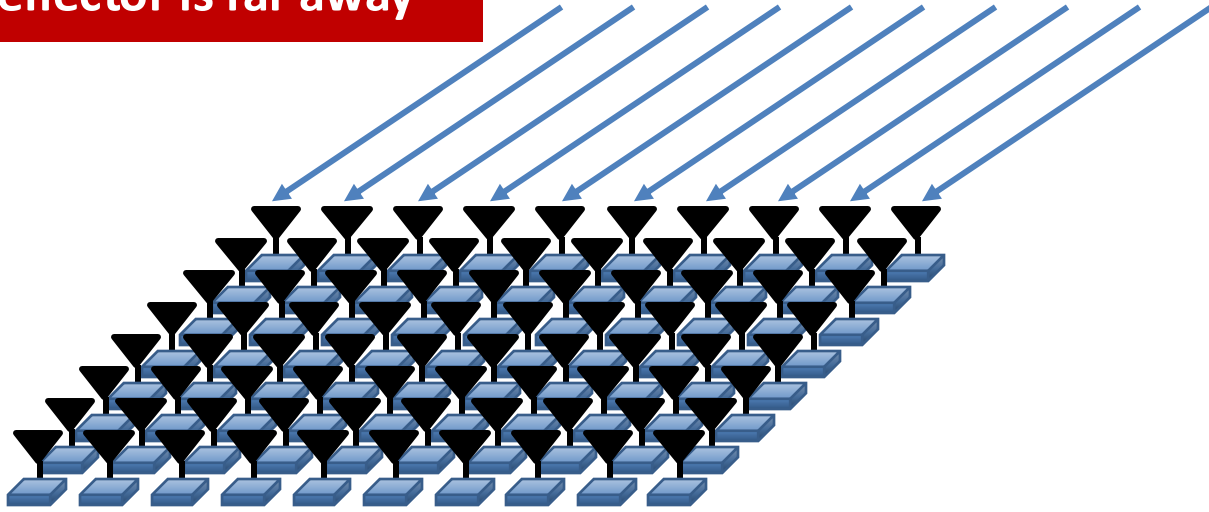
Assume antenna 0,0 at origin  
 $x_{m,k} = mS, y_{m,k} = kS, z_{m,k} = z_c$

$$s_{m,k}(t) = \alpha_l e^{-j2\pi(p\tau_l t + f_0\tau_l)} = \alpha_l e^{-j2\pi(p2d_l t/c + 2d_l/\lambda)}$$

$$d_l = \sqrt{(x_l - x_{m,k})^2 + (y_l - y_{m,k})^2 + (z_l - z_{m,k})^2} = \sqrt{(x_l - mS)^2 + (y_l - kS)^2 + (z_l - z_c)^2}$$

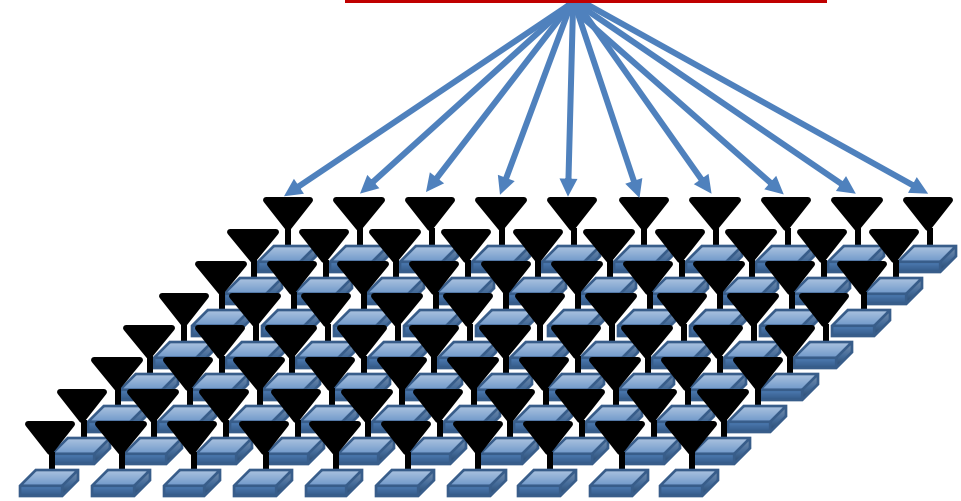
# Phased Arrays Primer

Assumes parallel waves  
Reflector is far away



$N \times N$  RX Chains

Reflector is not far



$N \times N$  RX Chains

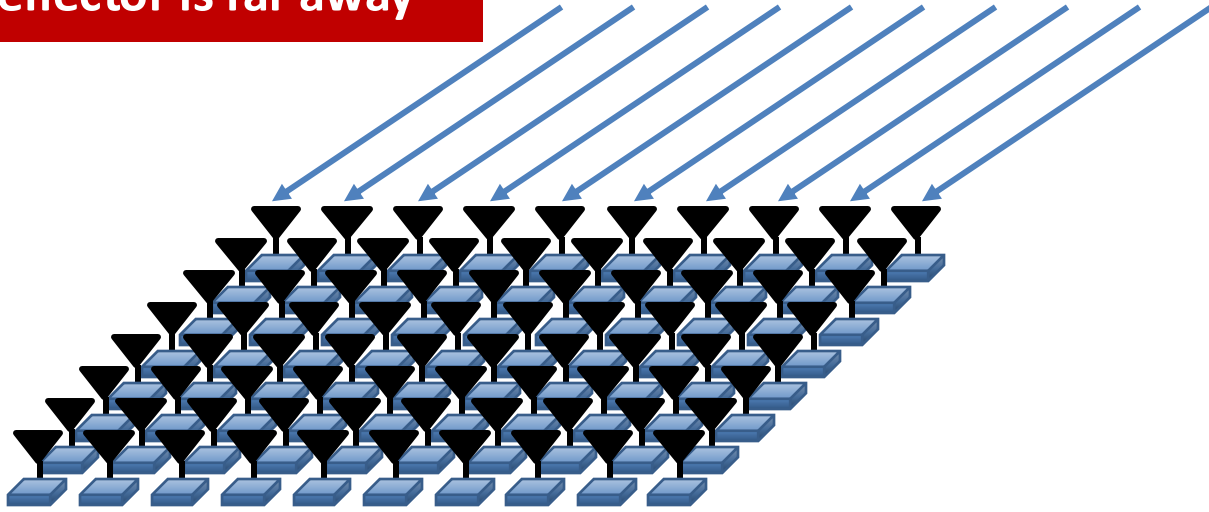
**Algorithm 3: (More Accurate)** IDEA: use the exact equation

Assume antenna 0,0 at origin  
 $x_{m,k} = mS, y_{m,k} = kS, z_{m,k} = z_c$

$$\begin{aligned} s_{m,k}(t) &= \alpha_l e^{-j2\pi(p\tau_l t + f_0 \tau_l)} = \alpha_l e^{-j2\pi(p2d_l t/c + 2d_l/\lambda)} \\ &= \alpha_l e^{-j4\pi(pt/c + 1/\lambda)\sqrt{(x_l - ms)^2 + (y_l - ks)^2 + (z_l - z_c)^2}} \end{aligned}$$

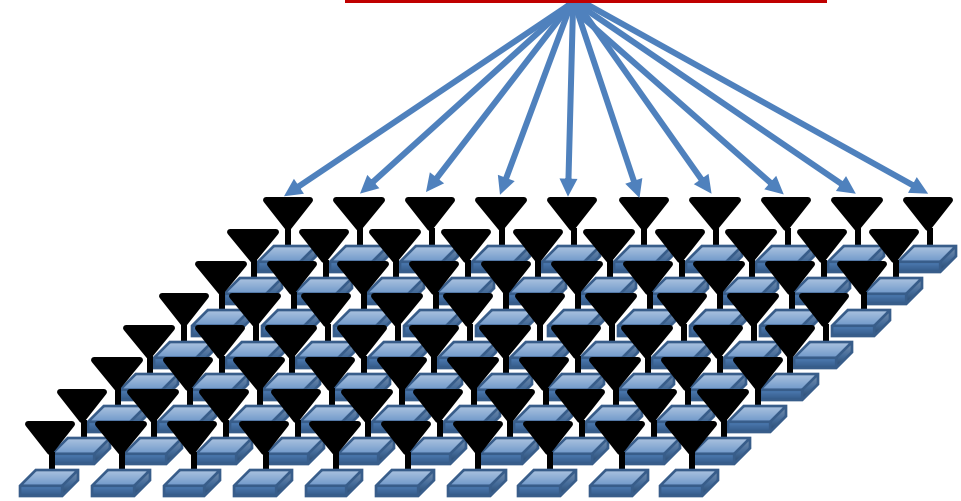
# Phased Arrays Primer

Assumes parallel waves  
Reflector is far away



$N \times N$  RX Chains

Reflector is not far



$N \times N$  RX Chains

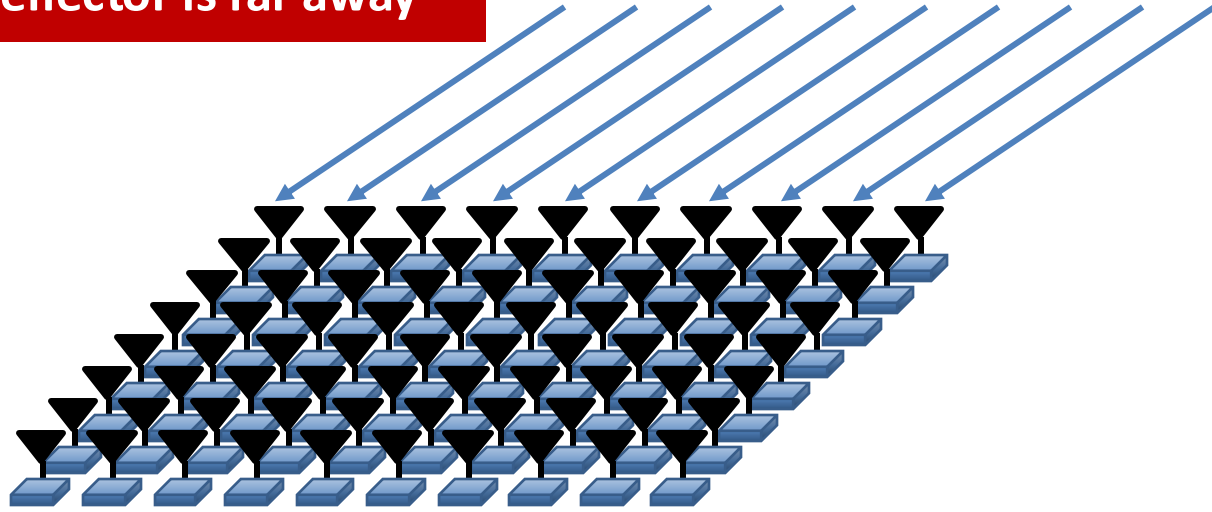
**Algorithm 3: (More Accurate)** IDEA: use the exact equation

Assume antenna 0,0 at origin  
 $x_{m,k} = ms, y_{m,k} = ks, z_{m,k} = z_c$

$$\begin{aligned}
 s_{m,k}(t) &= \sum_l \alpha_l e^{-j2\pi(p\tau_l t + f_0 \tau_l)} = \sum_l \alpha_l e^{-j2\pi(p2d_l t/c + 2d_l/\lambda)} \\
 &= \sum_l \alpha_l e^{-j4\pi(pt/c + 1/\lambda) \sqrt{(x_l - ms)^2 + (y_l - ks)^2 + (z_l - z_c)^2}}
 \end{aligned}$$

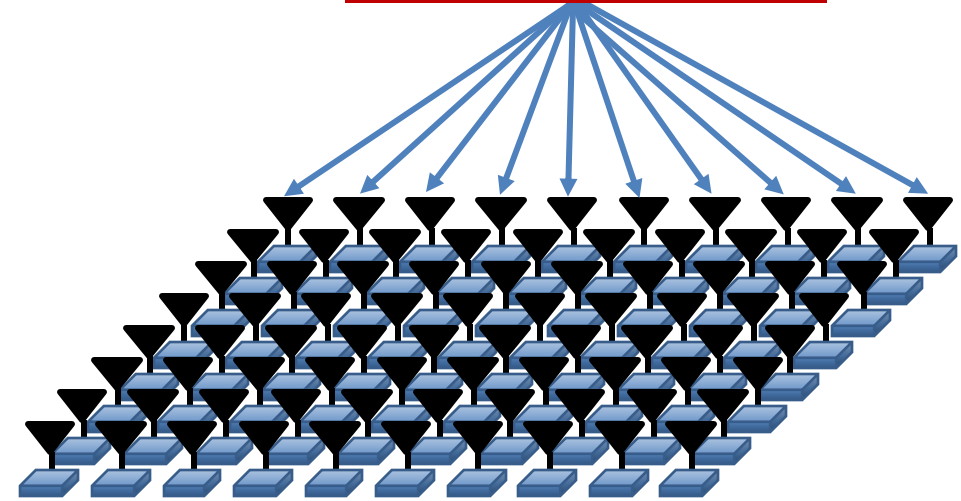
# Phased Arrays Primer

Assumes parallel waves  
Reflector is far away



$N \times N$  RX Chains

Reflector is not far



$N \times N$  RX Chains

**Algorithm 3: (More Accurate)** IDEA: use the exact equation

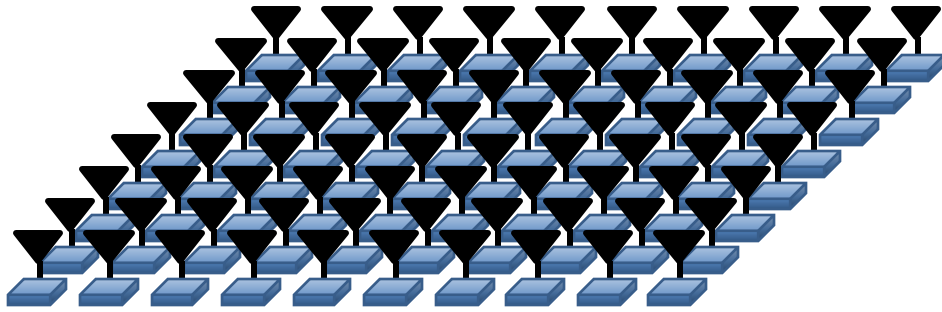
Assume antenna 0,0 at origin  
 $x_{m,k} = mS, y_{m,k} = kS, z_{m,k} = z_c$

$$s_{m,k}(t) = \sum_l \alpha_l e^{-j4\pi(pt/c + 1/\lambda) \sqrt{(x_l - ms)^2 + (y_l - ks)^2 + (z_l - z)^2}}$$

$$P(x, y, z) = \sum_m \sum_k \sum_t s_{m,k}(t) \times e^{j4\pi(pt/c + 1/\lambda) \sqrt{(x - ms)^2 + (y - ks)^2 + (z - z_c)^2}} = N^2 T \alpha_l$$

# Phased Arrays Primer

## Digital Phased Arrays



$N \times N$  RX Chains

**Algorithm 3: (More Accurate)**  $O(L^3 N^2 T) = o(N^5 T)$

- 1) Descritize space into  $L \times L \times L$  grid.
- 2) For each point in space compute the received signal using the below equation.

$$P(x, y, z) = \sum_m \sum_k \sum_t s_{m,k}(t) \times e^{j4\pi(kt/c + 1/\lambda)\sqrt{(x-ms)^2 + (y-ks)^2 + (z-z_c)^2}}$$

**Algorithm 1:**  $O(N^4 T + N^2 T \log T)$

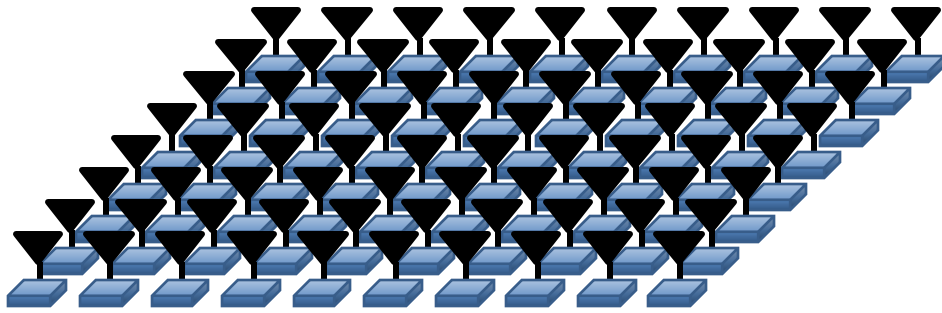
- 1) Mix the RX signal with TX.
- 2) Multiply the resulting signal on each antenna with  $e^{j\phi_{m,k}}$  and sum the signals.
- 3) Compute Range FFT.
- 4) Repeat in every direction.

**Algorithm 2: (Faster)**  $O(N^2 T \log NT)$  **3D FFT**

- 1) Mix the RX signal with TX.
- 2) Compute 2D FFT across antennas
- 3) Compute Range FFT.

# Phased Arrays Primer

## Digital Phased Arrays



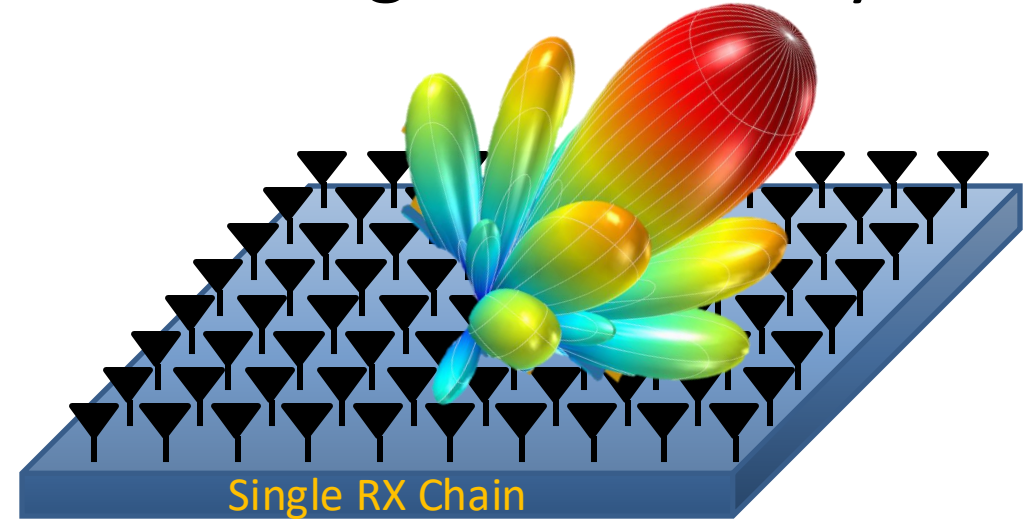
$N \times N$  RX Chains

**Algorithm 1:**  $O(N^4T + N^2T \log T)$  **Similar to Analog**

**Algorithm 2: (Faster)**  $O(N^2T \log NT)$  **3D FFT**

**Algorithm 3: (More Accurate)**  $O(L^3 N^2 T) = o(N^5 T)$

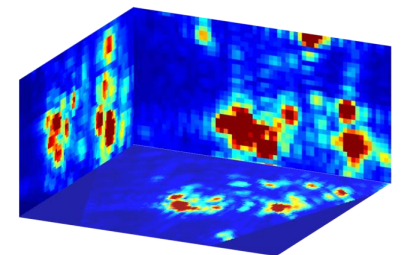
## Analog Phased Arrays



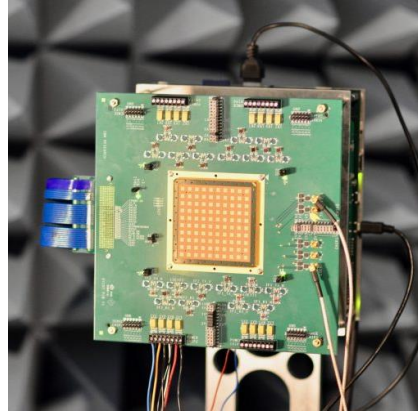
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- 3) Mix RX signal with TX and take range FFT .
- 4) Repeat in every direction.



**3D Heatmap Image**



# 2D Phased Arrays



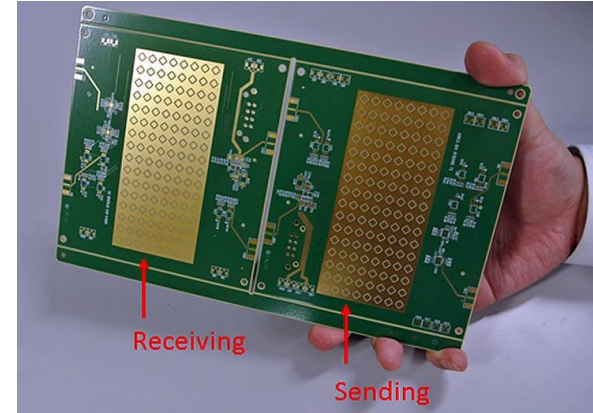
IBM

64 elements



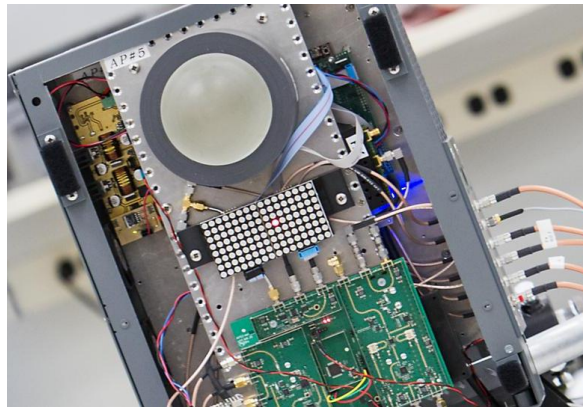
Anokiwave

256 elements

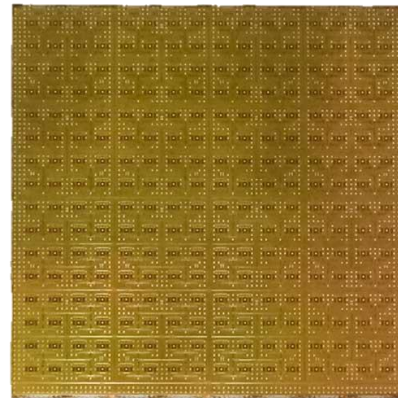


Fujitsu

64 elements

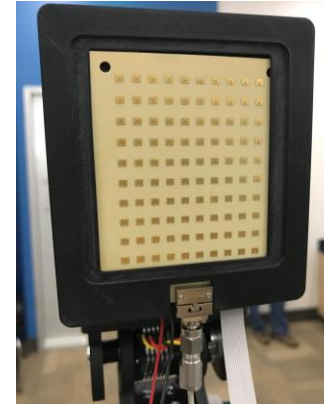


Nokia & National  
Instruments



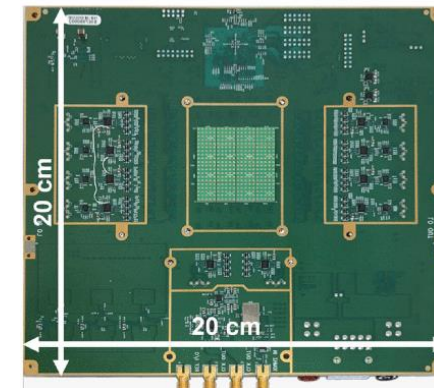
UCSD

256 elements



UCSD

64 elements



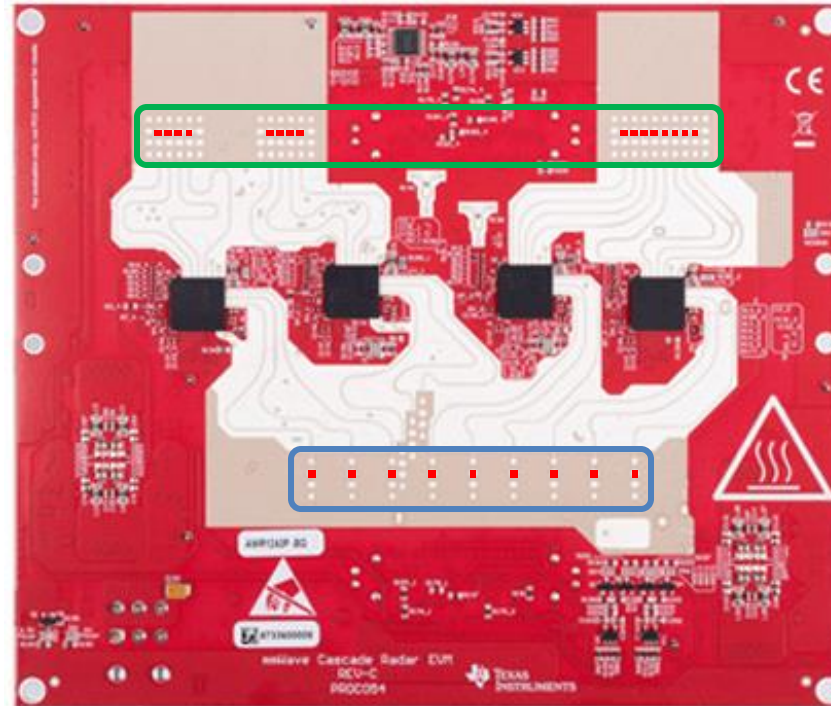
Bell Labs

384 elements

# MIMO Radar

Receiver  
Array  
(16 elements)

Transmitter  
Array  
(9 elements)



**TI**  
**MIMO Radar**

Angular resolution of radar is proportional to antenna array size.

# Virtual Antenna Array

Physical TX Array

Physical RX Array

Location



# Virtual Antenna Array

Physical TX Array

Physical RX Array

Location



# Virtual Antenna Array

Physical TX Array

Physical RX Array

Location



# Virtual Antenna Array

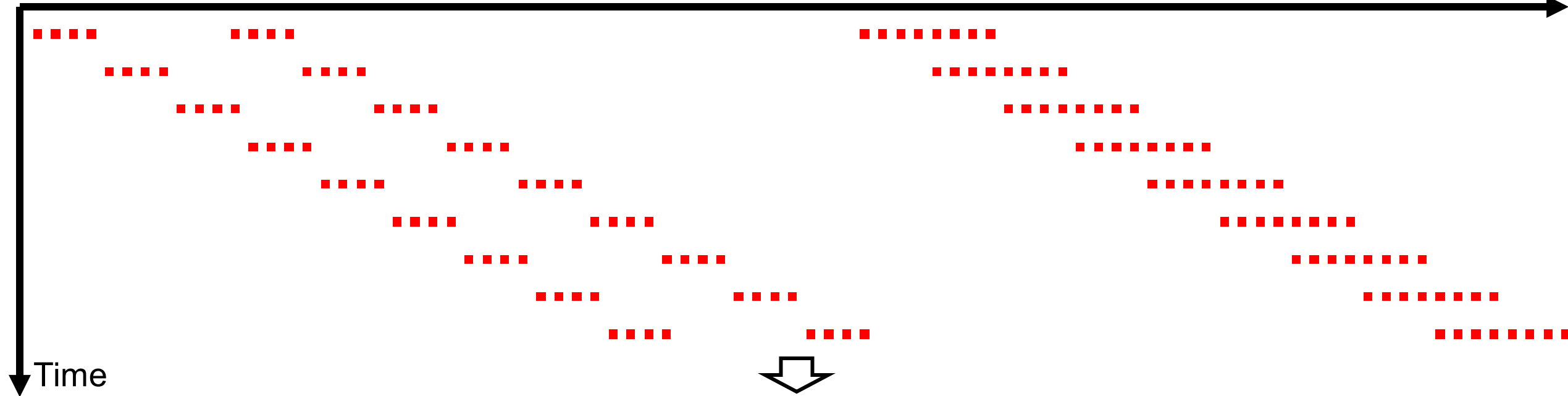
Physical TX Array



Physical RX Array



Location



86 element array

1.8° azimuth resolution

10x higher than single-chip radars

# MIMO Radar

Receiver  
Array  
(20 elements)

Transmitter  
Array  
(20 elements)

